# DAQ

# Lab-PC-1200/AI Register-Level Programmer Manual

Multifunction I/O Board for AT Bus Computers



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## **About This Manual**

Organization of This Manuali	х
Conventions Used in This Manualx	ζ.
National Instruments Documentationx	ci
Related Documentationx	cii
Customer Communicationx	cii

# Chapter 1 General Description

General Characteristics	1-1
Board Configuration Overview	1-2
Analog Input Configuration	
Analog Output Configuration (Lab-PC-1200 Only)	
Digital I/O Configuration	1-3
Counter Configuration	

# Chapter 2 Register Map and Descriptions

Register Map	2-1
Register Description Overview	2-3
Configuration and Status Register Group	2-4
Command Register 1	2-5
Command Register 2	2-7
Command Register 3	
Command Register 4	
Command Register 5	2-13
Command Register 6	
Status Register 1	
Status Register 2	
Analog Input Register Group	
A/D FIFO Register	
A/D FIFO Clear Register	
Start Convert Register	
DMATC Interrupt Clear Register	
Analog Output Register Group (Lab-PC-1200 Only)	
DAC0 Low-Byte, DAC0 High-Byte, DAC1 Low-Byte, and	
DAC1 High-Byte Registers	

82C53 Counter/Timer Register Groups A and B	
Counter A0 Data Register	
Counter A1 Data Register	
Counter A2 Data Register	
Counter A Mode Register	
Timer Interrupt Clear Register	
Counter B0 Data Register	
Counter B1 Data Register	
Counter B2 Data Register	
Counter B Mode Register	
82C55A Digital I/O Register Group	
Port A Register	
Port B Register	
Port C Register	
Digital Control Register	
Interval Counter Register Group	
Interval Counter Data Register	
Interval Counter Strobe Register	

# Chapter 3 Programming

Register Programming Considerations	. 3-1
Programming Examples	. 3-1
Lab-PC-1200/AI Companion Disk	. 3-2
Assigning Lab-PC-1200/AI Resources	3-2
Initializing the Lab-PC-1200/AI Circuitry	. 3-3
Programming the Analog Input Circuitry for Single A/D Conversions	3-4
Clearing the Analog Input Circuitry	. 3-4
Configuring the Analog Input Circuitry	. 3-5
Performing Single A/D Conversions	. 3-7
Programming a DAQ Operation Using Internal Timing	. 3-8
Programming Counter A0 and Counter B0	. 3-10
Programming Counter A1	. 3-11
Programming Counter B1 and the Interval Counter Register	. 3-11
Triggering the DAQ Operation	. 3-12
Servicing the DAQ Operation	. 3-12
Programming a DAQ Operation Using External Timing	3-13
Programming a DAQ Operation Using EXTCONV*	. 3-14
Programming a DAQ Operation Using EXTTRIG in Posttrigger Mode	3-14
Programming a DAQ Operation Using EXTTRIG in Pretrigger Mode	. 3-15
Programming a DAQ Operation Using OUTB1	. 3-15
DAQ Interrupt Programming	. 3-16

DAQ DMA Programming	.3-17
Programming the Analog Output Circuitry (Lab-PC-1200 Only)	.3-17
Configuring the Analog Output Circuitry	.3-17
Programming the Update Mode of the Analog Output Circuitry	.3-18
DAC Interrupt Programming	.3-20
Programming the Digital I/O Circuitry	.3-20
Programming the General-Purpose Counter/Timers	.3-21

# Chapter 4 Calibration

Storing User-Defined Constants	ĩ
Calibration DACs	
Analog Input Calibration	
Bipolar Input Calibration Procedure4-5	
Pregain Offset Coarse Calibration4-5	5
Pregain Offset Fine Calibration	5
Gain Calibration	5
Postgain Offset Calibration	5
Calibration at Higher Gains4-6	5
Unipolar Input Calibration Procedure4-6	5
Pregain Offset Calibration4-7	7
Gain Calibration	7
Postgain Offset Calibration	7
Analog Output Calibration (Lab-PC-1200 Only)	3
Bipolar Output Calibration Procedure4-8	3
Gain Calibration	)
Offset Calibration4-9	)
Unipolar Output Calibration Procedure4-9	)
Gain Calibration	0
Offset Calibration4-1	0
EEPROM Map4-1	0

# Appendix A Fujitsu MB88341/MB88342 Data Sheet

# Appendix B Xicor X25020 Data Sheet

## Appendix C OKI MSM82C53 Data Sheet

# Appendix D OKI MSM82C55A Data Sheet

## Appendix E Customer Communication

# Glossary

## Index

## Tables

Table 1-1.	Analog Input Settings	1-2
Table 1-2.	Analog Output Settings	1-2
Table 2-1.	Lab-PC-1200/AI Register Map	
Table 3-1.	Lab-PC-1200/AI Allowable Resources	3-2
Table 3-2.	Analog Output Voltage Versus Digital Code	
	(Unipolar Mode, Straight Binary Coding)	3-18
Table 3-3.	Analog Output Voltage Versus Digital Code	
	(Bipolar Mode, Two's Complement Coding)	3-18
Table 4-1.	Calibration DAC Characteristics for Analog Input Circuitry	4-3
Table 4-2.	Calibration DAC Characteristics for Analog Output Circuitry	4-3
Table 4-3.	Lab-PC-1200/AI EEPROM Map	4-11

This manual contains information about the internal operation and programming of the Lab-PC-1200/AI. The Lab-PC-1200 and Lab-PC-1200AI boards are low-cost multifunction analog, digital, and timing boards. The Lab-PC-1200/AI is a member of the National Instruments AT Series of expansion boards for AT/ISA bus computers. Additionally, the Lab-PC-1200 has two 12-bit DACs with voltage outputs. These boards are designed for high-performance data acquisition (DAQ) and control for applications in laboratory testing, production testing, and industrial process monitoring and control.

This manual assumes you are familiar with the *Lab-PC-1200/AI User Manual*. If you will be using National Instruments software with the Lab-PC-1200/AI, you do not need to read this manual. For information on the Lab-PC-1200/AI installation, signal connections, and theory of operation, consult your user manual.

# **Organization of This Manual**

The Lab-PC-1200/AI Register-Level Programmer Manual is organized as follows:

- Chapter 1, *General Description*, describes the general characteristics and gives a configuration overview of the Lab-PC-1200/AI.
- Chapter 2, *Register Map and Descriptions*, describes in detail the address and function of each of the Lab-PC-1200/AI registers.
- Chapter 3, *Programming*, contains programming instructions for operating the Lab-PC-1200/AI circuitry, and examples of the programming steps necessary to execute an operation.
- Chapter 4, *Calibration*, contains instructions for creating user-defined calibration constants for the Lab-PC-1200/AI CALDACs.
- Appendix A, *Fujitsu MB88341/MB88342 Data Sheet*, contains the manufacturer data sheet for the MB88341/MB88342 R-2R type 8-bit D/A converter manufactured by Fujitsu Microelectronics, Inc. The MB88341 D/A converter is used on the Lab-PC-1200/AI.
- Appendix B, *Xicor X25020 Data Sheet*, contains the manufacturer data sheet for the X25020 SPI serial EEPROM manufactured by Xicor, Inc. This EEPROM is used on the Lab-PC-1200/AI.

- Appendix C, *OKI MSM82C53 Data Sheet*, contains the manufacturer data sheet for the MSM82C53 CMOS programmable interval timer manufactured by OKI Semiconductor, Inc. This counter/timer is used on the Lab-PC-1200/AI.
- Appendix D, *OKI MSM82C55A Data Sheet*, contains the manufacturer data sheet for the MSM82C55A CMOS programmable peripheral interface manufactured by OKI Semiconductor, Inc. This interface is used on the Lab-PC-1200/AI.
- Appendix E, *Customer Communication*, contains a form you can use to comment on the product documentation. This appendix also contains information on how to access technical assistance for your National Instruments product.
- The *Glossary* contains an alphabetical list and description of terms used in this manual, including abbreviations, acronyms, metric prefixes, mnemonics, and symbols.
- The *Index* contains an alphabetical list of key terms and topics covered in this manual, including the page where you can find each one.

# **Conventions Used in This Manual**

The following conventions are used in this manual.

<>	Angle brackets containing numbers separated by an ellipsis represent a range of values associated with a bit or signal name—for example, DBIO<30>.
<u></u>	This icon to the left of bold italicized text denotes a note, which alerts you to important information.
1200 Series	1200 Series refers to both the Lab-PC-1200 and the Lab-PC-1200AI
bold	Bold text denotes the names of menus, menu items, dialog boxes, dialog box buttons or options.
bold italic	Bold italic text denotes a note, caution, or warning.
italic	Italic text denotes emphasis, a cross reference, or an introduction to a key concept.

monospace	Text in this font denotes text or characters that you should literally enter from the keyboard, sections of code, programming examples, and syntax examples. This font is also used for the proper names of disk drives, paths, directories, programs, subprograms, subroutines, device names, functions, operations, variables, filenames and extensions, and for statements and comments taken from programs.
NI-DAQ	NI-DAQ is used in this manual to refer to the NI-DAQ driver software, unless otherwise noted.
PC	PC refers to all PC compatible computers with PCI bus, unless otherwise noted.
SCXI	SCXI stands for Signal Conditioning eXtensions for Instrumentation and is a National Instruments product line designed to perform front-end signal conditioning for National Instruments plug-in DAQ boards.

# **National Instruments Documentation**

The *Lab-PC-1200/AI Register-Level Programmer Manual* is one piece of the documentation set for your DAQ system. You could have any of several types of manuals, depending on the hardware and software in your system. Use the different types of manuals you have as follows:

- *Getting Started with SCXI*—If you are using SCXI, this is the first manual you should read. It gives an overview of the SCXI system and contains the most commonly needed information for the modules, chassis, and software.
- Your SCXI hardware user manuals—If you are using SCXI, read these manuals next for detailed information about signal connections and module configuration. They also explain in greater detail how the module works and contain application hints.
- Your DAQ hardware user manuals—These manuals have detailed information about the DAQ hardware that plugs into or is connected to your computer. Use these manuals for hardware installation and configuration instructions, specification information about your DAQ hardware, and application hints.

- Software manuals—Examples of software manuals you may have are the LabVIEW or LabWindows<sup>™</sup>/CVI manual sets and the NI-DAQ manuals. After you set up your hardware system, use either the LabVIEW, LabWindows/CVI, or NI-DAQ manuals to help you write your application. If you have a large and complicated system, it is worthwhile to look through the software manuals before you configure your hardware.
- Accessory installation guides or manuals—If you are using accessory products, read the terminal block and cable assembly installation guides or accessory board user manuals. They explain how to physically connect the relevant pieces of the system. Consult these guides when you are making your connections.
- SCXI Chassis Manual—If you are using SCXI, read this manual for maintenance information on the chassis and installation instructions.

# **Related Documentation**

The following National Instruments document contains information that you may find helpful as you read this manual:

• Application Note 025, Field Wiring and Noise Considerations for Analog Signals

The following document also contains information that you may find helpful as you read this manual:

• Your computer's technical reference manual

# **Customer Communication**

National Instruments wants to receive your comments on our products and manuals. We are interested in the applications you develop with our products, and we want to help if you have problems with them. To make it easy for you to contact us, this manual contains a comment form for you to complete. This form is in Appendix E, *Customer Communication*, at the end of this manual.

# **General Description**

This chapter describes the general characteristics and gives a configuration overview of the Lab-PC-1200/AI.

# **General Characteristics**

Thank you for purchasing the Lab-PC-1200/AI, low-cost, high-performance multifunction analog, digital, and timing boards for AT/ISA bus computers. The Lab-PC-1200/AI boards have eight analog input channels that you can configure as eight single-ended or four differential inputs; a 12-bit successive-approximation ADC; two 12-bit DACs with voltage outputs; 24 lines of TTL-compatible digital I/O; and three 16-bit counter/timers for timing I/O. Additionally, the Lab-PC-1200 has two 12-bit DACs with voltage outputs.

The Lab-PC-1200/AI is a member of the National Instruments AT Series of expansion boards for AT/ISA bus computers. The 1200 Series boards are completely switchless and jumperless DAQ boards. This allows DMA, interrupts, and base I/O addresses to be assigned by your system to avoid resource conflicts with other boards in your system. These boards are designed for high-performance data acquisition and control for applications in laboratory testing, production testing, and industrial process monitoring and control. You can use the TTL-compatible digital I/O lines for switching external devices such as transistors and solid-state relays, for reading the status of external digital logic, and for generating interrupts. You can use the counter/timers to synchronize events, generate pulses, and measure frequency and time. The Lab-PC-1200/AI, used in conjunction with the computer, is a versatile, cost-effective platform for laboratory test, measurement, and control.

This manual is intended for programming at the register level. Even if you are an experienced register-level programmer, consider using NI-DAQ or other National Instruments application software to program the Lab-PC-1200/AI. If NI-DAQ does not support your operating system, or you have other reasons to write your own register-level programs, continue reading this manual.

# **Board Configuration Overview**

This section is a reference to the Lab-PC-1200/AI configuration options. You should already have unpacked and installed your Lab-PC-1200/AI. Refer to your *Lab-PC-1200/AI User Manual* if you have not already performed these tasks.

#### **Analog Input Configuration**

The Lab-PC-1200/AI is completely software configurable, and at startup, defaults to the following configuration:

- Referenced single-ended input mode
- ±5 V analog input range

Table 1-1 lists the available analog I/O configurations for the Lab-PC-1200/AI and shows the default settings.

Parameter	Configuration
Analog Input Range	Bipolar—±5 V (default settings) Unipolar—0 to 10 V
Analog Input Mode	Referenced single-ended (RSE) (default setting) Nonreferenced single-ended (NRSE) Differential (DIFF)

Table 1-1. Analog Input Settings

The analog input circuitry is software configurable.

## Analog Output Configuration (Lab-PC-1200 Only)

At startup, the two channels of analog output of the Lab-PC-1200 default to the following configuration:

 $\pm 5$  V analog input range

Table 1-2 lists the available analog I/O configurations for the Lab-PC-1200 and shows the default settings.

Parameter	Configuration
Analog Output Range	Bipolar—±5 V (default settings) Unipolar—0 to 10 V

 Table 1-2.
 Analog Output Settings

The analog output circuitry is software configurable.

## **Digital I/O Configuration**

The Lab-PC-1200/AI uses the MSM82C55A PPI, which provides 24 digital lines in the form of three ports—A, B, and C. On power up, all three ports reset to mode 0 input. Appendix D, *OKI MSM82C55A Data Sheet*, has the 82C55A data sheets that you need to program the digital I/O.

#### **Counter Configuration**

You can use the MSM82C53 counter/timers for general-purpose applications, such as pulse and square wave generation, event counting, and pulse width, time-lapse, and frequency measurement. Appendix C, *OKI MSM82C53 Data Sheet*, has the 82C53 data sheet that you need to program the counters/timers.

# 2

# **Register Map and Descriptions**

This chapter describes in detail the address and function of each of the Lab-PC-1200/AI registers.

# **Register Map**

Table 2-1 shows the register map for the Lab-PC-1200/AI and lists the register name, address, type (read-only, write-only, or read-write), and size in bits.

Table 2-1 divides the Lab-PC-1200/AI registers into seven groups. The Configuration and Status Register Group controls the overall operation of the Lab-PC-1200/AI. The Analog Input Register Group reads output from the 12-bit successive-approximation ADC and can initiate conversions. The Analog Output Register Group accesses the two 12-bit DACs on the Lab-PC-1200 only. The two Counter/Timer Register Groups (A and B) access each of the two onboard 82C53 counter/timer integrated circuits. The Digital I/O Register Group consists of the four registers of the onboard 82C55A PPI integrated circuit that are used for digital I/O. The Interval Counter registers are used in single-channel interval-scanning acquisition.

The Lab-PC-1200/AI registers are 8-bit registers. To transfer 16-bit data, you must perform two consecutive memory readings or writings. For example, to read the 16-bit A/D conversion result, you must make two consecutive 8-bit readings of the FIFO. The first reading returns the low byte of the 16-bit data, and the second returns the high byte of the data.

Register Name	Address Offset (Hex)	Туре	Size
Configuration and Status Register Group			
Command Register 1	00	Write-only	8-bit
Command Register 2	01	Write-only	8-bit
Command Register 3	02	Write-only	8-bit
Command Register 4	0F	Write-only	8-bit
Command Register 5	1C	Write-only	8-bit
Command Register 6	0E	Write-only	8-bit
Status Register 1	00	Read-only	8-bit
Status Register 2	1D	Read-only	8-bit
Analog Input Register Group A/D FIFO Register	0A	Read-only	8-bit
A/D FIFO Clear Register	08	Write-only	8-bit
CONFIGMEM Register	03	Write-only	8-bit
DMA TC Interrupt Clear Register	0A	Write-only	8-bit
Analog Output Register Group			
DAC0 Low-Byte Register	04	Write-only	8-bit
DAC1 High-Byte Register	05	Write-only	8-bit
DAC1 Low-Byte Register	06	Write-only	8-bit
DAC1 High-Byte Register	07	Write-only	8-bit
82C53 Counter/Timer Register Group A			
Counter A0 Data Register	14	Read-Write	8-bit
Counter A1 Data Register	15	Read-Write	8-bit
Counter A2 Data Register	16	Read-Write	8-bit
Counter A Mode Register	17	Write-only	8-bit
Timer Interrupt Clear Register	0C	Write-only	8-bit
82C53 Counter/Timer Register Group B			
Counter B0 Data Register	18	Read-Write	8-bit
Counter B1 Data Register	19	Read-Write	8-bit
Counter B2 Data Register	1A	Read-Write	8-bit
Counter B Mode Register	1B	Write-only	8-bit

Table 2-1.	Lab-PC-1200/AI	Register Map
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Register Name	Address Offset (Hex)	Туре	Size
82C55A Digital I/O Register Group			
Port A Register	10	Read-Write	8-bit
Port B Register	11	Read-Write	8-bit
Port C Register	12	Read-Write	8-bit
Digital Control Register	13	Write-only	8-bit
Interval Counter Register Group			
Interval Counter Data Register	1E	Write-only	8-bit
Interval Counter Strobe Register	1F	Write-only	8-bit

Table 2-1. Lab-PC-1200/AI Register Map (Continued)

# **Register Description Overview**

The remainder of this chapter discusses each of the Lab-PC-1200/AI registers in the order shown in Table 2-1. Each register group is introduced, followed by a detailed bit description of each register on the Lab-PC-1200/AI. For a detailed bit description of each register concerning the 82C53 (A or B) chip or the 82C55A chip on the Lab-PC-1200/AI, refer to Appendix C, *OKI MSM82C53 Data Sheet*, or Appendix D, *OKI MSM82C55A Data Sheet*. The individual register description gives the address, type, word size, and bit map of the register, followed by a description of each bit.

The register bit map shows a diagram of the register with the most significant bit (MSB), bit 7 for an 8-bit register, shown on the left, and the least significant bit (LSB), bit 0, shown on the right. A rectangle labeled with the bit name inside its rectangle represents each bit. An asterisk (\*) after the bit name indicates that the bit is inverted (negative logic).

In a few of the registers, several bits are labeled with an X, indicating *don't care bits*. When you read a register, these bits may appear set or cleared but should be ignored because they have no significance. When you write to a register, these bit locations should always be written with a 0.

The bit map field for some write-only registers states *not applicable, no bits used*. Writing to these registers causes some event to occur on the Lab-PC-1200/AI, such as clearing the analog input circuitry. The data is ignored when writing to these registers; therefore, any bit pattern will suffice.

Is Note Always write a 0 to don't care bits.

# **Configuration and Status Register Group**

The eight registers of the Configuration and Status Register Group allow general control and monitoring of the Lab-PC-1200/AI A/D and D/A circuitry.

Command Register 1 and Command Register 2 contain bits that control the operation modes of the A/D and D/A circuitry. Command Register 3 enables or disables interrupt operations. Use Command Register 4 to select the analog input mode and to allow certain DAQ signals to be externally driven at the I/O connector. Use Command Register 5 for software calibration of the A/D circuitry. Use Command Register 6 to enable and disable interrupt operations and to configure the A/D and D/A circuitry.

Status Register 1 reports the status of a DAQ operation and the status of analog output during waveform generation. Status Register 2 reports the status of a DAQ operation and gives access to the output of the EEPROM.

Upon power up, all of the Command Registers are cleared.

Bit descriptions for the registers in the Configuration and Status Register Group are on the following pages.

## **Command Register 1**

Use Command Register 1 to select the input channel you want to scan, the gain for the analog input circuitry, the DAQ scanning mode, and the coding used for the output of the ADC.

Address:	00 (hex)
Туре:	Write-only
Word Size:	8-bit
Bit Map:	

7	6	5	4	3	2	1	0
SCANEN	GAIN2	GAIN1	GAIN0	TWOSCMP	MA2	MA1	MA0

Bit	Name	Description
7	SCANEN	Scan Enable—This bit enables or disables multiple-channel scanning during data acquisition. Set this bit to scan the analog channels as specified by MA<20> and SE*/DIFF (bit 3 of Command Register 4). Clear this bit to sample a single analog channel specified by MA<20> and SE*/DIFF during the entire DAQ operation.

6–4	GAIN<20>	Gain—These three bits select the gain setting as follows:
-----	----------	---

GAIN<20>	Selected Gain
000	1
001	Invalid
010	2
011	5
100	10
101	20
110	50
111	100

TWOSCMP

Two's Complement—This bit selects the coding format of the ADC output. Set this bit to sign-extend the 12-bit data

3

from the ADC to 16 bits (two's complement). Clear this bit to make bits 12 through 15 return 0 (straight binary).

2–0 MA<2..0> Multiplexer Address—These three bits select which of the eight input channels are scanned. The analog input multiplexers depend on these bits and also on SCANEN, SCANUP (bit 7 of Command Register 6), and SE\*/DIFF. Input channels are selected as follows:

	Selected Analog Input Channels					
	Single-Ended Mode	Different	ial Mode			
MA<20>	Scan Disabled/Enabled	Scan Disabled	Scan Enabled			
000	0	0	0			
001	1	0	2			
010	2	2	4			
011	3	2	6			
100	4	4	0			
101	5	4	2			
110	6	6	4			
111	7	6	6			

In single-ended mode (SE\*/DIFF cleared), if you set SCANEN and clear SCANUP, analog channels MA<2..0> through 0 are sampled sequentially. In single-ended mode (SE\*/DIFF cleared), if you set SCANEN and set SCANUP, analog channels 0 through MA<2..0> are sampled sequentially. If you clear SCANEN, a single analog channel specified by MA<2..0> is sampled during the entire DAQ operation.

In DIFF mode, the number of analog inputs reduces to four. The single-ended input channels 0 and 1 (pins 3 and 4) become differential input channel 0. The single-ended input channels 2 and 3 (pins 5 and 6) become differential input channel 2. There are no odd differential input channels.

# **Command Register 2**

Command Register 2 contains eight bits that control the Lab-PC-1200/AI analog input trigger modes, analog output update modes, and the coding scheme of the DACs.

Address:	01 (hex)
Туре:	Write-only
Word Size:	8-bit
Bit Map:	

7	6	5	4	3	2	1	0
LDAC1	LDAC0	2SDAC1	2SDAC0	TBSEL	SWTRIG	HWTRIG	PRETRIG

Bit	Name	Description
7	LDAC1	Load DAC1—This bit determines how DAC1 will be updated. If you set this bit, DAC1 updates its output at regular intervals as determined by counter A2 or the EXTUPDATE* signal at the I/O connector. If you clear this bit, the voltage output of DAC1 is immediately updated when data is loaded into the DAC1 High-Byte Register.
6	LDAC0	Load DAC0—This bit determines how DAC0 will be updated. If you set this bit, DAC0 updates its output at regular intervals as determined by counter A2 or the EXTUPDATE* signal at the I/O connector. If you clear this bit, the voltage output of DAC0 is immediately updated when data is loaded into the DAC0 High-Byte Register.
5	2SDAC1	Two's Complement DAC1—This bit selects the binary coding scheme used for the DAC1 data. If you set this bit, a two's complement binary coding scheme is used for interpreting the 12-bit data. Two's complement is used with bipolar output mode. If you clear this bit, a straight binary coding scheme is used. Straight binary is used with unipolar output mode.
4	2SDAC0	Two's Complement DAC0—This bit selects the binary coding scheme used for the DAC0 data. If you set this bit, a two's complement binary coding scheme is used for interpreting the 12-bit data. Two's complement is used with bipolar output mode. If you clear this bit, a straight

binary coding scheme is used. Straight binary is used with unipolar output mode.

- 3 TBSEL Time Base Select—This bit selects the clock source for counter A0, the sample interval timer. If you clear this bit, a 1 MHz clock drives counter A0, and the interval between samples is the value loaded into counter A0 multiplied by 1 µs. If you set this bit, the output of counter B0 is used as the clock source. The timebase for counter B0 is fixed at 2 MHz. The sample interval is the value loaded into counter A0 multiplied by the period of the output signal from counter B0.
- 2 SWTRIG Software Trigger—This bit is a software trigger for a DAQ operation. You can trigger a DAQ operation by setting this bit. The terminal count signal of counter A1 or a cleared SWTRIG terminates a DAQ process.
- 1 HWTRIG Hardware Trigger—This bit enables or disables the posttrigger mode using the EXTTRIG signal at the I/O connector. If you set this bit, you can use the EXTTRIG signal to trigger a DAQ operation in place of SWTRIG. A DAQ process is terminated by a terminal count signal of counter A1 or by writing to the A/D FIFO Clear Register. You must clear PRETRIG to use this mode.
- 0 PRETRIG Pretrigger—This bit enables or disables the pretrigger mode using the EXTTRIG signal at the I/O connector. If you set this bit, you can use the EXTRIG signal at the I/O connector to terminate a DAQ operation by using counter A1. Data acquisition is terminated by a terminal count on A1. You must clear the HWTRIG to use this mode.

# **Command Register 3**

The Command Register 3 contains four bits that enable and disable interrupts for a DAQ operation and for digital I/O.

Address:	02 (hex)
Type:	Write-only
Word Size:	8-bit
Bit Map:	

7	6	5	4	3	2	1	0
0	0	FIFOINTEN	ERRINTEN	CNTINTEN	TCINTEN	DIOINTEN	DMAEN

Bit	Name	Description
7,6	0	Always leave these bits cleared.
5	FIFOINTEN	FIFO Interrupt Enable—This bit enables and disables the generation of an interrupt when an A/D conversion result is available to be read from the A/D FIFO. If you set FIFOINTEN, an interrupt is generated whenever the DAVAIL bit becomes set in Status Register 1. Service this interrupt by reading the data from the FIFO.
4	ERRINTEN	Error Interrupt Enable—This bit enables and disables the generation of an interrupt when an A/D error condition is detected. If you set ERRINTEN, an interrupt is generated whenever the OVERFLOW or OVERRUN bit becomes set in Status Register 1. Service the interrupt by writing to the A/D FIFO Clear Register.
3	CNTINTEN	Counter Interrupt Enable—This bit enables the counter A2 output or the EXTUPDATE* signal to generate an interrupt. If you set CNTINTEN, an interrupt occurs whenever the CNTINT bit becomes set in Status Register 1. Clear this interrupt by writing to the Timer Interrupt Clear Register. This interrupt allows waveform generation on the analog output because the same signal that sets the interrupt also updates the DAC output if the corresponding LDAC bit in Command Register 2 is set.
2	TCINTEN	DMA Terminal Count Interrupt Enable—This bit enables generation of an interrupt when a DMA terminal count pulse is received. If TCINTEN is set, an interrupt request

is generated when the DMA controller transfer count register decrements from 0 to FFFF (hex). The interrupt is serviced by writing to the DMATC Interrupt Clear Register.

 1
 DIOINTEN
 DIO Interrupt Enable—This bit enables or disables generation of an interrupt when either Port A or Port B is ready to transfer data, and an interrupt request is set via PC3 or PC0 of 82C55A. See Appendix D, OKI

 MSM82C55A Data Sheet, for details. Clear this interrupt by clearing PC3 or PC0. If you clear DIOINTEN, the interrupts from PC3 or PC0 are disabled.

0 DMAEN DMA Enable—This bit enables or disables Analog Input DMA transfers. If DMAEN is set, a DMA request is generated whenever an A/D conversion result is available. If DMAEN is cleared, no DMA request will be generated.

See the *DAQ DMA Programming* section in Chapter 3, *Programming*, for more information.

# **Command Register 4**

Use this register to select the analog input mode, to enable interval scanning, and to allow the I/O connector pins to externally drive certain DAQ signals.

Address:	0F (hex)
Туре:	Write-only
Word Size:	8-bit
Bit Map:	

7	6	5	4	3	2	1	0
0	0	0	ECLKRCV	SE*/DIFF	ECLKDRV	EOIRCV	INTSCAN

Bi	t	Name	Description				
7—	5	0	Always leave the	se bits cleared.			
4			External Clock Receive—This bit disables or enables the external signal EXTCONV*. If you set this bit, transitions on EXTCONV* will not effect data acquisition. If you clear this bit, a falling edge on EXTCONV* initiates an A/D conversion if ECLKDRV is cleared.				
3			A/D conversion if ECLKDRV is cleared. Single-Ended/Differential—This bit, along with bit 0 of Command Register 6 (RSE*/NRSE), selects one of three analog input modes of the Lab-PC-1200/AI. You can select the single-ended mode by clearing this bit and you can select the differential mode by setting this bit. Refer to the <i>Lab-PC-1200/AI User Manual</i> for an explanation of the different modes. The following table illustrates how to choose the various input modes by using SE*/DIFF and RSE*/NRSE.				
			RSE*/NRSE	SE*/DIFF	Input Mode		
			0	0	RSE (reset condition)		
			1	0	NRSE		

DIFF

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2	ECLKDRV	External Clock Drive—When you clear this bit (default power up), you can drive the EXTCONV* pin at the I/O connector to cause conversions (if ECLKRCV is also cleared). When you set this bit, you enable internally timed conversions and the conversion pulses are driven onto the EXTCONV* pin for synchronizing channels on SCXI modules (if used with SCXI).
1	EOIRCV	External Output Interval Clock Receive—This bit selects the clock source for interval scanning. If you clear this bit, counter B1 drives the interval scanning circuitry. This will also configure OUTB1 on the I/O connector as an output signal. If you set this bit, OUTB1 on the I/O connector is selected as an input signal and allows you to externally drive the interval scanning circuitry.
0	INTSCAN	Interval Scan—This bit selects the DAQ mode. When you set this bit, the Lab-PC-1200/AI performs interval data acquisition. If you clear this bit, freerun or controlled data acquisition occurs. For an explanation of the different modes, refer to Chapter 4, <i>Theory of Operation</i> , in the <i>Lab-PC-1200/AI User Manual</i> . Also, this bit selects the clock source for counter B1 used in interval scanning. If interval scanning is disabled (INTSCAN = 0), then counter B1 is available for user applications. You can then drive CLKB1 externally at the I/O connector. If interval scanning is enabled (INTSCAN = 1), the clock source of counter A0 also drives CLKB1. This source can further be selected by using the TBSEL bit in Command Register 2.

# **Command Register 5**

Use Command Register 5 for software calibration of the A/D and D/A circuitry, for interaction with the EEPROM, and for enabling dither.

Address:	1C (hex)
Туре:	Write-only
Word Size:	8-bit
Bit Map:	

7	6	5	4	3	2	1	0	
EEPROMCS	SDATA	SCLK	CALDACLD	DITHEREN	WRTPRT*	0	0	

Bit	Name	Description
7	EEPROMCS	EEPROM Chip Select—This bit enables and disables the EEPROM. You can enable the EEPROM for both read and write operations by setting this bit. You can disable the EEPROM by clearing this bit. Notice that this bit is inverted on the Lab-PC-1200/AI to make EEPROMCS* as explained in Chapter 5, <i>Calibration</i> , in your <i>Lab-PC-1200/AI User Manual</i> .
6	SDATA	Serial Data—This bit is a serial data input for both the calibration DACs and the EEPROM.
5	SCLK	Serial Clock—This bit is a serial clock for both the calibration DACs and the EEPROM. A low-to-high transition of this bit clocks data into the EEPROM (during a write operation) and the calibration DAC. A high-to-low transition of the bit clocks data out of the EEPROM (during a read operation).
4	CALDACLD	Calibration DAC Load—This bit updates the calibration DACs. After you load the calibration DAC address and data, set CALDACLD high to update the selected CALDAC output signal.
3	DITHEREN	Dither Enable—This bit enables or disables the dither circuitry. When you set this bit, 0.5 LSB of white Gaussian noise is added to the selected analog input signal. By enabling dither and using averaging, you can achieve greater input resolution.

2	WRTPRT*	Write Protect—This bit controls the write protect input signal for the EEPROM. When you set this bit, you enable normal write operations. When you clear this bit, you disable write operations.
1, 0	0	Always leave these bits cleared.

Lab-PC-1200/AI RLPM

# **Command Register 6**

Use Command Register 6 to enable A/D interrupts and to configure the A/D and D/A circuitry.

Address:	0E (hex)
Туре:	Write-only
Word Size:	8-bit
54.35	

Bit Map:

7	6	5	4	3	2	1	0		
SCANUP	DQINTEN	HFINTEN	0	DAC1UNI/BI*	DAC0UNI/BI*	ADCUNI/BI*	RSE*/NRSE		
Bi	it	Name	Dese	cription					
7		SCANUP	Scan Up—This bit selects the order in which the analog input channels are scanned. Clear this bit to select down counting (highest numbered channel scanned first). Set this bit to select up counting (channel 0 scanned first).						
6		DQINTEN	DAQ Interrupt Enable—This bit enables and disables the end of a DAQ operation interrupt. Set this bit to generate an interrupt whenever the OUTA1 bit in Status Register 2 becomes set. Service this interrupt by resetting counter A1. Clear this bit to disable interrupt generation.						
5		HFINTEN	Half-Full Interrupt Enable—This bit enables and disab the FIFO half-full interrupt. Set this bit to generate an interrupt whenever the FIFOHF* bit in Status Register becomes cleared. Service this interrupt by reading data from the FIFO. Clear this bit to disable interrupt generation.						
4		0	Always leave this bit cleared.						
3		DAC1UNI/BI*	outp a un to co	ut range for D ipolar (0 to +1	ipolar—This AC1. Set this 0 V) output v 1 for a bipola	bit to configu oltage range.	re DAC1 for Clear this bit		

2	DAC0UNI/BI*	DAC0 Unipolar/Bipolar—This bit sets the analog voltage output range for DAC0. Set this bit to configure DAC0 for a unipolar (0 to $+10$ V) output voltage range. Clear this bit to configure DAC0 for a bipolar ( $-5$ to $+5$ V) output voltage range.
1	ADCUNI/BI*	ADC Unipolar/Bipolar—This bit sets the analog voltage input range for data acquisition. Set this bit to select a unipolar (0 to $+10$ V) voltage input range. Clear this bit to select a bipolar ( $-5$ to $+5$ V) voltage input range.
0	RSE*/NRSE	Referenced Single-Ended/Nonreferenced Single-Ended—This bit, and bit 3 of Command Register 4 (SE*/DIFF), selects one of three input modes of the Lab-PC-1200/AI. The status of RSE*/NRSE is only important with the single-ended analog-input modes. Set this bit to select the nonreferenced single-ended mode. Clear this bit to select the referenced single-ended mode. For an explanation of the three input modes, refer to the <i>Lab-PC-1200/AI User Manual</i> .

#### Status Register 1

Status Register 1 indicates the status of the current DAQ operation and the status of analog output during waveform generation. These bits indicate if a DAQ operation is in progress or if data is available, whether any errors have been found, and the analog output interrupt status.

Address:	00 (hex)
Type:	Read-only
Word Size:	8-bit
Bit Map:	

5 7 6 4 3 2 1 0 Х EXTGATA0 GATA0 DMATC CNTINT OVERFLOW **OVERRUN** DAVAIL

Bit	Name	Description
7	Х	Don't care bits.
6	EXTGATA0	External Gate A0—This bit indicates the status of the external trigger signal (EXTTRIG) during a DAQ operation in posttrigger mode. If this bit is set, EXTTRIG has triggered a DAQ operation. Clear this bit by writing to the A/D FIFO Clear Register.
5	GATA0	Gate A0—This bit indicates the status of the GATEA0 input for counter A0. Use this bit as a busy indicator for DAQ operations because conversions are enabled as long as GATEA0 is high and counter A0 is programmed appropriately. A DAQ operation is terminated when GATA0 is cleared.
4	DMATC	DMA Terminal Count—This bit reflects the status of the DMA terminal count. If this bit is set and the TCINTEN bit in Command Register 3 is set, the current interrupt was generated by a DMA terminal count pulse. This bit is cleared by writing to the DMATC Interrupt Clear Register.
3	CNTINT	Counter Interrupt—This bit reflects the status of the interrupt caused by counter A2 output or the EXTUPDATE* signal. A low-to-high transition on counter A2 output or on EXTUPDATE* sets this bit. You can generate an interrupt if you set CNTINTEN in Command Register 3. Clear this bit by writing to the Timer Interrupt Clear Register.

2	OVERFLOW	Overflow—This bit indicates if an overflow error has occurred. If this bit is cleared, no error was encountered. If this bit is set, the A/D FIFO has overflowed because the DAQ servicing operation could not keep up with the sampling rate.
1	OVERRUN	Overrun—This bit indicates if an overrun error has occurred. If this bit is cleared, no error occurred. This bit is set if a convert command is issued to the ADC while the last conversion is still in progress.
0	DAVAIL	Data Available—This bit indicates if conversion output is available. If this bit is set, the ADC is finished with the last conversion and the result can be read from the FIFO. This bit is cleared if the FIFO is empty.

OUTA1

PROMOUT

## Status Register 2

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Status Register 2 reports the status of a DAQ operation and the FIFO half-full output and gives access to the output of the EEPROM.

Ado	dress:	1D (hex)					
Тур	be:	Read-only					
Wo	rd Size:	8-bit					
Bit Map:							
7	6	5	4	3	2	1	0

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FIFOHF\*

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Bit	Name	Description
7–3	Х	Don't care bits.
2	FIFOHF*	FIFO Half-Full—This bit indicates the status of the FIFO half-full output. If this bit is low, the FIFO is at least half full. If this bit is high, the FIFO is less than half full.
1	OUTA1	Output Counter A1—This bit indicates the status of the output signal of counter A1. If the user sets the counter A1 mode for a terminal count, OUTA1 low indicates counter A1 has not started counting or that counting is in progress. OUTA1 high indicates that counter A1 has finished counting.
0	PROMOUT	EEPROM Output—This bit allows access to the serial output pin of the EEPROM. During calibration procedures, the software reads the calibration data from the EEPROM through PROMOUT and writes the data to the calibration DACs.

## **Analog Input Register Group**

The three registers making up the Analog Input Register Group control the analog input circuitry and the FIFO. Reading the FIFO Register returns stored A/D conversion results. Writing to the Start Convert Register initiates a single A/D conversion. Writing to the A/D FIFO Clear Register clears the analog input circuitry.

Bit descriptions for the registers of the Analog Input Register Group are on the following pages.

# A/D FIFO Register

The 12-bit A/D conversion results are sign-extended to 16-bit data in either two's complement or straight binary format and are stored into a 4 Kword-deep A/D FIFO buffer. Two consecutive 8-bit readings of the A/D FIFO Register return an A/D conversion value stored in the A/D FIFO. The first reading returns the low byte of the 16-bit value, and the second reading returns the high byte. The value read is removed from the A/D FIFO, thereby freeing space for another A/D conversion value to be stored.

The A/D FIFO is empty when all values it contains are read. The DAVAIL bit in Status Register 1 should be read before the A/D FIFO Register is read. If the A/D FIFO contains one or more A/D conversion values, the DAVAIL bit is set and the A/D FIFO Register can be read to retrieve a value. If the DAVAIL bit is cleared, the A/D FIFO is empty. Therefore, reading the A/D FIFO Register returns meaningless information.

The values returned by reading the A/D FIFO Register are available in two different binary formats: straight binary or two's complement binary. The binary format used is selected by the TWOSCMP bit in Command Register 1. The bit pattern returned for either format is as follows.

Address:	0A (hex)
Туре:	Read-only
Word Size:	8-bit
Bit Map:	Straight binary mode
High Byte	

7	6	5	4	3	2	1	0
0	0	0	0	D11	D10	D9	D8

Low Byte

				3				
D7	D6	D5	D4	D3	D2	D1	D0	

Bit	Name	Description
High Byte		
7–4	0	These bits always return 0 in straight binary mode.
3 -0	D<118>	Data—These bits contain the high byte of the straight binary result of a 12-bit A/D conversion. Values made up of D<110> range from 0 to +4,095 decimal (0000 to

			0FFF hex). Straight binary mode is useful for unipolar analog input readings because all values read reflect a positive polarity input signal.					
Lov	w Byte							
7–0			binary res	Data—These bits contain the low byte of the straight binary result of a 12-bit A/D conversion. The first of the two consecutive readings of the A/D FIFO Register returns this byte.				
	Map:	Two's con	plement bin	ary mode				
Hig	h Byte							
7	6	5	4	3	2	1	0	
D15	D14	D13	D12	D11	D10	D9	D8	
Lov	w Byte							
7	6	5	4	3	2	1	0	
D7	D6	D5	D4	D3	D2	D1	D0	
Bit	Na	ime	Descripti	on				
Hig	h Byte							
7–0	7–0 D<158>		sign-exter conversio from -2,0 compleme readings b	Data—These data bits contain the high byte of the 16-bit, sign-extended two's complement result of a 12-bit A/D conversion. Values made up of D<150>, therefore, range from $-2,048$ to $+2,047$ decimal (F800 to 07FF hex). Two's complement mode is useful for bipolar analog input readings because the values read reflect the polarity of the input signal.				
Lov	w Byte							
7–0 D<70>			sign-exter conversio	ese data bits ided two's c n. The first c Register ret	omplement in the two co	result of a 12 onsecutive re	2-bit A/D	

### A/D FIFO Clear Register

Write to this register to reset the ADC FIFO. This operation clears the FIFO, clears the DAVAIL bit, and sets the FIFOHF\* bit. All error bits in Status Register 1 are cleared.

Address:	08 (hex)
Туре:	Write-only
Word Size:	8-bit
Bit Map:	Not applicable, no bits used

### **Start Convert Register**

Write to the Start Convert Register to initiate a single A/D conversion.

Address:	03 (hex)
Type:	Write-only
Word Size:	8-bit
Bit Map:	Not applicable, no bits used

### **DMATC Interrupt Clear Register**

Writing to the DMA Terminal Count (DMATC) Clear Register clears the interrupt request asserted when a DMA terminal count pulse is detected.

Address:	0A (hex)
Туре:	Write-only
Word Size:	8-bit
Bit Map:	Not applicable, no bits used

### Analog Output Register Group (Lab-PC-1200 Only)

Use the four registers of the Analog Output Register Group to load the two 12-bit DACs. DAC0 controls analog output channel 0. DAC1 controls analog output channel 1. Write to these DACs individually.

Bit descriptions of the registers making up the Analog Output Register Group are on the following pages.

## Image: NoteDACx represents DAC0 and DAC1 registers. LDACx represents LDAC0 and<br/>LDAC1 bits.

## DACO Low-Byte, DACO High-Byte, DAC1 Low-Byte, and DAC1 High-Byte Registers

Write to DAC0 Low-Byte and then to DAC0 High-Byte to load DAC0. Write to DAC1 Low-Byte and then to DAC1 High-Byte to load DAC1. If you clear the LDACx bit in Command Register 2, then the corresponding analog output channel is updated immediately after you write to the DACx High-Byte register. If you set the LDACx bit, then the corresponding analog output channel is updated when an active low pulse occurs on the output of counter A2 or on the EXTUPDATE\* line on the I/O connector.

Ado	dress:	04 (hex) 05 (hex) 06 (hex) 07 (hex)	DAC0 Low Byte DAC0 High Byte DAC1 Low Byte DAC1 High Byte				
Тур	be:	Write-only	v (all)				
Wo	rd Size:	8-bit (all)					
Bit	Мар:						
DA	CxH						
7	6	5	4	3	2	1	0
D15	D14 D13		D12	D11	D10	D9	D8
DA	CxL						
7	6	5	4	3	2	1	0
D7	D6	D5	D4	D3	D2	D1	D0
<b>Bit</b> DA	Na CxH	Name		on			
7–4	D<	<1512>	Data—Zero in straight binary mode, sign extension in two's complement mode.			nsion in	
3–0	3–0 D<118>		Data—Th DAC high	ese four bits 1 byte.	are loaded	into the spec	cified

DACxL

7–0	D<70>	Data—These eight bits are loaded into the specified
		DAC low byte should be loaded first, followed by the
		corresponding high byte.

### 82C53 Counter/Timer Register Groups A and B

The nine registers of the two counter/timer register groups access the two onboard 82C53 counter/timers. Each 82C53 has three counters. For convenience, the two counter/timer groups and their respective 82C53 integrated circuits have been designated A and B. The three counters of group A control onboard DAQ timing and waveform generation. You can use the three counters of group B for general-purpose timing functions.

Each 82C53 has three independent 16-bit counters and one 8-bit Mode Register. The Mode Register sets the mode of operation for each of the three counters.

Writing to the Timer Interrupt Clear Register clears the interrupt request asserted when a low pulse is detected on the output of counter A2 or on the EXTUPDATE\* line.

Bit descriptions for the registers in the Counter/Timer Register Groups are in the following pages.

### **Counter AO Data Register**

Use the Counter A0 Data Register to write data and read back the contents of 82C53(A) counter 0. Counter A0 is the sample interval counter for a DAQ operation.

Tyj Wo	dress: pe: rd Size: Map:	14 (hex) Read-and- 8-bit	write				
7	6 6	5	4	3	2	1	0
D7	D6	D5	D4	D3	D2	D1	D0
Bit	t Name		Descripti	ion			
7–0	) D<70>		Data—8-	bit counter A	0 contents.		

### **Counter A1 Data Register**

Use the Counter A1 Data Register to write data and read back the contents of 82C53(A) counter 1. Counter A1 is the sample counter for a DAQ operation.

Address:	15 (hex)
Type:	Read-and-write
Word Size:	8-bit
Bit Map:	

	7	6	5	4	3	2	1	0
ſ	D7	D6	D5	D4	D3	D2	D1	D0
	Bit	Na	me	Descripti	on			

7–0 D<7..0> Data—8-bit counter A1 contents.

### **Counter A2 Data Register**

Use the Counter A2 Data Register to write data and read back the contents of 82C53(A) counter A2. Counter A2 is the DAC update timer for waveform generation.

Address:16 (hex)Type:Read-and-Word Size:8-bitBit Map:		write					
7	6	5	4	3	2	1	0
D7	D6	D5	D4	D3	D2	D1	D0
Bit	Name		Descripti				
7–0 D<70>		Data—8-I	Data—8-bit counter A2 contents.				

### **Counter A Mode Register**

The Counter A Mode Register determines the operation mode for each of the three counters on the 82C53(A) chip. The Counter A Mode Register selects the counter involved, its read/write mode, its operation mode (that is, any of the 82C53 six operation modes), and the counting mode (binary or BCD counting).

The Counter A Mode Register is an 8-bit register. Bit and programming descriptions for each of these bits are in Appendix C, *OKI MSM82C53 Data Sheet*.

Address:	17 (hex)
Туре:	Write-only
Word Size:	8-bit
Bit Map:	

7	6	5	4	3	2	1	0
SC1	SC0	RL1	RL0	M2	M1	M0	BCD

### **Timer Interrupt Clear Register**

Write to the Timer Interrupt Clear Register to clear the interrupt request asserted when a low pulse is detected on the counter A2 output or on EXTUPDATE\* line.

Address:	0C (hex)
Туре:	Write-only
Word Size:	8-bit
Bit Map:	Not applicable, no bits used.

### **Counter BO Data Register**

Use the Counter B0 Data Register to write data and read back the contents of 82C53(B) counter 0. Counter B0 either supplies the time base clock for counter A0 or is reserved for external usage.

Address:	18 (hex)
Туре:	Read-and-write
Word Size:	8-bit
Bit Map:	

7	6	5	4	3	2	1	0
D7	D6	D5	D4	D3	D2	D1	D0
DI	20	20				DI	

Bit	Name	Description
7–0	D<70>	Data—8-bit counter B0 contents.

### **Counter B1 Data Register**

Use the Counter B1 Data Register to write data and read back the contents of 82C53(B) counter 1. Counter B1 is either the interval timer for a DAQ operation or is reserved for external usage.

Ado	dress:	19 (hex)					
Тур	be:	e: Read-and-write					
Word Size: 8-bit							
Bit	Мар:						
7	6	5	4	3	2	1	0
D7	D6	D5	D4	D3	D2	D1	D0
Bit	Na	ime	Description				
7–0	D<	<70>	Data—8-bit counter B1 contents.				

### **Counter B2 Data Register**

Use the Counter B2 Data Register to write data and read back the contents of 82C53(B) counter 2. Counter B2 is reserved for external usage.

Address:	1A (hex)
Туре:	Read-and-write
Word Size:	8-bit
Bit Map:	

7	6	5	4	3	2	1	0
D7	D6	D6	D4	D3	D2	D1	D0

Bit	Name	Description
7–0	D<70>	Data—8-bit counter B2 contents.

### **Counter B Mode Register**

The Counter B Mode Register determines the operation mode for each of the three counters on the 82C53(B) chip. The Counter B Mode Register selects the counter involved, its read/write mode, its operation mode (that is, any of the 82C53 six operation modes), and the counting mode (binary or BCD counting).

The Counter Mode Register is an 8-bit register. Bit descriptions for each of these bits are in Appendix C, *OKI MSM82C53 Data Sheet*.

Address:1B (hex)Type:Write-onlyWord Size:8-bitBit Map:

7	6	5	4	3	2	1	0
SC1	SC0	RL1	RL0	M2	M1	M0	BCD

### 82C55A Digital I/O Register Group

The four registers of the Digital I/O register group access the 82C55A peripheral interface. The 82C55A is a general-purpose peripheral interface containing 24 programmable I/O pins. These pins represent the three 8-bit I/O ports (A, B, and C) of the 82C55A and the 24 digital I/O lines on the I/O connector. You can program these ports as two groups of 12 signals or as three individual 8-bit ports. You can also configure them as either input or output pins. Use the Digital Control Register to configure the three ports. For further information on the 82C55A, refer to Appendix D, *OKI MSM82C55A Data Sheet*.

Bit descriptions for the registers in the Digital I/O Register Group are on the following pages.

### Port A Register

Use the Port A Register to write and to read the eight digital I/O lines constituting port A on the I/O connector (PA<0..7>). See Appendix D, *OKI MSM82C55A Data Sheet*, for information on how to configure port A for input or output.

Ade	dress:	10 (hex)					
Тур	pe:	Read-and-write					
Word Size: 8-bit							
Bit	Map:						
7	6	5	4	3	2	1	0
D7	D6	D5	D4	D3	D2	D1	D0
Bit	Na	ime	Description				
7–0	) D<	<70>	Data—8-bit port A data.				

### Port B Register

Use the Port B Register to write and to read the eight digital I/O lines constituting port B, that is, PB<0..7>. See Appendix D, *OKI MSM82C55A Data Sheet*, for information on how to configure port B for input or output.

Address:	11 (hex)
Type:	Read-and-write
Word Size:	8-bit
D!4 M	

#### Bit Map:

7	6	5	4	3	2	1	0
D7	D6	D5	D4	D3	D2	D1	D0

data.

Bit	Name	Description
7–0	D<70>	Data—8-bit port B

### **Port C Register**

You can use port C as an 8-bit I/O port like port A and port B if neither port A nor port B is used in handshaking mode. If either port A or port B is configured for mode 1 or mode 2, some of the bits in port C are used for handshaking signals. See Appendix D, *OKI MSM82C55A Data Sheet*, for a description of the individual bits in the Port C Register.

<b>Address:</b> 12 (		12 (hex)					
Type: Read-and-		write					
Word Size: 8-bit							
Bit	Map:						
7	6	5	4	3	2	1	0
D7	D6	D5	D4	D3	D2	D1	D0
Bit	Na	ime	Descripti	on			
7–0 D<70>			Data—8-bit port C data.				

### **Digital Control Register**

You can use the Digital Control Register to configure port A, port B, and port C as inputs or outputs, and you can select simple mode (basic I/O) or handshaking mode (strobed I/O) for transfers. See Appendix D, *OKI MSM82C55A Data Sheet*, for a description of the individual bits in the Digital Control Register.

Address:	13 (hex)
Туре:	Write-only
Word Size:	8-bit
Bit Map:	

	7	6	5	4	3	2	1	0
ſ	CW7	CW6	CW5	CW4	CW3	CW2	CW1	CW0
	Bit	Na	ime	Descripti	on			
	7–0	CV	N<70>	Control W	Vord—8-bit	control word	1.	

### **Interval Counter Register Group**

Use the 8-bit Interval Counter only in the single-channel interval scanning mode (SCANEN = 0 and INTSCAN = 1). Refer to the *DAQ Operations* section in Chapter 4, *Theory of Operation*, of your *Lab-PC-1200/AI User Manual* for an explanation of interval scanning mode. The Interval Counter consists of two 8-bit registers—the Interval Counter Data Register and the Interval Counter Strobe Register. Load the Interval Counter Data Register with the count. Writing to the Interval Counter Strobe Register loads this count into the Interval Counter. The Interval Counter decrements with each conversion. When the count reaches 0, the Interval Counter autoinitializes, restoring the original count value.

Bit descriptions for the registers in the Interval Counter Register Group are on the following pages.

### **Interval Counter Data Register**

Load the Interval Counter Data Register with the desired number of samples to be acquired within a scan interval of a single channel DAQ operation.

		1E (hex) Write-only	7	-			
Bit Map:		8-011					
7	6	5	4	3	2	1	0
D7	D6	D5	D4	D3	D2	D1	D0
Bit Name		Descripti	on				
7–0 D<70>		Data—Interval counter count.					

### **Interval Counter Strobe Register**

Writing to the Interval Counter Strobe Register strobes the contents of the Interval Counter Data Register into the Interval Counter. This action arms the Interval Counter, which decrements with each conversion pulse.

Address:	1F (hex)
Type:	Write-only
Word Size:	8-bit
Bit Map:	

7	6	5	4	3	2	1	0
0	0	0	0	0	0	0	1

Bit	Name	Description
7–0	0	Each of these bits must be 0 for proper operation of the Lab-PC-1200/AI.
0	1	This bit must be 1 for proper operation of the Lab-PC-1200/AI.

## Programming

This chapter contains programming instructions for operating the Lab-PC-1200/AI circuitry, and examples of the programming steps necessary to execute an operation. If you are not using NI-DAQ, you must first initialize your board. The initialization steps are unique for PC and Macintosh users, so refer to the section pertaining to your platform.

Programming the Lab-PC-1200/AI involves writing to and reading from registers on the board. You will find a listing of these registers in Chapter 2, *Register Map and Descriptions*, of this manual.

## **Register Programming Considerations**

The Lab-PC-1200/AI supports 8-bit I/O transfers; thus, all the read-and-write operations are 8-bit operations. You must do 16-bit transfers in two consecutive 8-bit operations.

Several write-only registers on the Lab-PC-1200/AI contain bits that control several independent pieces of the onboard circuitry. In the set or clear instructions, specific register bits should be set or cleared without changing the current state of the remaining bits in the register. However, writing to these registers affects all register bits simultaneously. Because you cannot read these registers to determine their current status, you should maintain a software copy of the write-only registers. To change the state of a single bit without disturbing the remaining bits, set or clear the bit in the software copy and then write the modified software copy to the register.

## **Programming Examples**

The programming examples in this section demonstrate the programming steps needed to perform several different operations. The instructions are language independent; that is, they tell you to read or write a given register or to detect if a given bit is set or cleared, without presenting the actual code. The information given is not intended to be used without proper modification in a practical solution. Before you can implement any of the examples into a real application, you must know the base memory address for your board. To process any interrupts, you must write and install an applicable interrupt service routine.

In this chapter all numbers preceded by 0x are hexadecimal.

Common terms that you will see used in the programming examples are listed below:

Write (address, data)	Generic function call for a I/O space Write of data to address
Read (address)	Generic function call for a I/O space Read from address

### Lab-PC-1200/AI Companion Disk

The companion disk provides code to execute the ISA Plug and Play algorithm and assign resources to the card. This process is normally performed by your computer's operating system, such as Windows 95, but for programming environments that do not execute the Plug and Play algorithm, such as MS-DOS, the board will not be operable. The code provided can be used to execute the Plug and Play algorithm for environments such as MS-DOS. The code was compiled with Microsoft Visual C++, version 1.5, as an MS-DOS application.

### Assigning Lab-PC-1200/AI Resources

The companion disk code allows you to assign any of the allowable resources for the base I/O address, interrupt channel, and DMA channel. Table 3-1 lists the acceptable resources for the Lab-PC-1200/AI board. The function, PNPProgramSrom(), located in the file, pnp\_util.c, assigns the resources to the board. This function, as written, assigns all boards interrupt channel 5, DMA channel 3, and base I/O address  $0x220 + (0x20 \cdot j)$ , where *j* is the board number. Thus, the first Lab-PC-1200/AI board will be assigned base I/O address 0x220, the next 0x240, and so on. This function should be changed accordingly for your application.

Base I/O Address	IRQ Channels	DMA Channels
0x100-0x3e0 (in 0x20 increments)	3, 4, 5, 6, 7, 9	1, 3

## Initializing the Lab-PC-1200/AI Circuitry

Initializing the Lab-PC-1200/AI circuitry involves two steps—reset and calibration. The Lab-PC-1200/AI circuitry is reset on power up. Calibration of the analog input and output circuitry involves reading the factory-defined calibration constants from the onboard EEPROM and loading them into the appropriate calibration DACs. For information on using user-defined calibration constants, refer to Chapter 4, *Calibration*.

When reset, the Lab-PC-1200/AI is left in the following state:

- 1. All of the command registers are cleared.
- 2. All of the interrupts are disabled and cleared.
- 3. The 82C55A digital I/O is in Mode 0 input.
- 4. The analog output DACs are reset to 0.0 V (Lab-PC-1200 only).

Calibrating the Lab-PC-1200/AI involves two steps—reading eight factory-defined calibration constants from the EEPROM and writing each value to the appropriate CALDAC. Choose the desired factory-defined calibration constants as explained in Chapter 5, *Calibration*, of your *Lab-PC-1200/AI User Manual*.

Use the following sequence of steps to read a single byte from the EEPROM.

- 1. Set the EEPROMCS and WRTPRT\* bits in Command Register 5.
- 2. Serially write the READ instruction, 0x03, to the EEPROM to start a read operation.
- 3. Serially write the 8-bit address.
- 4. Serially read one byte of data from the EEPROM.
- 5. Clear the EEPROMCS and WRTPRT\* bits in Command Register 5 to end the read operation.

Repeat the following two steps eight times to serially write one byte to the EEPROM.

- 1. Write a single bit of the 8-bit value, MSB first, by setting or clearing the SDATA bit in Command Register 5. The SCLK bit in Command Register 5 should be cleared during this step.
- 2. Set the SCLK bit in Command Register 5.

Repeat the following three steps eight times to serially read one byte from the EEPROM.

- 1. Set the SCLK bit in Command Register 5.
- 2. Clear the SCLK bit in Command Register 5.
- 3. Read a single bit of the 8-bit value, MSB first, by reading the PROMOUT bit of Status Register 2.

After reading a single calibration constant from the EEPROM, you must write this value to the appropriate CALDAC. Use the following steps to write a single byte to a CALDAC.

- 1. Write the 4-bit address of the desired CALDAC (address 0x3-0xA), LSB first. CALDACLD should be cleared during this step.
- 2. Write the 8-bit calibration constant, MSB first.
- 3. Set the CALDACLD bit in Command Register 5.
- 4. Clear the CALDACLD bit in Command Register 5.

Repeat the following steps the appropriate number of times to serially write required bits to the CALDAC.

- 1. Write a single bit of the value by setting or clearing the SDATA bit in Command Register 5. SCLK should be cleared during this step.
- 2. Set the SCLK bit in Command Register 5.

Repeat the calibration operation for each CALDAC to fully calibrate the Lab-PC-1200/AI.

# Programming the Analog Input Circuitry for Single A/D Conversions

This section explains how to clear the analog input circuitry, how to configure the analog input circuitry, and how to perform single A/D conversions.

### **Clearing the Analog Input Circuitry**

Before clearing the analog input circuitry, perform the following steps:

- 1. Clear SWTRIG in Command Register 2.
- 2. Ensure that EXTCONV\* from the I/O connector does not cause any conversions.

3. Ensure that EXTTRIG from the I/O connector does not cause any conversions.

Clearing the analog input circuitry does not stop a DAQ operation that was started by the SWTRIG bit in Command Register 2. Therefore, you should clear this bit before clearing the analog input circuitry.

While clearing the analog input circuitry, do not externally drive the EXTCONV\* pin on the I/O connector or, if you do, drive it high. Another option is to set the ECLKRCV bit in Command Register 4. This disables any transitions on the EXTCONV\* pin and no unwanted conversions will occur while you are clearing the analog input circuitry.

While clearing the analog input circuitry, do not externally drive the EXTTRIG pin on the I/O connector or, if you do, drive it high. Another option is to make sure the HWTRIG bit in Command Register 2 is cleared. This disables any transitions on the EXTTRIG pin and no unwanted conversions will occur while you are clearing the analog input circuitry.

The analog input circuitry can be cleared by writing to the A/D FIFO Clear Register, which leaves the analog input circuitry in the following state:

- 1. Analog input status bits OVERRUN, OVERFLOW, and DAVAIL are cleared and FIFOHF\* is set.
- 2. Pending interrupt requests from the analog input circuitry are cleared.

The command registers are not cleared, so you do not necessarily have to reconfigure the Lab-PC-1200/AI before initiating another DAQ sequence.

### **Configuring the Analog Input Circuitry**

Configure the analog input circuitry after initializing the Lab-PC-1200/AI and any time the characteristics of the analog input signals change. Program the appropriate register bits as follows (not necessarily in this order):

- Select the appropriate input mode (DIFF, NRSE, or RSE).
- Select the input polarity (bipolar or unipolar) and coding of the resulting conversions (straight binary or two's complement).
- Select the analog input channels to be scanned and gain.

You determine the input mode by identifying the types of signal sources that you are using. For more information about determining the input mode, consult the *Lab-PC-1200/AI User Manual*. Select the input mode by setting

or clearing the RSE\*/NRSE and SE\*/DIFF bits in Command Register 6 and Command Register 4, respectively.

You determine the input polarity according to the voltage range of the analog input signal. For more information about determining the input polarity, consult the *Lab-PC-1200/AI User Manual*. The conversion coding should be selected according to the input polarity. Select the input polarity by setting or clearing the ADCUNI/BI\* bit in Command Register 6. Select the respective coding by setting or clearing the TWOSCMP bit in Command Register 1.

If you will be performing single-channel data acquisition, select the desired analog channel and gain by setting or clearing the GAIN<2..0> and MA<2..0> bits in Command Register 1. You must also clear SCANEN in Command Register 1. If you will be doing multiple-channel data acquisition, set MA<2..0> to specify the highest numbered channel in the scan sequence. Set or clear the SCANUP bit in Command Register 6 for the desired scanning order. For example, if you set MA<2..0> to 011 (binary) and clear the SCANUP bit, the following scan sequence is used:

channel 3, channel 2, channel 1, channel 0, channel 3, channel 2, and so on.

If you set MA<2..0> to 011 (binary) and set the SCANUP bit, the following scan sequence is used:

channel 0, channel 1, channel 2, channel 3, channel 0, channel 1, and so on.

Select the analog input channel and gain for multiple-channel data acquisition in the following order.

- 1. Set or clear the SCANUP bit for the desired scanning order.
- 2. Set the gain and the highest channel number in the scan sequence in Command Register 1. Clear the SCANEN bit during this first write to Command Register 1.
- 3. Write to Command Register 1 again, but this time set the SCANEN bit. This latches the channel number into the scan counter. You must write all other bits in Command Register 1 as you did in the first write, which set the gain and highest channel number.

### Performing Single A/D Conversions

After configuring the analog input circuitry, you can perform single A/D conversions and then read the conversion result from the A/D FIFO. Perform the following steps:

- 1. Set counter A0 so that OUTA0 is high.
- 2. Initiate a conversion by writing to the Start Convert Register.
- 3. Read the conversion result from the A/D FIFO.

You must program counter A0 so that OUTA0 is high. You can do this by writing 0x14 (select counter A0, mode 2) to Counter A Mode Register. This enables the conversions initiated by writing to the Start Convert Register.

Initiate a conversion by writing to the Start Convert Register. When you initiate an A/D conversion, the ADC stores the result in the A/D FIFO at the end of its conversion cycle (approximately 10  $\mu$ s later). You can acquire a specific number of samples by writing to the Start Convert Register that same number of times. In multiple-channel data acquisition, the hardware scans the channels as described before with each write to the Start Convert Register.

You obtain A/D conversion results by reading the A/D FIFO Register. First, you must read the status registers to determine the state of the A/D FIFO. The useful status bits are OVERRUN, OVERFLOW, and DAVAIL in Status Register 1 and FIFOHF\* in Status Register 2. The DAVAIL bit will be set if there is at least one conversion result stored in the A/D FIFO. The DAVAIL bit should become set within a maximum of 12  $\mu$ s after you initiate an A/D conversion. If DAVAIL is set, you can follow the Status Register read by reading the A/D FIFO. If the FIFOHF\* bit is cleared, you can follow the Status Register read with 4,096 consecutive readings of the A/D FIFO. This corresponds to 2,048 samples, since each sample requires two readings of the A/D FIFO. If either the OVERFLOW or the OVERRUN bit is set, an error has occurred. You have either lost at least one conversion by overflowing the A/D FIFO, or you have attempted to initiate a conversion before the previous one has completed.

You must read the A/D FIFO twice to obtain the result. The first reading returns the low byte of the 16-bit data, and the second reading returns the high byte. Reading the A/D FIFO removes the A/D conversion result from the A/D FIFO. If the DAVAIL bit is cleared, then the A/D FIFO is empty and further reading of the A/D FIFO returns meaningless data.

## **Programming a DAQ Operation Using Internal Timing**

A sequence of timed A/D conversions is referred to in this manual as a DAQ operation. The parameters of concern in a DAQ operation are the sample interval, scan interval, and the total number of samples. A sample interval indicates the time to elapse between A/D conversions on each channel in the sequence. The scan interval is the time that elapses between the channel-scanning cycles. The scan interval is only used in interval scanning mode. There are four different internal counters that can help you time these parameters—counter A0, counter A1, counter B0, and counter B1. There are three different DAQ operation modes—interval scanning, controlled, and freerun acquisition mode. The Lab-PC-1200/AI can perform both single-channel data acquisition and multiple-channel data acquisition.

In a controlled acquisition mode, only one counter is required to time the sample intervals. In this mode, you can perform a specified number of conversions, after which the hardware ends the DAQ operation. The number of conversions in a single DAQ operation in this case is limited to a 16-bit count (or 65,535 samples). Use counter A0 to time the sample intervals without any delays. Use counter A1 as the sample counter. Each sample is taken with the same sample interval. Counter A0 is clocked by a 1 MHz clock. The period of counter A0, or the sample interval, is equal to the value programmed into counter A0 is 2  $\mu$ s. The sample interval period, or Counter A0, must be at least 10  $\mu$ s to avoid an overrun error. Choose your sample interval according to the specifications in Chapter 4, *Theory of Operation*, of your *Lab-PC-1200/AI User Manual*, to maintain 12-bit accuracy.

In freerun acquisition mode, only one counter is required for a DAQ operation. Counter A0 continuously generates the conversion pulses as long as GATEA0 is held at a high logic level. The software keeps track of the number of conversions that have occurred and turns off counter A0 after the required number of conversions have been obtained or after some other user-defined criteria have been met. The number of conversions in a single DAQ operation in this case is unlimited.

In an interval scanning acquisition mode, you need two counters. Use counter B1 to time the scan interval. Within the scan interval, each conversion occurs at regular sample intervals timed by counter A0. In multiple-channel data acquisition, the conversions stop after the sample counter A1 counts down to 0. In single-channel interval data acquisition, the conversions stop after the programmed count in the Interval Counter Register has expired. An interval scanning DAQ operation consists of back-to-back scan intervals. Counter B1 is clocked by the same timebase used for counter A0. If the 1 MHz clock is used for the timebase of counter A0, then the period of counter B1, or the scan interval, is the value programmed into counter B1 times 1  $\mu$ s.

Alternatively, a programmable timebase for counter A0 is available by using counter B0. Counter B0 has a fixed, unalterable 2 MHz clock as its own timebase. Therefore, its period is the value programmed into counter B0 multiplied by 500 ns. The minimum period that can be selected for counter B0 is 1  $\mu$ s. The period of counter A0, or the sample interval, is then equal to the period of counter B0 multiplied by the value programmed into counter A0. The maximum sample interval is approximately 35 minutes. You can use counter B0 in any DAQ operation as an alternative timebase for the sample interval.

Programming a DAQ operation requires setting up the four available counters, setting up the Interval Counter Register in single-channel interval scanning mode, and then triggering the DAQ operation. If you are using counter B0 or counter B1 internally for a DAQ operation, be sure that GATB0 or GATB1 are not being driven externally through the I/O connector or are being driven high. Perform the following steps to program a DAQ operation:

- 1. Clear the analog input circuitry.
- 2. Configure the analog input circuitry.
- 3. Set up the four available counters (and the Interval Counter Register if necessary).
- 4. Trigger the operation.
- 5. Service the operation.

Perform the configuration steps and clear the analog input circuitry as described in *Programming the Analog Input Circuitry for Single A/D Conversions* section of this chapter.

### **Programming Counter A0 and Counter B0**

You must always program counter A0 for a DAQ operation using internal timing. A high-to-low transition on OUTA0 (counter A0 output) initiates a conversion. You can program counter A0 to generate a pulse once every  $N \mu s$ , where N is the value programmed into counter A0 and the clock input is a 1 MHz signal. The minimum number that you can use for N is 2. The sample interval can then be programmed to be between 2  $\mu s$  and 65,535  $\mu s$ . Use the following equation to determine the sample interval:

#### *sample interval* = $N \cdot 1 \mu s$

Use the following sequence to program counter A0, the sample interval counter. All writes are 8-bit write operations.

- 1. Write 0x34 (select counter A0, mode 2) to the Counter A Mode Register.
- 2. Write the least significant byte of the sample interval (*N*) to the Counter A0 Data Register.
- 3. Write the most significant byte of the sample interval (*N*) to the Counter A0 Data Register.

You can achieve a larger option of sample intervals by using counter B0 along with counter A0. The resulting sample interval is then  $N_b$  multiplied by  $N_a$  multiplied by 500 ns, where  $N_b$  is the value programmed into counter B0 and  $N_a$  is the value programmed into counter A0. The sample interval can then be programmed to be between 2 µs and [(65,535)<sup>2</sup>/2]µs. Use the following equation to determine the sample interval:

sample interval = 
$$N_a \cdot N_b \cdot 500$$
 ns

Use the following programming sequence to program counter B0 as an alternative timebase for counter A0. All writes are 8-bit operations.

- 1. Set the TBSEL bit in Command Register 2.
- 2. Write 0x36 (select counter B0, mode 3) to the Counter B Mode Register.
- 3. Write the least significant byte of the sample interval  $(N_b)$  to the Counter B0 Data Register.
- 4. Write the most significant byte of the sample interval  $(N_b)$  to the Counter B0 Data Register.

### **Programming Counter A1**

You must program counter A1 even if you are planning on doing a freerun DAQ operation. In a freerun DAQ operation, set the output of counter A1 low to gate counter A0 on and thus allow conversions to occur. The number of samples you want to acquire in a controlled DAQ operation is equal to N + 2, where N is the value programmed into counter A1. Therefore the number of samples can range from 3 to 65,537. Use the following equation to determine the number of samples:

*number of samples* = N + 2

Use the following sequence to program counter A1. All writes are 8-bit operations.

1. Write 0x70 (select counter A1, mode 0) to the Counter A Mode Register. This step sets the output of counter A1 (OUTA1) low.

**Continue** to steps 2 and 3 only for controlled DAQ operations.

- 2. Write the least significant byte of the sample count (*N*) to the Counter A1 Data Register.
- 3. Write the most significant byte of the sample count (*N*) to the Counter A1 Data Register.

### **Programming Counter B1 and the Interval Counter Register**

If you are doing interval scanning data acquisition, you must program counter B1. CLKB1 (the clock input of counter B1) is the same as that used for counter A0 in order to synchronize the individual conversions and the scan interval. The scan interval is equal to *N* (the value programmed into counter B1) multiplied by the clock period used for counter A0. The scan interval should be longer than the total conversion time. In a multiple-channel interval DAQ operation, the total conversion time is equal to the number of channels being scanned, multiplied by the sample interval. In a single-channel interval DAQ operation, the total conversion time is equal to the number loaded into the Interval Counter Register, multiplied by the sample interval.

For single-channel interval scanning data acquisition, the number that you program into the Interval Counter Register should be smaller than the total number of desired samples. For example, if you want to acquire 2,000 samples in batches of 100, load the Interval Counter Register with 100 and the sample counter (counter A1) with 2,000. In this example, 20 scan intervals are required to convert 2,000 samples. The Interval Counter Register is an 8-bit register. Therefore, you can convert up to 255 samples

in a single scan interval. Use the following sequence to program the Interval Counter Register:

- 1. Write the desired number of samples of a single channel that will be acquired during a scan interval to the Interval Counter Data Register.
- 2. Write 0x01 to the Interval Counter Strobe Register to latch the Interval Counter Register.

Use the following sequence to program counter B1 for interval data acquisition. All of the writes are 8-bit operations.

- 1. Set the INTSCAN bit in Command Register 4.
- 2. Clear the EOIRCV bit in Command Register 4.
- 3. Write 0x74 (select counter B1, mode 2) to the Counter B Mode Register.
- 4. Write the least significant byte of the scan interval (*N*) to the Counter B1 Data Register.
- 5. Write the most significant byte of the scan interval (*N*) to the Counter B1 Data Register.

### **Triggering the DAQ Operation**

To start the DAQ operation, set the SWTRIG bit in Command Register 2, which enables counter A0 to start counting. In a freerun DAQ operation (or in any other type of DAQ operation in which you want to stop the operation prematurely), you can stop the DAQ operation by clearing the SWTRIG bit.

**Note** In interval DAQ operations, the first scan interval is not synchronized with counter B1. Therefore you may wish to discard the conversions acquired in the first scan interval.

### Servicing the DAQ Operation

When you start a DAQ operation, you must service the operation by reading the A/D FIFO Register. You can either read the A/D FIFO every time a conversion is available or when the A/D FIFO is half full. You read the A/D FIFO as explained in the section *Performing Single A/D Conversions*. You can also use interrupts to service the DAQ operation. In order to process interrupts, you must install an interrupt handler. See *Programming Options* earlier in this chapter for information on installing interrupt handlers.

Two error conditions, overflow or overrun, may occur during a DAQ operation. If these error conditions occur, the OVERFLOW and/or the OVERRUN bits are set in Status Register 1. Check these bits every time you read Status Register 1 to check the DAVAIL bit.

An overflow condition occurs if more than 4,096 A/D conversions have been stored in the A/D FIFO without the A/D FIFO being read; that is, the A/D FIFO is full and cannot accept any more data. This condition occurs if the software loop reading the A/D FIFO is not fast enough to keep up with the A/D conversion rate. When an overflow occurs, at least one A/D conversion result is lost.

An overrun condition occurs if a second A/D conversion is initiated before the previous conversion is finished. This condition may result in one or more missing A/D conversions. This condition occurs if the sample interval is smaller than the conversion time of the ADC, which is 10  $\mu$ s.

### **Programming a DAQ Operation Using External Timing**

You can use three external timing signals, EXTTRIG, EXTCONV\* and OUTB1, to time a DAQ operation. Use EXTTRIG to initiate a DAQ operation (posttrigger mode) or to terminate an ongoing DAQ operation (pretrigger mode). In posttrigger mode, you use EXTTRIG in place of writing to the SWTRIG bit in Command Register 2. Use the EXTCONV\* signal to time the individual A/D conversions in place of counter A0. If you are performing an interval-scanning DAQ operation and are timing the individual A/D conversions using EXTCONV\*, you must time the scan interval through OUTB1 in place of using counter B1. For signal specifications of these external timing signals, see Chapter 3, *Signal Connections*, in the *Lab-PC-1200/AI User Manual*.

### **Programming a DAQ Operation Using EXTCONV\***

If you want to use EXTCONV\* instead of counter A0 for sample-interval timing, you can follow the same sequence of steps described in *Programming a DAQ Operation Using Internal Timing*, except for programming counters A0 and B0. Replace the steps for programming counters A0 and B0 with the following sequence.

- 1. Write 0x34 (select counter A0, mode 2) to the Counter A Mode Register.
- 2. Clear the ECLKRCV bit in Command Register 4.

You must set up counter A0 to force OUTA0 high and enable conversions initiated by EXTCONV\*. You also need to ensure that the ECLKRCV bit in Command Register 4 is cleared to enable EXTCONV\*.

NoteAfter you trigger the DAQ operation, using either SWTRIG or EXTTRIG in<br/>posttrigger mode, the first EXTCONV\* pulse may not cause an A/D conversion.<br/>See Chapter 3, Signal Connections, in the Lab-PC-1200/AI User Manual for<br/>specifications regarding EXTCONV\*.

### Programming a DAQ Operation Using EXTTRIG in Posttrigger Mode

If you want to use EXTTRIG in posttrigger mode instead of SWTRIG to trigger a DAQ operation, you can follow the same sequence of steps described in *Programming a DAQ Operation Using Internal Timing* except for the section describing triggering of the DAQ operation. Use the following sequence to trigger a DAQ operation using EXTTRIG:

- 1. Set the HWTRIG bit in Command Register 2.
- 2. Trigger a DAQ operation with a rising edge on EXTTRIG.

When you set the HWTRIG bit in Command Register 2, the next rising edge on EXTTRIG will trigger a DAQ operation. Further transitions on EXTTRIG do not affect anything. In a freerun DAQ operation (or in a controlled DAQ operation triggered by EXTTRIG in which you want to stop the operation prematurely), you can stop conversions by first clearing the HWTRIG bit in Command Register 2 and then writing to the A/D FIFO Clear Register. After writing to the A/D FIFO Clear Register, any remaining data in the FIFO will have been cleared.

### Programming a DAQ Operation Using EXTTRIG in Pretrigger Mode

If you want to use EXTTRIG in a pretrigger mode, you must trigger the DAQ operation using SWTRIG, and you must program counter A1 as described in *Programming a DAQ Operation Using Internal Timing*. However, the number of samples that you want to occur after the EXTTRIG trigger is equal to N + 1, where N is the programmed count in counter A1. The number of samples that can occur after the EXTTRIG trigger ranges from 3 to 65,536. Use the following sequence to use EXTTRIG in a pretrigger mode:

- 1. During configuration of the analog input circuitry, set the PRETRIG bit in Command Register 2.
- 2. Clear the analog input circuitry, program the appropriate counters, and trigger the DAQ operation as described in *Programming a DAQ Operation Using Internal Timing*.
- 3. Gate on counter A1 with a low-to-high transition on EXTTRIG.

The first rising edge on EXTTRIG should occur after you trigger the DAQ operation using SWTRIG. This rising edge gates on counter A1. The DAQ operation stops after the programmed count in counter A1 has expired.

### **Programming a DAQ Operation Using OUTB1**

If you want to drive your own interval-scanning pulse on OUTB1 instead of using counter B1 to time the scan interval, you can follow the same sequence of steps as described in *Programming a DAQ Operation Using Internal Timing*, except for the section describing the programming of counter B1. Use the following sequence of steps to use OUTB1 in place of counter B1:

- 1. Program the Interval Counter Register (if necessary).
- 2. Set the INTSCAN bit in Command Register 4.
- 3. Set the EOIRCV bit in Command Register 4.

If you want to externally time the scan interval, you should also externally time the sample interval through EXTCONV\* to synchronize the sample interval and the scan interval. Refer to Chapter 3, *Signal Connections*, in the *Lab-PC-1200/AI User Manual* for timing specifications.

## **DAQ Interrupt Programming**

Five different interrupts can be generated by the analog input circuitry, as follows:

- When a conversion is available to be read from the A/D FIFO
- When the A/D FIFO is half full
- When an error condition (overflow or overrun) is detected
- When a DAQ operation is terminated by counter A1
- When a DMA TC pulse is received

You can enable these five interrupts individually.

Set the FIFOINTEN bit in Command Register 3 to enable interrupt generation when a conversion is available to be read from the A/D FIFO. If FIFOINTEN is set, an interrupt is generated whenever the DAVAIL bit in Status Register 1 is set.

Set the HFINTEN bit in Command Register 6 to enable interrupt generation when the A/D FIFO becomes half full (2,048 samples). If HFINTEN is set, an interrupt is generated whenever the FIFOHF\* bit is cleared in Status Register 2.

Set the ERRINTEN bit in Command Register 3 to enable interrupt generation when an error condition is detected. If ERRINTEN is set, an interrupt is generated whenever either the OVERFLOW or OVERRUN bits are set in Status Register 1.

Set the DQINTEN bit in Command Register 6 to enable interrupt generation when a DAQ operation is terminated by counter A1. If DQINTEN is set, an interrupt is generated whenever OUTA1 is set in Status Register 2.

Set the TCINTEN bit in Command Register 3 to enable interrupt generation when a DMA TC is received. If TCINTEN is set, an interrupt is generated whenever DMA TC is set in Status Register 1.

## **DAQ DMA Programming**

You can program the Lab-PC-1200/AI so that the FIFO generates a DMA request whenever one or more A/D conversions are available in the FIFO. To use DMA, the AT DMA controller must be properly configured. Perform the following steps to configure the board.

- 1. Configure the analog input circuitry.
- 2. Set the DMAEN bit in Command Register 3.
- 3. Program the AT DMA controller.
- 4. Trigger the operation.

The DMA controller automatically transfers data from the AI FIFO into a buffer in system memory when properly configured.

# Programming the Analog Output Circuitry (Lab-PC-1200 Only)

This section explains how to configure the analog output circuitry and how to update the analog output voltage.

### **Configuring the Analog Output Circuitry**

You must configure the analog output circuitry after initializing the Lab-PC-1200/AI and anytime the characteristics of the analog output signals change. Program the appropriate register bits as follows:

- 1. Select the output polarity (unipolar or bipolar) and the coding of the digital code (straight binary or two's complement).
- 2. Set or clear the DAC1UNI/BI\* and DAC0UNI/BI\* bits in Command Register 6 and the 2SDAC1 and 2SDAC0 bits in Command Register 2.

If you select a unipolar output polarity, the straight binary coding is recommended. If you select a bipolar output polarity, the two's complement coding is recommended.

Use the following formula to calculate the output voltage versus digital code for a unipolar analog output configuration and straight binary coding. The digital code is a decimal value ranging from 0 to +4,095.

$$V_{out} = 10.0 \bullet \frac{Code}{4,096}$$

The following formula calculates the output voltage versus digital code for a bipolar analog output configuration and two's complement coding. The digital code is a decimal value ranging from -2,048 to +2,047.

$$V_{out} = 5.0 \bullet \frac{Code}{2.048}$$

Tables 3-1 and 3-2 show the analog output voltage versus the input digital code in unipolar and bipolar mode, respectively.

Digita		
Decimal	Hex	Voltage Output
0	0000	0 V
1	0001	2.4414 mV
2,048	0800	5.0 V
4,095	0FFF	9.9976 V

 
 Table 3-2.
 Analog Output Voltage Versus Digital Code (Unipolar Mode, Straight Binary Coding)

 
 Table 3-3.
 Analog Output Voltage Versus Digital Code (Bipolar Mode, Two's Complement Coding)

Digita		
Decimal	Hex	Voltage Output
-2,048	F800	-5.0 V
-1,024	FC00	2.5 V
0	0000	0.0 V
1,024	0400	2.5 V
2,047	07FF	4.9976 V

### Programming the Update Mode of the Analog Output Circuitry

The analog output circuitry on the Lab-PC-1200/AI uses double-buffered DACs. Therefore the output voltages (DAC0OUT and DAC1OUT on the I/O connector) do not have to be updated immediately with each write to the DAC Data Registers. You can update the analog output in

synchronization with counter A2 or the external update timing signal EXTUPDATE\*. This is useful for waveform generation applications because the timed update pulses eliminate the timing jitter associated with software writes to the DAC Data Registers.

You can operate the analog output circuitry in three ways—immediate update, update on OUTA2, or update on EXTUPDATE\*.

Use the following sequence to program for immediate update mode:

- 1. Clear the LDAC1 or LDAC0 bit in Command Register 2.
- 2. Write the low byte to the DAC0 or DAC1 Low-Byte Register.
- 3. Write the high byte to the DAC0 or DAC1 High-Byte Register.

The analog output voltage is updated immediately after you write the high byte to the DAC0 or DAC1 High-Byte Register.

Use the following sequence to program for update on OUTA2 or EXTUPDATE\*.

- 1. Set the LDAC1 or LDAC0 bit in Command Register 2.
- 2. Set the CNTINTEN bit in Command Register 3 to enable timer interrupts.
- 3. Program counter A2.
- 4. Service the interrupts for waveform generation.

You must program counter A2 even if you are using EXTUPDATE\*. If you are using EXTUPDATE\*, you must set OUTA2 high to enable updates caused by EXTUPDATE\*. If you are using counter A2, do not drive EXTUPDATE\* externally or, if you do, drive it high. You can not block EXTUPDATE\* through software. The clock for counter A2 is the same as that used for counter A0, so you can use a 1 MHz clock or the output of counter B0 to clock counter A2. The update period is equal to the value programmed into counter A2 times the period of the clock. You should set the update period to be longer than the time it takes to write a new value to the DAC Registers. Use the following sequence to program counter A2.

- 1. Write 0xB4 (select counter A2, mode 2) to the Counter A Mode Register.
- **Note** Continue to steps 2 and 3 if you are using counter A2 to update the analog output circuitry.

- 2. Write the least significant byte of the update period to the Counter A2 Data Register.
- 3. Write the most significant byte of the update period to the Counter A2 Data Register.

The update cycle starts immediately after you write the most significant byte to the Counter A2 Data Register. If you are using EXTUPDATE\*, the update cycle starts after setting OUTA2 high and on the first falling edge of EXTUPDATE\*.

### **DAC Interrupt Programming**

Set the CNTINTEN bit in Command Register 3 to enable interrupt generation on the rising edge of either OUTA2 or EXTUPDATE\*. If CNTINTEN is set, then an interrupt is generated whenever the CNTINT bit in Status Register 1 is set.

You service this interrupt by writing a new value to DAC0 or DAC1 Low-Byte and High-Byte Registers. To clear the interrupt, write to the Timer Interrupt Clear Register. This allows continuous waveform generation. The DAC output voltage is then updated by a high-to-low transition on either OUTA2 or EXTUPDATE\*.

## **Programming the Digital I/O Circuitry**

Digital I/O on the Lab-PC-1200/AI uses the 82C55A integrated circuit. Programming the digital I/O circuitry involves setting the mode of the 82C55A by writing to the Digital Control Register and then writing and reading from the three port registers (port A, port B, and port C). The various modes of the 82C55A are illustrated in Appendix D, *OKI MSM*82C55A Data Sheet. Examples for using the digital I/O circuitry are in the Lab-PC-1200/AI User Manual.

You can generate interrupts through PC0 and PC3 on the I/O connector. Enable digital I/O interrupts by setting the DIOINTEN bit in Command Register 3. There are no status bits associated with the digital I/O interrupts. You clear this interrupt by clearing PC0 and PC3.

### **Programming the General-Purpose Counter/Timers**

You can use Counter/Timer Group B of the 82C53 timing circuitry as general-purpose counters when they are not being used for internal timing. To program the general-purpose counters, set the mode of the 82C53 by writing to the Counter B Mode Register, then write and read from the three data registers (counter B0, counter B1, and counter B2). See Appendix C, *OKI MSM82C53 Data Sheet* for information on the various modes of the 82C53. Examples for using the general-purpose counters are given in the *Lab-PC-1200/AI User Manual*. You cannot generate interrupts with the general-purpose counter/timers.

## Calibration

This chapter contains instructions for creating user-defined calibration constants for the Lab-PC-1200/AI CALDACs. This information is important if you do not want to use NI-DAQ to create user-defined calibration constants to be stored in the user areas of the Lab-PC-1200/AI EEPROM. Since the calibration process is quite complicated, the user is advised to use NI-DAQ whenever possible. If NI-DAQ does not support your operating system, only then should you try to write register-level code to perform calibration. Also, if you accidentally overwrite the factory area, you will permanently lose factory-calibration information, and may have to send your unit back to National Instruments for recalibration.

## **Note** National Instruments is not liable for accidental overwriting of the calibration *EEPROM* in the field.

For information concerning writing register-level programs to write to the CALDACs, refer to Chapter 3, *Programming*. For information on calibration equipment requirements, refer to the *Lab-PC-1200/AI User Manual*.

### **Storing User-Defined Constants**

You should store only one set of user-defined calibration constants in one user area. One set of calibration constants consists of twenty constants, six for CALDACs 3, 4 and 7–10, seven for CALDAC6, and seven for CALDAC5 (one at each gain setting). Therefore, one set of calibration constants can calibrate the Lab-PC-1200/AI analog input and analog output circuitry in either bipolar or unipolar polarity and at all gains.

Store your user-defined calibration constants in the same format as that shown for the factory-defined calibration constants in Table 4-1. For example, if you use user area 1, store the calibration constants for CALDACs 3–10 in EEPROM addresses 117–110, then store the seven gain constants for CALDAC6 in EEPROM locations 109–102, followed by the seven postgain offset values for each gain in locations 101–94. Table 4-3 shows the Lab-PC-1200/AI EEPROM map.

Note that the location for gain 1.25 is provided; however, the corresponding value is ignored. Also note that within each user area, the AI CALDAC 3 value is the analog input gain calibration constant for the gain at which the calibration is performed. This value must also be duplicated in the corresponding location in the GAIN *X* value in that user area. For example, if the calibration is performed at gain = 5, then the calibration constant must be written in both locations 114 and 106 for user area 1. Similarly, the postgain offset value for that gain must be duplicated in the corresponding GAIN *X* value-offset location.

When the Lab-PC-1200/AI is shipped, the contents of factory area for bipolar mode are copied in all the user areas. Consequently, there will be constants in the user areas that will be very accurate when the Lab-PC-1200/AI is used in the bipolar mode for both analog input and output.

To save your user-defined calibration constants in the EEPROM, you need programming instructions for writing to the EEPROM. Use the following sequence of steps to store a calibration constant to the EEPROM. For information on user areas in the EEPROM, refer to the EEPROM memory map at the end of this chapter.

- 1. Set the EEPROMCS and WRTPRT\* bits in Command Register 5.
- 2. Serially write the WRITE instruction, 0x06, to the EEPROM to enable a write operation.
- 3. Clear the EEPROMCS bit in Command Register 5.
- 4. Set the EEPROMCS bit in Command Register 5.
- 5. Serially write the WRITE instruction, 0x02, to the EEPROM to start a write operation.
- 6. Serially write the 8-bit user area address to the EEPROM.
- 7. Serially write the 8-bit calibration constant to the EEPROM.
- 8. Clear the EEPROMCS and WRTPRT\* bits in Command Register 5.

Repeat the following two steps eight times to serially write one byte to the EEPROM.

- Clear the SCLK bit in Command Register 5. Write a single bit of the 8-bit value, MSB first, by setting or clearing the SDATA bit in Command Register 5.
- 2. Set the SCLK bit in Command Register 5.

After you store a single calibration constant, you must wait a minimum of 10 ms before accessing the EEPROM again. Alternatively, you may use the RDSR (Read status register) instruction of the X25020 and poll the WIP (write-in-process) bit. If this bit is 1, then the previous write is still in progress. You must wait until this bit reads 0, then you may proceed with the next write.

# **Calibration DACs**

There are eight 8-bit calibration DACs (CALDACs) on the Lab-PC-1200/AI that are used for calibration. These DACs are described in Tables 4-1 and 4-2 for analog input and output calibration respectively. The tolerance in both tables is simply the adjustment range divided by 255 (8-bit). In Table 4-1,  $V_{ref}$  refers to the reference voltage value that you provide during the gain calibration procedure. In Table 4-2,  $V_{ref}$  refers to the voltage that you write to either DAC0 or DAC1 during the gain calibration procedure.

DAC Name	Function	Adjustment Range	Tolerance
CALDAC3	Pregain offset, coarse	820 LSBs (gain = 100)	3.2 LSBs
CALDAC4	Pregain offset, fine	8 LSBs (gain =100)	0.04 LSBs
CALDAC5	Postgain offset	82 LSBs	0.32 LSBs
CALDAC6	Gain	2% of $V_{ref}$	0.008% of $V_{ref}$

Table 4-1. Calibration DAC Characteristics for Analog Input Circuitry

**Table 4-2.** Calibration DAC Characteristics for Analog Output Circuitry

DAC Name	Function	Adjustment Range	Tolerance
CALDAC7	Offset, DAC0	31 LSBs	0.12 LSBs
CALDAC8	Gain, DAC0	2% of $V_{ref}$	0.008% of $V_{ref}$
CALDAC9	Offset, DAC1	31 LSBs	0.12 LSBs
CALDAC10	Gain, DAC1	2% of $V_{ref}$	0.008% of $V_{ref}$

# **Analog Input Calibration**

To null out error sources, you must calibrate the analog input circuitry by adjusting the following potential sources of error (not necessarily in this order):

- Offset error at the input of the instrumentation amplifier.
- Offset error at the input of the ADC.
- Gain error of the analog input circuitry.

Offsets at the input to the instrumentation amplifier contribute gain-dependent error to the analog input system. This offset is multiplied by the gain of the instrumentation amplifier. To calibrate this offset, you must ground the inputs of the instrumentation amplifier, measure the input at two different gains, and adjust CALDAC3 and CALDAC4 until the measured offset in LSBs is independent of the gain setting. Calibration of this pregain offset is done in bipolar mode for both bipolar and unipolar analog input configurations.

Offset error at the input of the ADC is the total of the voltage offsets contributed by the circuitry from the output of the instrumentation amplifier to the ADC input (including the ADC's own offsets). Offset errors appear as a voltage added to the input voltage being measured. To calibrate this offset, you must connect either AGND in bipolar mode or an external voltage source in unipolar mode to the inputs of the ADC and adjust CALDAC5 until the measured voltage is equal to either AGND or (external voltage source + gain error) respectively.

If the three analog input offset DACs are adjusted in this way, there is no significant residual offset error, and reading a grounded channel returns, on average, less than 0.5 LSB regardless of the gain setting.

All the stages up to and including the input of the ADC contribute to the gain error of the analog input circuitry. With the instrumentation amplifier set to a gain of 1, the gain of the analog input circuitry is ideally 1. The gain error is the deviation of the gain from 1 and appears as a multiplication of the input voltage being measured. To eliminate this error source, you must measure the input first with the inputs grounded and then with the inputs connected to the external voltage source. Then adjust CALDAC6 until the measured difference between the two voltages is equal to the value of the external voltage source. After the Lab-PC-1200/AI is calibrated at a gain of 1, there is only a small residual gain error ( $\pm 0.5\%$  max) at the other gains. To reduce this error, calibrate the board at all other gains and for corresponding values stored in the EEPROM User gain area.

In both bipolar and unipolar modes, calibration of postgain offset does not affect the gain characteristics. However, gain calibration does affect postgain offset. Therefore, you must perform gain calibration before postgain calibration.

Perform the calibration procedure for both bipolar and unipolar mode with dither enabled (by setting the DITHEREN bit in Command Register 5) and in referenced single-ended mode (clear the RSE\*/NRSE bit in Command Register 6 and the SE\*/DIFF bit in Command Register 4). Refer to Table 4-1 for calibration tolerances.

## **Bipolar Input Calibration Procedure**

If your board is configured for bipolar input, which provides the -5 to +5 V range, complete the following procedures. This procedure assumes that ADC readings are in the -2,048 to +2,047 range; that is, you have selected the two's complement coding scheme.

Because adjusting the gain affects the postgain offset adjustment, you must calibrate gain before calibrating postgain offset. Also, initialize all of the CALDACs for analog input (3, 4, 5, and 6) to 128 before starting the calibration procedure. This sets each CALDAC at midscale.

# **Pregain Offset Coarse Calibration**

- 1. Connect ACH0 (pin 1 on the rear panel 50-pin I/O connector) to AGND (pin 11).
- 2. Take 1,024 readings from channel 0 at a gain of 1. Take the mean of these readings and call it *mean1*.
- 3. Take 1,024 readings from channel 0 at a gain of 100. Take the mean of these readings and call it *mean100*.
- 4. Adjust CALDAC3 so that:

 $|mean1 - mean100| \le pregain offset coarse calibration tolerance.$ 

# **Pregain Offset Fine Calibration**

- 1. Repeat steps 1–3 of the pregain offset coarse procedure.
- 2. Adjust CALDAC4 so that:

 $|mean1 - mean100| \le pregain offset fine calibration tolerance.$ 

At this point, the pregain offset is nulled out. However, there is a residual postgain offset remaining.

# **Gain Calibration**

- 1. Take 1,024 samples from channel 0 (still connected to AGND) at a gain of 1. Take the mean and call it *postgain\_offset*.
- 2. Connect the voltage reference  $(V_{ref})$  between ACH1 (pin 2) and AGND (pin 11). Choose a voltage reference between 3.0 and 4.5 V.
- 3. Take 1,024 samples from channel 1 at a gain of 1. Take the mean and call it *mean1*.
- 4. Adjust CALDAC6 so that:  $|(mean1 - postgain_offset) - \frac{Vref}{5V} \cdot 2,047| \le gain \ calibration$ tolerance.

# **Postgain Offset Calibration**

- 1. Take 1,024 samples from channel 0 (still connected to AGND) at a gain of 1. Take the mean and call this *postgain\_offset*.
- 2. Adjust CALDAC5 so that:

 $|postgain_offset| \le postgain \ calibration \ tolerance.$ 

# **Calibration at Higher Gains**

If you have performed gain calibration at a gain of 1 and the gain is changed to a value not equal to 1 (2, 5, 10, 20, 50, or 100), you will get a maximum gain error of 0.5%. In addition, the postgain offset will change, so you must recalibrate both the postgain offset (CALDAC5) as well as gain (CALDAC6) at that gain. If you perform gain and postgain offset calibrations at all other gains and store these values in the EEPROM, the maximum gain error will be 0.02% at all gains. Follow the same steps as given in the *Gain Calibration* section, but use a gain not equal to 1.

C Note

When you use a gain not equal to 1, remember that the voltage reference  $(V_{ref})$  multiplied by the gain should be less than 4.5 V.

# **Unipolar Input Calibration Procedure**

If your board is configured for unipolar input, which has an input range of 0 to +10 V, complete the following steps. This procedure assumes that your ADC readings are in the 0 to 4,095 range; that is, you have selected the straight binary coding scheme.

In unipolar mode, the offset can be negative and while doing offset adjustment, all of the acquired samples can be 0 V. This results in incorrect offset calibration. To correct this problem, initialize CALDAC5 to 0 to create a maximum positive offset. Initialize the other CALDACs (3, 4, and 6) to 128 as before.

# **Pregain Offset Calibration**

Follow the same steps as in the *Bipolar Input Calibration Procedure* section for pregain offset coarse and pregain offset fine calibration. Remember to configure the analog input for bipolar mode when performing pregain offset calibration. Reconfigure the analog input for unipolar mode for gain and postgain offset calibration.

After the pregain offset calibration, there will be a residual positive postgain offset remaining because the postgain CALDAC is biased to one extreme.

# **Gain Calibration**

For gain calibration, use the following procedure:

- 1. Take 1,024 samples from channel 0 (still connected to AGND) at a gain of 1. Take the mean and call it *postgain\_offset*.
- 2. Connect the voltage reference  $(V_{ref})$  between ACH1 (pin 2) and AGND (pin 11). Choose a voltage reference between 8.0 and 9.5 V.
- 3. Take 1,024 samples from channel 1 at a gain of 1. Take the mean and call it *mean1*.
- 4. Adjust CALDAC6 so that:

 $|(mean1 - postgain_offset) - \frac{Vref}{10V} \cdot 4,095| \le gain \ calibration \ tolerance.$ 

Perform the calibration for gains not equal to 1 as you did for bipolar input. Remember that the voltage reference ( $V_{ref}$ ) multiplied by the gain used should be less than 9.5 V.

# **Postgain Offset Calibration**

For postgain offset calibration, use the following procedure:

- 1. Take 1,024 samples from channel 1 (still connected to  $V_{ref}$ ) at a gain of 1. Take the mean and call this *mean1*.
- 2. Adjust CALDAC5 so that:

 $|mean1| \le postgain \ calibration \ tolerance.$ 

# Analog Output Calibration (Lab-PC-1200 Only)

To null out error sources that affect the accuracy of the output voltages generated, you must calibrate the analog output circuitry by adjusting the following potential sources of error.

- Analog output offset error
- Analog output gain error

Offset error in the analog output circuitry equals the total of the voltage offsets contributed by each component in the circuitry. This error appears as a voltage difference between the desired voltage and the actual output voltage generated and is independent of the D/A voltage setting. To correct this offset error, the routine sets the D/A to 0 V and adjusts CALDAC7 or CALDAC9 (for DAC0 or DAC1 respectively) until the output voltage is 0 V.

Gain error in the analog output circuitry is the product of the gains contributed by each component in the circuitry. This error appears as a voltage difference between the desired voltage and the actual output voltage generated and is dependent on the D/A voltage setting. To correct this gain error, the routine sets the D/A to a positive voltage ( $V_{ref}$ ) and adjusts CALDAC8 or CALDAC10 (for DAC0 or DAC1 respectively) until the output voltage corresponds to  $V_{ref}$ .

You must calibrate the analog input circuitry before calibrating the analog output circuitry because the output calibration procedure depends on the analog input circuitry. Also, for analog output calibration, set the analog input circuitry calibration to referenced single-ended (RSE), bipolar mode. Refer to Tables 4-1 and 4-2 for calibration tolerances.

# **Bipolar Output Calibration Procedure**

If your board is configured for bipolar output, which provides the -5 to +5 V range, complete the following procedure. This procedure assumes that DAC coding is in the -2,048 to +2,047 range; that is, you have selected the two's complement coding scheme.

Initialize all of the CALDACs for analog output (7, 8, 9, and 10) to 128 before you start the calibration procedure. This sets each CALDAC at midscale. Perform gain calibration before offset calibration.

# **Gain Calibration**

- 1. Write a value  $(-V_{ref})$  to DAC0 or DAC1. Choose a value of  $V_{ref}$  between 3 and 4.5 V.
- 2. Take 1,024 readings from analog input channel 2 (still connected to DAC0OUT or DAC1OUT) at a gain of 1. Take the mean and call it *mean\_gain\_neg\_vref*.
- 3. Write the value  $V_{ref}$  to DAC0 or DAC1.
- 4. Take 1024 readings from analog input channel 2 at a gain of 1. Take the mean and call it *mean\_gain\_vref*.
- 5. Adjust CALDAC8 (or CALDAC10) so that:

 $|(mean\_gain\_ref - mean\_gain\_neg\_ref) - \frac{Vref}{5V} \cdot 4095| \le gain$  calibration tolerance.

# **Offset Calibration**

- 1. Connect DACOOUT, pin 10, (or DAC1OUT, pin 12) to analog input channel 2 (pin 3).
- 2. Write a 0 to DAC0 (or DAC1).
- 3. Take 1,024 readings from analog input channel 2 at a gain of 1. Take the mean and call it *offset*.
- 4. Adjust CALDAC7 (or CALDAC9) so that:

 $|offset| \leq offset \ calibration \ tolerance.$ 

## **Unipolar Output Calibration Procedure**

If your board is configured for unipolar output, which provides the 0 to +10 V range, complete the following procedure. This procedure assumes that DAC coding is in the 0 to +4,095 range; that is, you have selected the straight binary coding scheme.

Set analog input to bipolar mode for analog output unipolar offset calibration. For gain calibration, set the analog input mode to unipolar. Initialize all of the CALDACs for analog output (7–10) to 128 before starting the gain calibration procedure. This sets each CALDAC to mid-scale.

# **Gain Calibration**

- 1. Write 0 to DAC0 (or DAC1). Choose a value between 8.0 and 9.5 V.
- 2. Take 1,024 readings from analog input channel 2 (still connected to DAC0OUT or DAC1OUT) at a gain of 1. Take the mean and call it *mean\_gain\_zero*.
- 3. Write the value  $v_{ref}$  to DAC0 or DAC1. Choose a value between 8.0 and 9.5 V.
- 4. Take 1,024 readings from analog input channel 2 at a gain of 1. Take the mean and call it *mean\_gain\_v<sub>ref</sub>*.
- 5. Adjust CALDAC8 (or CALDAC10) so that:

 $|(mean\_gain\_v_{ref} - mean\_gain\_zero) - \frac{Vref}{10V} \cdot 4,095| \le gain$  calibration tolerance.

# **Offset Calibration**

- 1. Connect DACOOUT, pin 10, (or DAC1OUT, pin 12) to analog input channel 2 (pin 3).
- 2. Write a 0 to DAC0 (or DAC1).
- 3. Take 1,024 readings from analog input channel 2 at a gain of 1. Take the mean and call it *offset*.
- 4. Adjust CALDAC7 (or CALDAC9) so that:

 $|offset| \leq offset \ calibration \ tolerance.$ 

# **EEPROM Map**

Table 4-3 shows part of the EEPROM map for the Lab-PC-1200/AI. Locations 180–255 contain information about the Lab-PC-1200/AI that NI-DAQ uses. These locations are not shown and you should not access them. The factory bipolar area contains locations 156–179 and the factory unipolar area contains 132–155. The user areas are in the lower half of the EEPROM. The pointers from 120–127 are initialized to point to factory locations. To use user area calibration data, you must set these pointers to point to the appropriate user area. Notice that if you point the AI bipolar frame to user area 1, you must also point the corresponding gain and offset pointers to the user area 1 gain and offset frames respectively.

Note NI-DAQ uses these pointers to load a set of calibration constants into the CALDAC each time you run a function pertaining to the Lab-PC-1200/AI. Hence, if you use NI-DAQ along with your register-level code, you must assign these pointers correctly or NI-DAQ will load incorrect values into the CALDACs.

Chapter 4 Calibration

Location	Hex	Decimal	Description
179	00	0	Factory AI CALDAC0 bipolar value
178	00	0	Factory AI CALDAC1 bipolar value
177	00	0	Factory AI CALDAC2 bipolar value
176	00	0	Factory AI CALDAC3 bipolar value
175	00	0	Factory AO CALDAC4 bipolar value
174	00	0	Factory AO CALDAC5 bipolar value
173	00	0	Factory AO CALDAC6 bipolar value
172	00	0	Factory AO CALDAC7 bipolar value
171	00	0	Factory Gain 1 bipolar value-gain
170	00	0	Factory Gain 1.25 bipolar value-gain
169	00	0	Factory Gain 2 bipolar value-gain
168	00	0	Factory Gain 5 bipolar value-gain
167	00	0	Factory Gain 10 bipolar value-gain
166	00	0	Factory Gain 20 bipolar value-gain
165	00	0	Factory Gain 50 bipolar value-gain
164	00	0	Factory Gain 100 bipolar value-gain
163	00	0	Factory Gain 1 bipolar value-offset
162	00	0	Factory Gain 1.25 bipolar value-offset
161	00	0	Factory Gain 2 bipolar value-offset
160	00	0	Factory Gain 5 bipolar value-offset
159	00	0	Factory Gain 10 bipolar value-offset
158	00	0	Factory Gain 20 bipolar value-offset
157	00	0	Factory Gain 50 bipolar value-offset
156	00	0 Factory Gain 100 bipolar value-offset	
155	00	0	Factory AI CALDAC0 unipolar value
154	00	0	Factory AI CALDAC1 unipolar value

Table 4-3. Lab-PC-1200/AI EEPROM Map

Location	Hex	Decimal	Description
			-
153	00	0	Factory AI CALDAC2 unipolar value
152	00	0	Factory AI CALDAC3 unipolar value
151	00	0	Factory AO CALDAC4 unipolar value
150	00	0	Factory AO CALDAC5 unipolar value
149	00	0	Factory AO CALDAC6 unipolar value
148	00	0	Factory AO CALDAC7 unipolar value
147	00	0	Factory Gain 1 unipolar value-gain
146	00	0	Factory Gain 1.25 unipolar value-gain
145	00	0	Factory Gain 2 unipolar value-gain
144	00	0	Factory Gain 5 unipolar value-gain
143	00	0	Factory Gain 10 unipolar value-gain
142	00	0	Factory Gain 20 unipolar value-gain
141	00	0	Factory Gain 50 unipolar value-gain
140	00	0	Factory Gain 100 unipolar value-gain
139	00	0	Factory Gain 1 unipolar value-offset
138	00	0	Factory Gain 1.25 unipolar value-offset
137	00	0	Factory Gain 2 unipolar value-offset
136	00	0	Factory Gain 5 unipolar value-offset
135	00	0	Factory Gain 10 unipolar value-offset
134	00	0	Factory Gain 20 unipolar value-offset
133	00	0	Factory Gain 50 unipolar value-offset
132	00	0	Factory Gain 100 unipolar value-offset
131	00	0	Not used
130	00	0	Not used
129	00	0	Not used
128	00	0	Not used

Table 4-3. Lab-PC-1200/AI EEPROM Map (Continued)

Location	Hex	Decimal	Description	
127	B3	179	Point to AI bipolar frame	
126	A3	155	Point to AI unipolar frame	
125	AF	175	Point to AO bipolar frame	
124	9F	151	Point to AO unipolar frame	
123	AB	171	Point to bipolar Gain frame	
122	9B	147	Point to unipolar Gain frame	
121	00	163	Point to bipolar offset frame	
120	00	139	Point to unipolar offset frame	
119	00	0	Not used	
118	00	0	Not used	
117	00	0	User 1 AI CALDAC0 value	
116	00	0	User 1 AI CALDAC1 value	
115	00	0	User 1 AI CALDAC2 value	
114	00	0	User 1 AI CALDAC3 value	
113	00	0	User 1 AO CALDAC4 value	
112	00	0	User 1 AO CALDAC5 value	
111	00	0	User 1 AO CALDAC6 value	
110	00	0	User 1 AO CALDAC7 value	
109	00	0	User 1 Gain 1 value-gain	
108	00	0	User 1 Gain 1.25 value-gain	
107	00	0	User 1 Gain 2 value-gain	
106	00	0	User 1 Gain 5 value-gain	
105	00	0	User 1 Gain 10 value-gain	
104	00	0	User 1 Gain 20 value-gain	
103	00	0	User 1 Gain 50 value-gain	
102	00	0	User 1 Gain 100 value-gain	

 Table 4-3.
 Lab-PC-1200/AI EEPROM Map (Continued)

Location	Location Hex		Description	
101	00	0	User 1 Gain 1 value-offset	
100	00	0	User 1 Gain 1.25 value-offset	
99	00	0	User 1 Gain 2 value-offset	
98	00	0	User 1 Gain 5 value-offset	
97	00	0	User 1 Gain 10 value-offset	
96	00	0	User 1 Gain 20 value-offset	
95	00	0	User 1 Gain 50 value-offset	
94	00	0	User 1 Gain 100 value-offset	
93	00	0	Not used	
92	00	0	Not used	
91	00	0	User 2 AI CALDAC0 value	
90	00	0	User 2 AI CALDAC1 value	
89	00	0	User 2 AI CALDAC2 value	
88	00	0	User 2 AI CALDAC3 value	
87	00	0	User 2 AO CALDAC4 value	
86	00	0	User 2 AO CALDAC5 value	
85	00	0	User 2 AO CALDAC6 value	
84	00	0	User 2 AO CALDAC7 value	
83	00	0	User 2 Gain 1 value-gain	
82	00	0	User 2 Gain 1.25 value-gain	
81	00	0	User 2 Gain 2 value-gain	
80	00	0	User 2 Gain 5 value-gain	
79	00	0	User 2 Gain 10 value-gain	
78	00	0	User 2 Gain 20 value-gain	
77	00	0	User 2 Gain 50 value-gain	
76	00	0	User 2 Gain 100 value-gain	

Table 4-3. Lab-PC-1200/AI EEPROM Map (Continued)

Location	Hex	Decimal	Description	
75	00	0	User 2 Gain 1 value-offset	
74	00	0	User 2 Gain 1.25 value-offset	
73	00	0	User 2 Gain 2 value-offset	
72	00	0	User 2 Gain 5 value-offset	
71	00	0	User 2 Gain 10 value-offset	
70	00	0	User 2 Gain 20 value-offset	
69	00	0	User 2 Gain 50 value-offset	
68	00	0	User 2 Gain 100 value-offset	
67	00	0	Not used	
66	00	0	Not used	
65	00	0	User 3 AI CALDAC0 value	
64	00	0	User 3 AI CALDAC1 value	
63	00	0	User 3 AI CALDAC2 value	
62	00	0	User 3 AI CALDAC3 value	
61	00	0	User 3 AO CALDAC4 value	
60	00	0	User 3 AO CALDAC5 value	
59	00	0	User 3 AO CALDAC6 value	
58	00	0	User 3 AO CALDAC7 value	
57	00	0	User 3 Gain 1 value-gain	
56	00	0	User 3 Gain 1.25 value-gain	
55	00	0	User 3 Gain 2 value-gain	
54	00	0	User 3 Gain 5 value-gain	
53	00	0	User 3 Gain 10 value-gain	
52	00	0	User 3 Gain 20 value-gain	
51	00	0	User 3 Gain 50 value-gain	
50	00	0	User 3 Gain 100 value-gain	

Table 4-3. Lab-PC-1200/AI EEPROM Map (Continued)

Location Hex Decimal Description					
			-		
49	00	0	User 3 Gain 1 value-offset		
48	00	0	User 3 Gain 1.25 value-offset		
47	00	0	User 3 Gain 2 value-offset		
46	00	0	User 3 Gain 5 value-offset		
45	00	0	User 3 Gain 10 value-offset		
44	00	0	User 3 Gain 20 value-offset		
43	00	0	User 3 Gain 50 value-offset		
42	00	0	User 3 Gain 100 value-offset		
41	00	0	Not used		
40	00	0	Not used		
39	00	0	User 4 AI CALDAC0 value		
38	00	0	User 4 AI CALDAC1 value		
37	00	0	User 4 AI CALDAC2 value		
36	00	0	User 4 AI CALDAC3 value		
35	00	0	User 4 AO CALDAC4 value		
34	00	0	User 4 AO CALDAC5 value		
33	00	0	User 4 AO CALDAC6 value		
32	00	0	User 4 AO CALDAC7 value		
31	00	0	User 4 Gain 1 value-gain		
30	00	0	User 4 Gain 1.25 value-gain		
29	00	0	User 4 Gain 2 value-gain		
28	00	0	User 4 Gain 5 value-gain		
27	00	0	User 4 Gain 10 value-gain		
26	00	0	User 4 Gain 20 value-gain		
25	00	0	User 4 Gain 50 value-gain		
24	00	0	User 4 Gain 100 value-gain		

Table 4-3. Lab-PC-1200/AI EEPROM Map (Continued)

Location	Location Hex Decimal		Description
23	00	0	User 4 Gain 1 value-offset
22	00	0	User 4 Gain 1.25 value-offset
21	00	0	User 4 Gain 2 value-offset
20	00	0	User 4 Gain 5 value-offset
19	00	0	User 4 Gain 10 value-offset
18	00	0	User 4 Gain 20 value-offset
17	00	0	User 4 Gain 50 value-offset
16	00	0	User 4 Gain 100 value-offset
15	00	0	Not used
14	00	0	Not used
13	00	0	Not used
12	00	0	Not used
11	00	0	Not used
10	00	0	Not used
9	00	0	Not used
8	00	0	Not used
7	00	0	Not used
6	00	0	Not used
5	00	0	Not used
4	00	0	Not used
3	00	0	Not used
2	00	0	Not used
1	00	0	Not used
0	00	0	Not used

Table 4-3. Lab-PC-1200/AI EEPROM Map (Continued)



# Fujitsu MB88341/MB88342 Data Sheet<sup>1</sup>

This appendix contains the manufacturer data sheet for the MB88341/MB88342 R-2R type 8-bit D/A converter manufactured by Fujitsu Microelectronics, Inc. The MB88341 D/A converter is used on the Lab-PC-1200/AI.

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October 1989 Edition 1.0

= DATA SHEET =

# MB88341/MB88342 R-2R TYPE 8-BIT D/A CONVERTER

#### DESCRIPTION

The Fujitsu MB88341 and MB88342 are R-2R type 8-bit resolution digital-to-analog converters (DAC), designed for interface with a wide range of general 4-bit and 8-bit microcomputers including Fujitsu's MB8840/50 series and MB88400/500 series 4-bit single-chip microcomputers.

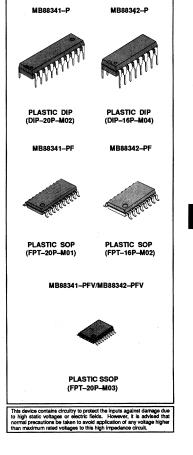
The MB88341 has an 8-bit x 12-channel D/A converter and the MB88342 has an 8-bit x 8-channel D/A converter. Digital data are input serially by individual channel units. The loaded digital data are convertered into analog DC voltages by the D/A converter in 60 µs settling time. The MB88341 and MB88342 are suitable for electronic volumes and replacement for potentiometers for adjustment, in addition to normal D/A converter applications.

## FEATURES

- · Conversion method : R-2R resistor ladder
- MB88341 : 8-bit x 12-channel D/A converter
- MB88342 : 8-bit x 8-channel D/A converter
- Serial data input
- · Serial data output for cascade connection
- 60 µs DAC output settling time
- Two separate power supply/ground lines for digital and analog blocks.
- Single +5V power supply
- Wide operating temperature range: -20°C to +85°C
- Silicon-gate CMOS process
- Three package options :

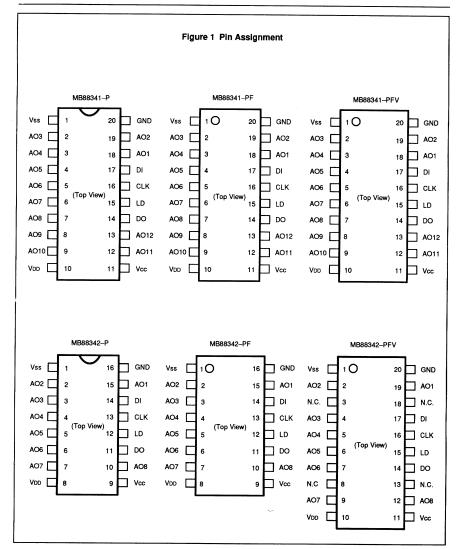
-MB88341: 20-pin plastic DIP (Suffix : -P), 20-pin plastic SOP (Suffix : -PF), 20-pin plastic SSOP (Suffix : -PFV)

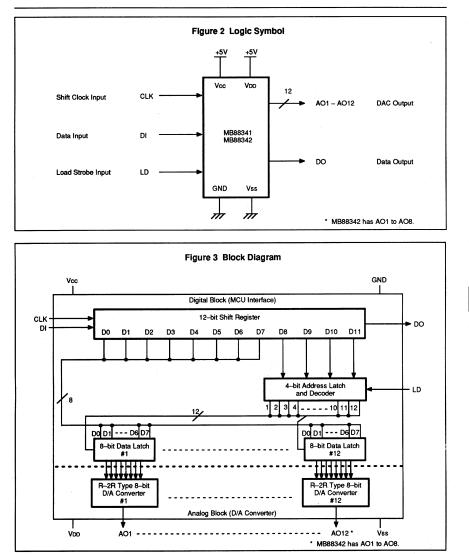
-MB88342: 16-pin plastic DIP (Suffix : -P), 16-pin plastic SOP (Suffix : -PF), 20-pin plastic SSOP (Suffix : -PFV)



FUITSU

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## **PIN DESCRIPTION**

Figure 1 and Table 1 show the pin assignment and pin description of the MB88341 and MB88342.

Symbol	Pin MB88341	No. MB88342	Туре	Name & Function
Power S	upply			
Vcc	11	9 (11)	-	+5V DC power supply pin for the digital block (MCU interface).
GND	20	16 (20)	- '	Ground pin for the digital block (MCU interface).
VDD	10	8 (10)	-	+5V DC power supply pin for the analog block (D/A converter).
Vss	1	1 (1)	-	Ground pin for the analog block (D/A converter).
Control	nput			
CLK	16	13 (16)	1	Shift clock input to the internal 12-bit shift register: At the rising edge of CLK data on the DI pin is shifted into the LSB of the shift register and contents of the shift register are shifted right (to the MSB).
LD	15	12 (15)	l	Load strobe input for a 12-bit address/data : A high level on the LD pin latches a 4-bit address (upper 4 bits: D11 to D8) of the internal 12-bit shift register into the internal address latch/decoder, and writes 8-bit data (lower 8 bits: D7 to D0) of the shift register into an internal data latch selected by the latched address.
Data Inp	ut/Output			
DI	17	14 (17)	I	Serial address/data input to the internal 12-bit shift register: The address/data format is that upper 4 bits (D1 to D8) indicate an address and lower 8 bits (D7 to D0) indicate data. The D11 (MSB) is the first-in bit and D0 (LSB) is the last-in bit.
DO	14	11 (14)	0	Serial address/data output from the internal 12-bit shift register. This is an output pin of the MSB bit data of the 12-bit shift register. This pin allows a cascade connection of the device.
DAC Out	put			
AO1 AO2 AO3 AO4 AO5 AO6 AO7 AO8 AO9 AO10 AO11 AO12	18 19 2 3 4 5 6 7 8 9 12 13	15 (19) 2 (2) 3 (4) 4 (5) 5 (6) 6 (7) 7 (9) 10 (12) - (-) - (-) - (-) - (-)	0	8-bit resolution D/A converter outputs : MB88341: 12 channels (AO1 to AO12) MB88342: 8 channels (AO1 to AO8)

Note : Pin numbers in parentheses are applied to MB88342-PFV.

## **FUNCTIONAL DESCRIPTION**

### **OVERVIEW**

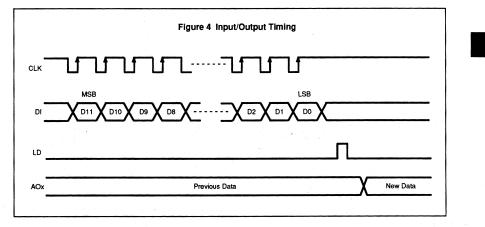
The MB88341 and MB88342 are R–2R resistor ladder type, 8-bit resolution digital-to-analog converter (DAC) devices. The MB88341 has 12 channels, and MB88342 has 8 channels of D/A converters. 8-bit digital data are loaded into internal data latches by individual DAC channel units. The loaded digital data are converted into analog DC voltages through the internal D/A converter in 60 µs settling time. For cascade connection, a serial data output is provided.

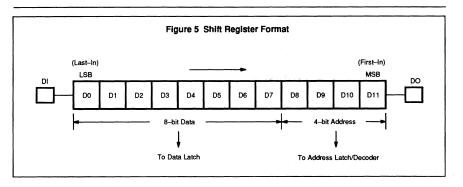
#### **DEVICE CONFIGURATION**

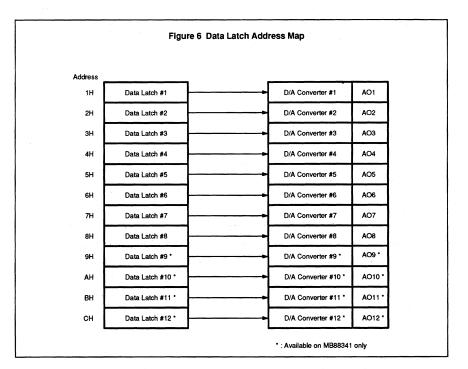
As illustrated in Figure 3 block diagram, the MB88341 (MB88342) device is composed by the digital block (MCU interface) and analog block (D/A converter). The digital block consists of a 12-bit shift register, a 4-bit address latch/decoder, and 12 (8) 8-bit data latches. The analog block includes 12 (8) 8-bit D/A converters connecting to the data latches. For electrically stable operation the power supply and ground lines are separate between the digital block (for MCU interface) and analog block (for D/A converter).

#### **DEVICE OPERATION**

Figure 4 shows the input/output timing. À 12-bit address/data is serially input into the shift register through the DI pin synchronously with the rising edge of CLK. The format of the shift register is shown in Figure 5. The lower 8 bits (D7 to D0) are data bits to be converted, and the upper 4 bits are address bits (D1 to D8) to select a data latch to be written. A high level on the LD pin loads the address latch/decoder with the 4-bit address to select a data latch, and writes the 8-bit data into a selected data latch. Figure 6 shows the data latch address imap, and Table 2, address decoding. 8-bit data written into individual data latches are converted into analog DC voltages, dividing the supply voltage IVoo-Vsst through R--2R resistor ladders of D/A converters. Figure 7 shows the R-2R resistor ladder D/A converter configuration, and Table 3 analog DC voltages corresponding to each digital data.



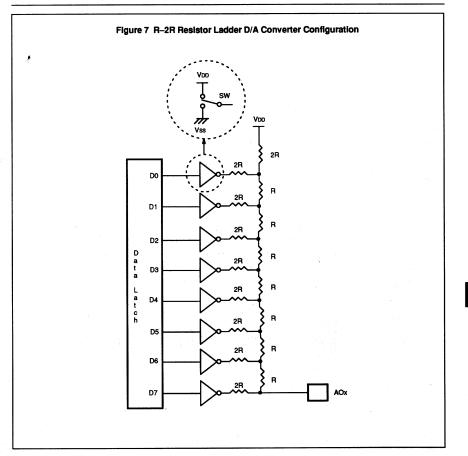




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7–122





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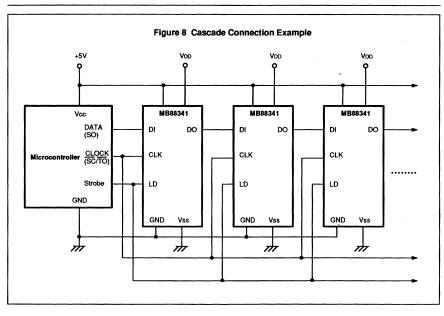
	Addre	\$\$		Data Latch Selected		
08	D9	D10	011	MB88341 MB88342		
0	0	0	0	Desel	ected	
0	0	0	1	Data La	tch #1	
0	. 0	1	0	Data La	tch #2	
0	0	1	1	Data La	tch #3	
0	1	0	0	Data La	tch #4	
0	1	0	1	Data La	tch #5	
0	1	1	0	Data Latch #6		
0	1	1	1	Data Latch #7		
1	0	0	0	Data La	tch #8	
1	0	0	1	Data Latch #9	Deselected	
1	0	1	0	Data Latch #10	Deselected	
1	0	1	1	Data Latch #11 Deselected		
1	1	0	0	Data Latch #12 Deselected		
1	1	0	1	Deselected		
1	1	1	0	Deselected		
1	1	1	1	Deselected		

## Table 2 Address Decoding

7

#### Table 3 Data Conversion

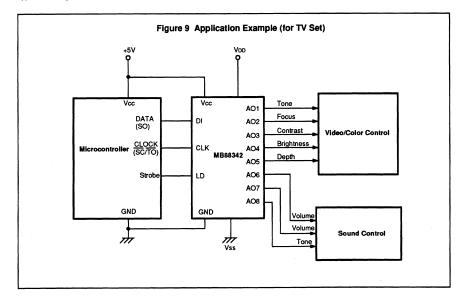
			D	ste				DAC Output Level
07	D6	D6	D4	Da	D2	01	De	AOx
0	0	0	0	0	0	0	0	≈  VDD - Vss   x 1/256
0	0	0	0	0	0	0	1	≈  VDD - Vss   x 2/256
0	0	0	0	0	0	1	0	≈  VDD - Vss   x 3/256
0	0	0	0	0	0	1	1	≈  VDD VSS   x 4/256
:	:	:	:	:	:	:	:	
1	1	1	. 1	1	1	1	0	≈  VDD - Vss   x 255/256
1	1	1	1 <b>1</b> 1	1	1	1	1	≈  VDD – VSS



7

## **APPLICATION DESCRIPTION**

The MB88341 and MB88342 are suitable for electronic volumes and replacement for adjustment potentiometers, in addition to normal D/A converter applications. Figure 8 illustrates an application example for TV set.



7

## **ELECTRICAL CHARACTERISTICS**

## ABSOLUTE MAXIMUM RATINGS<sup>†</sup>

		Rating					
Parameter	Symbol	Min Typ		Max	Unit	Condition	
	Vcc	-0.3		7.0	v	Ta = 25°C	
Supply Voltage	VDD	-0.3		7.0	V	GND = 0 V	
Input Voltage	Vin	-0.3		7.0	v	Ta = 25°C GND = 0 V	
Output Voltage	νουτ	Vout -0.3 7.0	v	Should not exceed Vcc + 0.3V			
Power Dissipation	PD			250	mW		
Operating Ambient Temperature	Та	-20		+85	°C		
Storage Temperature	Тята	-55		+150	°C		

Permanent device damage may occur if the above ABSOLUTE MAXIMUM RATINGS are exceeded. Functional operation should be restricted to the conditions as detailed in the operational sections of this data sheet. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

			Value				
Parameter	Symbol	Min Typ		Max	Unit	Condition	
Supply Voltage	Vcc	4.5	5.0	5.5	v		
(for Digital Block)	GND		0		v		
Supply Voltage (for Analog Block)	Vdd	3.0		Vcc	v	VDD ≤ Vcc,	
	Vss	0		1.0	v	Monotonicity, No load	
Operating Ambient Temperature	Та	-20		+85	°C		

### RECOMMENDED OPERATING CONDITIONS

## DC CHARACTERISTICS

(Recommended operating conditions unless otherwise noted.)

Digital	Block	(MCU	Inter	ace)	

Parameter	Symbol	Min	Velue Typ.	Max	Unit	Condition
Active Supply Current	lcc			1.0	mA	CLK = 1MHz
Standby Supply Current	lccs			10	μA	All inputs (including CLK) fixed at Vcc or GND. All outputs open.
Input Leakage Current	lilk	-10		10	μА	Vin = 0 to Vcc
Input Low Voltage	VIL	1. 1.		0.2•Vcc	v	
Input High Voltage	Viн	0.8•Vcc			v	
Output Low Voltage	Vol			0.4	v	lol = 2.5 mA
Output High Voltage	Vон	Vcc-0.4			v	юн =400 µА

## Analog Block (D/A Converter)

Parameter	Symbol	Min	Value Typ	Max	Unit	Condition	
Supply Current	loo		1.5	3.0	mA	MB88341	Network
Supply Current			1.2	2.5	mA	MB88342	No load
Resolution		8			bit	Monotonicity, lo	UT = -0.01 μA
Variation of Linearity among Channels				±3	LSB	Monotonicity, N	o load

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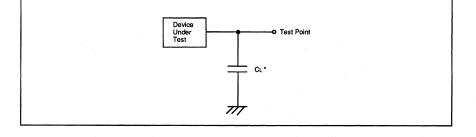
MB88341	
MB88342	

## AC CHARACTERISTICS

(Recommended operating conditions unless otherwise noted.)

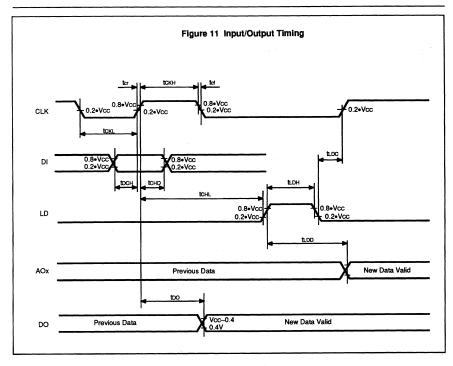
Parameter	Symbol	Value		Unit	Condition	
		Min	Max			
Clock Low Time	<b>tCKL</b>	200		ns		
Clock High Time	тскн	200		ns		
Clock Rise Time	tCr		200	ns		
Clock Fall Time	tcr		200	ns		
Data Setup Time	tDCH	30		ns		
Data Hold Time	tCHD	60		ns		
Load Strobe High Time	1LDH	100		ns		
Load Strobe Setup Time	<b>t</b> CHL	200		ns		
Load Strobe Hold Time	tLDC	100		ns		
DAC Output Settling Time	1LDD		60	μs	No load	
Data Output Delay Time	tDO	70	350	ns	*CL = 20 pF (Min.), 100 pF (Max.)	





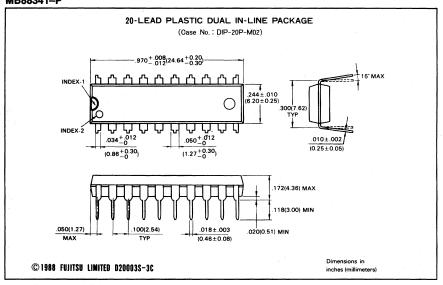
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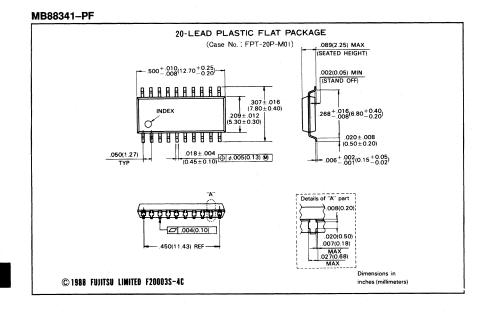
7–130

# PACKAGE DIMENSIONS MB88341-P



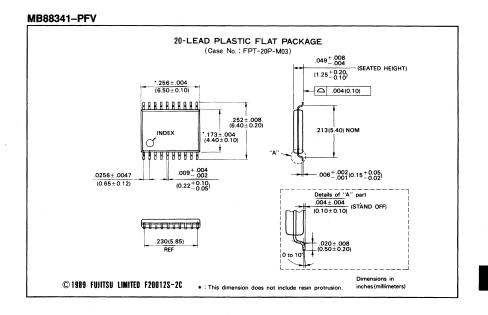
7



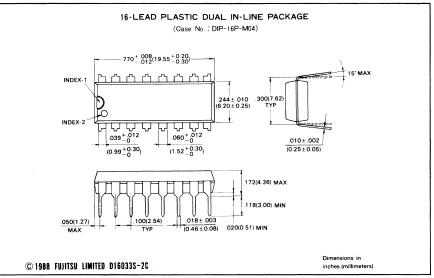


7-132

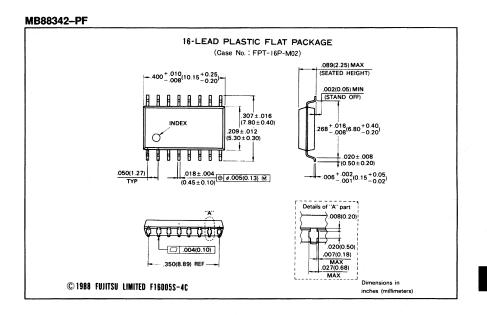






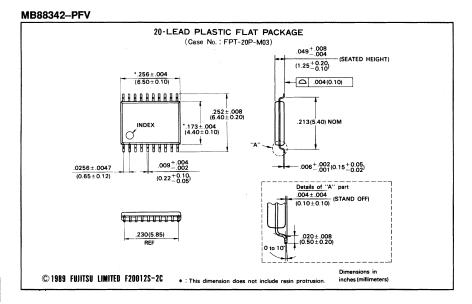






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# MB88341 MB88342



7

# B

# Xicor X25020 Data Sheet<sup>1</sup>

This appendix contains the manufacturer data sheet for the X25020 SPI serial EEPROM manufactured by Xicor, Inc. This EEPROM is used on the Lab-PC-1200/AI.

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	Appl	ICATI	оN No	TES	
Α	VΑ	IL	A B	L	Е
AN9 • .	AN18	• AN3	31 • AN	137 •	AN40



# 256 x 8 Bit

# SPI Serial E<sup>2</sup>PROM With BLOCK LOCK<sup>™</sup> PROTECTION

# **FEATURES**

2K

- 1MHz Clock Rate
- SPI Modes (0,0 & 1,1)
- 256 X 8 Bits
- -4 Byte Page Mode Low Power CMOS -150µA Standby Current
- -3mA Active Write Current
- 2.7V To 5.5V Power Supply
- Block Lock Protection -Protect 1/4, 1/2 or all of E<sup>2</sup>PROM Array
- Built-in Inadvertent Write Protection ---Power-Up/Power-Down protection circuitry -Write Latch
  - -Write Protect Pin
- Self-Timed Write Cycle -5ms Write Cycle Time (Typical)
- High Reliability
  - -Endurance: 100,000 cycles per byte -Data Retention: 100 Years
  - -ESD protection: 2000V on all pins
- 8-Pin Mini-DIP Package
- 8-Lead SOIC Package
- 8-Lead TSSOP

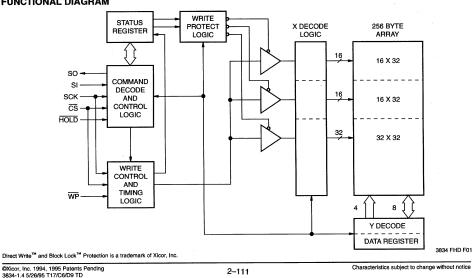
### FUNCTIONAL DIAGRAM

# DESCRIPTION

The X25020 is a CMOS 2048-bit serial E<sup>2</sup>PROM, internally organized as 256 x 8. The X25020 features a serial interface and software protocol allowing operation on a simple three-wire bus. The bus signals are a clock input (SCK) plus separate data in (SI) and data out (SO) lines. Access to the device is controlled through a chip select (CS) input, allowing any number of devices to share the same bus.

The X25020 also features two additional inputs that provide the end user with added flexibility. By asserting the HOLD input, the X25020 will ignore transitions on its inputs, thus allowing the host to service higher priority interrupts. The WP input can be used as a hardwire input to the X25020 disabling all write attempts, thus providing a mechanism for limiting end user capability of altering the memory.

The X25020 utilizes Xicor's proprietary Direct Write™ cell, providing a minimum endurance of 100,000 cycles per byte and a minimum data retention of 100 years.



#### PIN DESCRIPTIONS

#### Serial Output (SO)

SO is a push/pull serial data output pin. During a read cycle, data is shifted out on this pin. Data is clocked out by the falling edge of the serial clock.

#### Serial Input (SI)

SI is the serial data input pin. All opcodes, byte addresses, and data to be written to the memory are input on this pin. Data is latched by the rising edge of the serial clock.

#### Serial Clock (SCK)

The Serial Clock controls the serial bus timing for data input and output. Opcodes, addresses, or data present on the SI pin are latched on the rising edge of the clock input, while data on the SO pin change after the falling edge of the clock input.

#### Chip Select (CS)

When  $\overline{CS}$  is HIGH, the X25020 is deselected and the SO output pin is at high impedance and unless an internal write operation is underway, the X25020 will be in the standby power mode.  $\overline{CS}$  LOW enables the X25020, placing it in the active power mode. It should be noted that after power-up, a HIGH to LOW transition on  $\overline{CS}$  is required prior to the start of any operation.

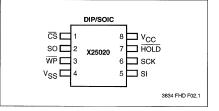
#### Write Protect (WP)

When  $\overline{WP}$  is LOW, nonvolatile writes to the X25020 are disabled, but the part otherwise functions normally. When  $\overline{WP}$  is held HIGH, all functions, including nonvolatile writes operate normally.  $\overline{WP}$  going LOW while  $\overline{CS}$  is still LOW will interrupt a write to the X25020. If the internal write cycle has already been initiated,  $\overline{WP}$  going LOW will have no affect on a write.

#### Hold (HOLD)

HOLD is used in conjunction with the CS pin to select the device. Once the part is selected and a serial sequence is underway, HOLD may be used to pause the serial communication with the controller without resetting the serial sequence. To pause, HOLD must be brought LOW while SCK is LOW. To resume communication, HOLD is brought HIGH, again while SCK is LOW. If the pause feature is not used, HOLD should be held HIGH at all times.

#### **PIN CONFIGURATION**



### PIN NAMES

Symbol	Description
CS	Chip Select Input
SO	Serial Output
SI	Serial Input
SCK	Serial Clock Input
WP	Write Protect Input
V <sub>SS</sub>	Ground
Vcc	Supply Voltage
HOLD Hold Input	

3834 PGM T01.1

2–112

#### **PRINCIPLES OF OPERATION**

The X25020 is a 256 x 8 E<sup>2</sup>PROM designed to interface directly with the synchronous serial peripheral interface (SPI) of many popular microcontroller families.

The X25020 contains an 8-bit instruction register. It is accessed via the SI input, with data being clocked in on the rising SCK. CS must be LOW and the HOLD and WP inputs must be HIGH during the entire operation.

Table 1 contains a list of the instructions and their opcodes. All instructions, addresses and data are transferred MSB first.

Data input is sampled on the first rising edge of SCK after  $\overline{CS}$  goes LOW. SCK is static, allowing the user to stop the clock and then resume operations. If the clock line is shared with other peripheral devices on the SPI bus, the user can assert the  $\overline{HOLD}$  input to place the X25020 into a "PAUSE" condition. After releasing  $\overline{HOLD}$ , the X25020 will resume operation from the point when  $\overline{HOLD}$  was first asserted.

#### Write Enable Latch

The X25020 contains a "write enable" latch. This latch must be SET before a write operation will be completed internally. The WREN instruction will set the latch and the WRDI instruction will reset the latch. This latch is automatically reset upon a power-up condition and after the completion of a byte, page, or status register write cycle.

#### Status Register

The RDSR instruction provides access to the status register. The status register may be read at any time, even during a write cycle. The status register is formatted as follows:

7	6	5	4	3	2	1	0
Х	Х	Х	Х	BP1	BP0	WEL	WIP
						3	834 PGM T02

BP0 and BP1 are set by the WRSR instruction. WEL and WIP are read-only and automatically set by other operations.

The Write-In-Process (WIP) bit indicates whether the X25020 is busy with a write operation. When set to a "1", a write is in progress, when set to a "0", no write is in progress. During a write, all other bits are set to "1".

The Write Enable Latch (WEL) bit indicates the status of the "write enable" latch. When set to a "1", the latch is set, when set to a "0", the latch is reset.

The Block Protect (BP0 and BP1) bits are nonvolatile and allow the user to select one of four levels of protection. The X25020 is divided into four 512-bit segments. One, two, or all four of the segments may be protected. That is, the user may read the segments but will be unable to alter (write) data within the selected segments. The partitioning is controlled as illustrated below.

Status Register Bits		Array Addresses	
BP1	BP0	Protected	
0	0	None	
0	: <b>1</b>	\$C0-\$FF	
1 0		\$80-\$FF	
1	1	\$00\$FF	
	· · · · · · · · · · · · · · · · · · ·	3834 PGN	

#### Table 1. Instruction Set

Instruction Name	Instruction Format*	Operation
WREN	0000 0110	Set the Write Enable Latch (Enable Write Operations)
WRDI	0000 0100	Reset the Write Enable Latch (Disable Write Operations)
RDSR	0000 0101	Read Status Register
WRSR	0000 0001	Write Status Register
READ	0000 0011	Read Data from Memory Array beginning at selected address
WRITE	0000 0010	Write Data to Memory Array beginning at Selected Address (1 to 32 Bytes)

\*Instructions are shown MSB in leftmost position. Instructions are transferred MSB first.

3834 PGM T04

2

#### **Clock and Data Timing**

Data input on the SI line is latched on the rising edge of SCK. Data is output on the SO line by the falling edge of SCK.

#### **Read Sequence**

When reading from the E<sup>2</sup>PROM memory array,  $\overline{CS}$  is first pulled LOW to select the device. The 8-bit READ instruction is transmitted to the X25020, followed by the 8-bit address. After the READ opcode and address are sent, the data stored in the memory at the selected address is shifted out on the SO line. The data stored in memory at the next address can be read sequentially by continuing to provide clock pulses. The address is automatically incremented to the next higher address after each byte of data is shifted out. When the highest address is reached (\$FF) the address counter rolls over to address \$00 allowing the read cycle to be continued indefinitely. The read operation is terminated by taking  $\overline{CS}$  HIGH. Refer to the read E<sup>2</sup>PROM array operation sequence illustrated in Figure 1.

To read the status register CS line is first pulled LOW to select the device followed by the 8-bit RDSR instruction. After the read status register opcode is sent, the contents of the status register are shifted out on the SO line. Figure 2 illustrates the read status register sequence.

#### Write Sequence

Prior to any attempt to write data into the X25020 the "write enable" latch must first be set by issuing the WREN instruction (See Figure 3).  $\overline{CS}$  is first taken LOW, then the WREN instruction is clocked into the X25020. After all eight bits of the instruction are transmitted,  $\overline{CS}$  must then be taken HIGH. If the user continues the write operation without taking  $\overline{CS}$  HIGH after issuing the WREN instruction, the write operation will be ignored.

To write data to the E<sup>2</sup>PROM memory array, the user issues the WRITE instruction, followed by the address and then the data to be written. This is minimally a thirty-two clock operation. CS must go LOW and remain LOW for the duration of the operation. The host may continue to write up to 4 bytes of data to the X25020. The only restriction is that the 4 bytes must reside on the same page. If the address counter reaches the end of the page and the clock continues, the counter will "roll over" to the first address of the page and overwrite any data that may have been written.

For the write operation (byte or page write) to be completed,  $\overline{CS}$  can only be brought HIGH after bit 0 of data byte N is clocked in. If it is brought HIGH at any other time the write operation will not be completed. Refer to Figures 4 and 5 below for a detailed illustration of the write sequences and time frames in which  $\overline{CS}$  going HIGH are valid.

To write to the status register, the WRSR instruction is followed by the data to be written. Data bits 0, 1, 4, 5, 6 and 7 must be "0". Figure 6 illustrates this sequence.

While the write is in progress following a status register or E<sup>2</sup>PROM write sequence, the status register may be read to check the WIP bit. During this time the WIP bit will be HIGH.

#### **Hold Operation**

The HOLD input should be HIGH (at  $V_{IH}$ ) under normal operation. If a data transfer is to be interrupted HOLD can be pulled LOW to suspend the transfer until it can be resumed. The only restriction is the SCK input must be LOW when HOLD is first pulled LOW and SCK must also be LOW when HOLD is released.

The HOLD input may be tied HIGH either directly to  $V_{CC}$  or tied to  $V_{CC}$  through a resistor.

#### **Operational Notes**

The X25020 powers-up in the following state:

• The device is in the low power standby state.

- A HIGH to LOW transition on  $\overline{\text{CS}}$  is required to enter an active state and receive an instruction.
- SO pin is high impedance.
- The "write enable" latch is reset.

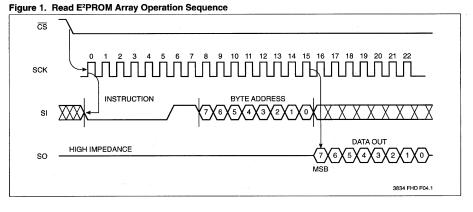
#### Data Protection

The following circuitry has been included to prevent inadvertent writes:

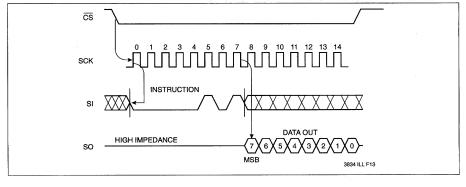
- The "write enable" latch is reset upon power-up.
- A WREN instruction must be issued to set the "write enable" latch.

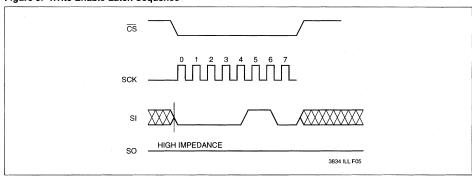
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• CS must come HIGH at the proper clock count in order to start a write cycle.

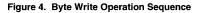


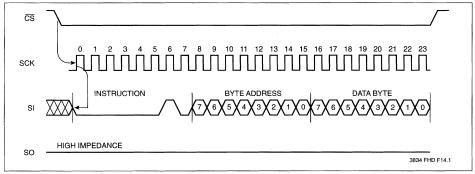
#### Figure 2. Read Status Register Operation Sequence





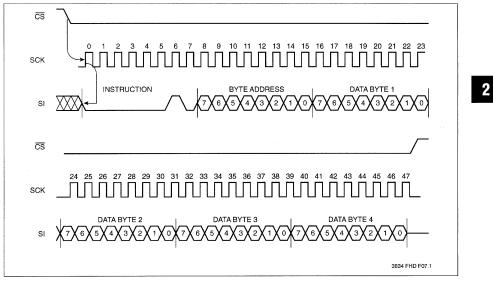


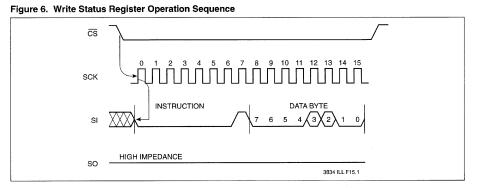




2–116







#### **ABSOLUTE MAXIMUM RATINGS\***

Temperature under Bias	-65°C to +135°C
Storage Temperature	-65°C to +150°C
Voltage on any Pin with Respect to VSS	s −1V to +7V
D.C. Output Current	5mA
Lead Temperature	
(Soldering, 10 seconds)	300°C

#### **RECOMMENDED OPERATING CONDITIONS**

Temp	Min.	Max.
Commercial	0°C	+70°C
Industrial	-40°C	+85°C
Military	–55°C	+125°C
		3834 PGM T05.1

# \*COMMENT

Stresses above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress rating only and the functional operation of the device at these or any other conditions above those listed in the operational sections of this specification is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

Supply Voltage	Limits
X25020	5V ±10%
X25020–3	3V to 5.5V
X25020–2.7	2.7V to 5.5V

3834 PGM T06.1

D.C. OPERATING CHARACTERISTICS	<ul> <li>Over the recommended operating</li> </ul>	conditions unless otherwise specified.)
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	all and a second	Limits				
Symbol	Parameter	Min.	Max.	Units	Test Conditions	
ICC	V <sub>CC</sub> Supply Current (Active)	-	3	mA	SCK = V <sub>CC</sub> x 0.1/V <sub>CC</sub> x 0.9 @ 1MHz, SO = Open	
I <sub>SB</sub>	V <sub>CC</sub> Supply Current (Standby)		150	μA	$\overline{CS} = V_{CC}, V_{IN} = V_{SS} \text{ or } V_{CC} - 0.3V$	
ILI.	Input Leakage Current		10	μA	V <sub>IN</sub> = V <sub>SS</sub> to V <sub>CC</sub>	
LO	Output Leakage Current		10	μA	V <sub>OUT</sub> = V <sub>SS</sub> to V <sub>CC</sub>	
V <sub>IL</sub> (1)	Input LOW Voltage	-1	V <sub>CC</sub> x 0.3	V		
VIH <sup>(1)</sup>	Input HIGH Voltage	V <sub>CC</sub> x 0.7	V <sub>CC</sub> + 0.5	V		
VOL	Output LOW Voltage		0.4	V	I <sub>OL</sub> = 2mA	
VOH	Output HIGH Voltage	V <sub>CC</sub> 0.8		V	$I_{OH} = -1mA$	
		•			3834 PGM T07.3	

#### **POWER-UP TIMING**

Symbol	Parameter	Min.	Max.	Units
tPUR <sup>(2)</sup>	Power-up to Read Operation		1	ms
tPUW <sup>(2)</sup>	Power-up to Write Operation	a an	5	ms

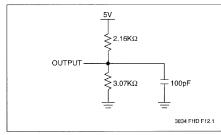
3834 PGM T08

#### **CAPACITANCE** $T_A = +25^{\circ}C$ , f = 1MHz, $V_{CC} = 5V$ .

Symbol	Test	Max.	Units	Conditions
COUT <sup>(2)</sup>	Output Capacitance (SO)	8	pF	$V_{OUT} = 0V$
C <sub>IN</sub> <sup>(2)</sup>	Input Capacitance (SCK, SI, CS, WP, HOLD)	6	pF	$V_{IN} = 0V$
				3834 PGM T09.1

Notes: (1)  $V_{IL}$  min. and  $V_{IH}$  max. are for reference only and are not tested. (2) This parameter is periodically sampled and not 100% tested.

# EQUIVALENT A.C. LOAD CIRCUIT AT 5V VCC



#### A.C. TEST CONDITIONS

Input Pulse Levels	V <sub>CC</sub> x 0.1 to V <sub>CC</sub> x 0.9
Input Rise and Fall Times	10ns
Input and Output Timing Level	V <sub>CC</sub> x 0.5
	3834 PGM T10

2

C. CHARACTERISTICS (Over recommended operating conditions, unless otherwise specified)
--

Data	In	nut	Tim	nnin

Symbol	Parameter	Min.	Max.	Units	
fsck	Clock Frequency	0	1	MHz	
tCYC	Cycle Time	1000		ns	
tLEAD	CS Lead Time	500		ns	
tLAG	CS Lag Time	500		ns	
twn	Clock HIGH Time	400		ns	
twL	Clock LOW Time	400		ns	
tsu	Data Setup Time	100		ns	
tH	Data Hold Time	100		ns	
t <sub>RI</sub>	Data In Rise Time		2	μs	
tFI	Data In Fall Time		2	μs	
tHD	HOLD Setup Time	200		ns	
tCD	HOLD Hold Time	200		ns	
t <sub>CS</sub>	CS Deselect Time	500		ns	
twc <sup>(3)</sup>	Write Cycle Time		10	ms	

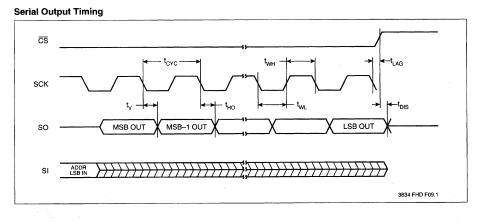
3834 PGM T11.2

# **Data Output Timing**

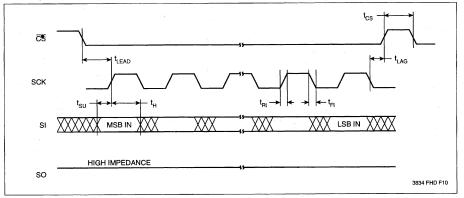
Symbol	Parameter	Min.	Max.	Units	
fsck	Clock Frequency	0	1	MHz	
tDIS	Output Disable Time		500	ns	
tv	Output Valid from Clock LOW		360	ns	
tно	Output Hold Time	0		ns	
tRO	Output Rise Time		300	ns	
tFO	Output Fall Time		300	ns	
tLZ	HOLD HIGH to Output in Low Z	100		ns	
tHZ	HOLD LOW to Output in High Z	100		ns	
				3834 PG	

Notes: (3) two is the time from the rising edge of CS after a valid write sequence has been sent to the end of the self-timed internal nonvolatile write cycle.

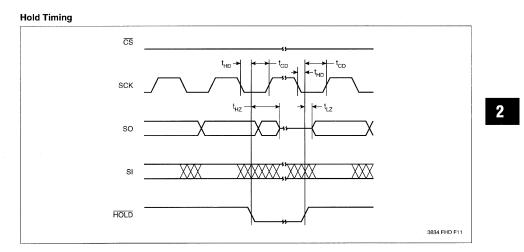
2–119



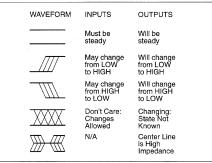




2–120

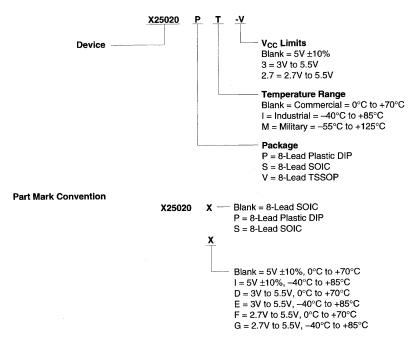


# SYMBOL TABLE



2–121

#### ORDERING INFORMATION



#### LIMITED WARRANTY

Devices sold by Xicor, Inc. are covered by the warranty and patent indemnification provisions appearing in its Terms of Sale only. Xicor, Inc. makes no warranty, express, statutory, implied, or by description regarding the information set forth herein or regarding the freedom of the described devices from patent infringement. Xicor, Inc. makes no warranty of merchantability or fitness for any purpose. Xicor, Inc. reserves the right to discontinue production and change specifications and prices at any time and without notice.

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#### U.S. PATENTS

Xicor products are covered by one or more of the following U.S. Patents: 4,263,664; 4,274,012; 4,300,212; 4,314,265; 4,326,134; 4,393,481; 4,404,475; 4,450,402; 4,486,769; 4,488,060; 4,520,461; 4,533,846; 4,599,706; 4,617,652; 4,666,932; 4,752,912; 4,829,482; 4,874,967; 4,883, 976. Foreign patents and additional patents pending.

#### LIFE RELATED POLICY

In situations where semiconductor component failure may endanger life, system designers using this product should design the system with appropriate error detection and correction, redundancy and back-up features to prevent such an occurrence.

Xicor's products are not authorized for use in critical components in life support devices or systems.

1. Life support devices or systems are devices or systems which, (a) are intended for surgical implant into the body, or (b) support or sustain life, and whose failure to perform, when property used in accordance with instructions for use provided in the labeling, can be reasonably expected to result in a significant injury to the user.

2. A critical component is any component of a life support device or system whose failure to perform can be reasonably expected to cause the failure of the life support device or system, or to affect its safety or effectiveness.

# OKI MSM82C53 Data Sheet<sup>1</sup>

This appendix contains the manufacturer data sheet for the MSM82C53 CMOS programmable interval timer manufactured by OKI Semiconductor, Inc. This counter/timer is used on the Lab-PC-1200/AI.

<sup>&</sup>lt;sup>1</sup> Copyright © OKI Semiconductor, Inc. Reprinted with permission of copyright owner. All rights reserved. OKI Semiconductor. *Microprocessor Data Book 1993*.

# **OKI** semiconductor MSM82C53-2RS/GS/JS

CMOS PROGRAMMABLE INTERVAL TIMER

# GENERAL DESCRIPTION

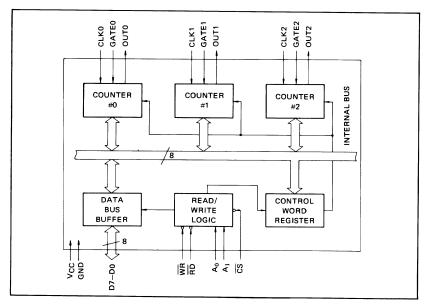
The MSM82C53-2RS/GS/JS are programmable universal timers designed for use in microcomputer systems. Based on silicon gate CMOS technology, it requires a standby current of only 100 µA (max.) when the chip is in the nonselected state. During timer operation, power consumption is still very low with only 8 mA (max.) at 8 MHz of current required.

The devices consist of three independent counters, and can count up to a maximum of 8 MHz (MSM82C53-2) The timer features six different counter modes, and binary count/BCD count functions. Count values can be set in byte or word units, and all functions are freely programmable.

# **FEATURES**

- Maximum operating frequency of 8 MHz (MSM82C53-2)
   Six counter modes available for each counter
- High speed and low power consumption achieved through silicon gate CMOS technology.
- Completely static operation
- Three independent 16-bit down-counters
- 3V to 6V single power supply

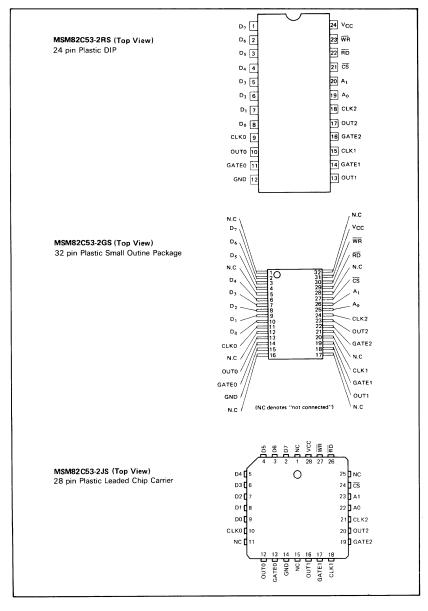
- Binary and decimal counting possible
- •24 pin Plastic DIP (DIP24-P-600)
- 28 pin PLCC (QFJ28-P-S450)
- 32 pin-V Plastic SOP (SSOP32-P-430-VK)



# FUNCTIONAL BLOCK DIAGRAM

# ■ I/O·MSM82C53-2RS/GS/JS ■

# PIN CONFIGURATION



# 

# ABSOLUTE MAXIMUM RATINGS

Parameter	Symbol	Conditions		Limits		
	Symbol	Conditions	MSM82C53-2RS	MSM82C53-2GS	MSM82C53-2JS	Unit
Ssupply Voltage	vcc			v		
Input Voltage	VIN	$V_{IN}$ Respect to GND -0.5 to $V_{CC}$ + 0.5				V
Output Voltage	Vout	1	_	-0.5 to V <sub>cc</sub> + 0.5		
Storage Temperature	T <sub>stg</sub>	- 55 to + 150			°C	
Power Dissipation	PD	Ta = 25°C	0.9	0.7	0.9	w

# **OPERATING RANGES**

Parameter	Symbol	Limits	Conditions	Unit
Supply Voltage	Vcc	3 to 6	V <sub>IL</sub> = 0.2V, V <sub>IH</sub> = V <sub>CC</sub> - 0.2V, operating frequency 2.6 MHz	v
Operating Temperature	Тор	-40 to +85		°c

# **RECOMMENDED OPERATING CONDITIONS**

Parameter	Symbol	Min.	Тур.	Max.	Unit
Supply Voltage	Vcc	4.5	5	5,5	v
Operating Temperature	ТОР	-40	+25	+85	°C
"L" Input Voltage	VIL	-0.3		+0.8	v
"H" Input Voltage	VIH	2.2		V <sub>CC</sub> + 0.3	v

# DC CHARACTERISTICS

Parameter	Symbol	Cond	Min.	Тур.	Max.	Unit	
"L" Output Voltage	VOL	10L = 4mA				0.45	v
"H" Output Voltage	∨он	I <sub>OH</sub> = -1mA		3.7			v
Input Leak Current	1 <sub>L1</sub>	$0 \le V_{IN} \le V_{CC}$	V <sub>CC</sub> =4.5V to 5.5V	-10		10	μA
Output Leak Current	LO	$0 \le V_{OUT} \le V_{CC}$	Ta=40°C to +85°C	-10		10	μA
Standby Supply Current	Iccs	$ \overline{\text{CS}} \geq \text{V}_{\text{CC}} - 0.2\text{V} \\ \text{V}_{\text{IH}} \geq \text{V}_{\text{CC}} - 0.2\text{V} \\ \text{V}_{\text{IL}} \leq 0.2\text{V} $				100	μΑ
Operating Supply Current	<sup>1</sup> cc	<sup>t</sup> CLK = 125 ns C <sub>L</sub> =0pF				8	mA

# ■ I/O·MSM82C53-2RS/GS/JS ■---

# AC CHARACTERISTICS

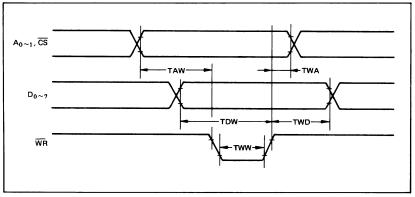
 $(V_{CC}=4.5V\sim5.5V,\ Ta=-40\sim+85^{\circ}C)$ 

		MSM8	2C53-2			
Parameter	Symbol	Min.	Max.	Unit	Co	onditions
Address Set-up Time before reading	TAR	30		ns		CL = 150pF
Address Hold Time after reading	TRA	0		ns	Read	
Read Pulse Width	TRR	150		ns	cycle	
Read Recovery Time	TRVR	200		ns	1	
Address Set-up Time before writing	TAW	0		ns		
Address Hold Time after writing	TWA	20		ns	1	
Write Pulse Width	TWW	150		ns	Write	
Data Input Set-up Time before writing	TDW	100		ns	cycle	
Data Input Hold Time after writing	TWD	20		ns	-	
Write Recovery time	TRVW	200		ns		
Clock Cycle Time	TCLK	125	D.C.	ns		
Clock "H" Pulse Width	TPWH	60		ns		
Clock "L" Pulse Width	TPWL	60		ns	Clock	
"H" Gate Pulse Width	TGW	50		ns	and gate	
"L" Gate Pulse Width	TGL	50		ns	timing	
Gate Input Set-up Time before clock	TGS	50		ns	1	
Gate Input Hold Time after clock	TGH	50		ns	1	
Output Delay Time after reading	TRD		120	ns		
Output Floating Delay Time after reading	TDF	5	90	ns	1	
Output Delay Time after gate	TODG		120	ns	Delay time	
Output Delay Time after clock	TOD		150	ns		
Output Delay Time after address	TAD		180	ns	1	

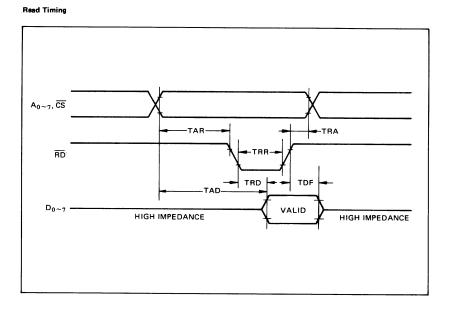
Note: Timing measured at  $V_L = 0.8V$  and  $V_H = 2.2V$  for both inputs and outputs.

# TIME CHART

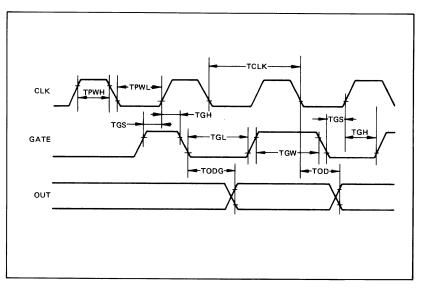
Write Timing







# Clock & Gate Timing

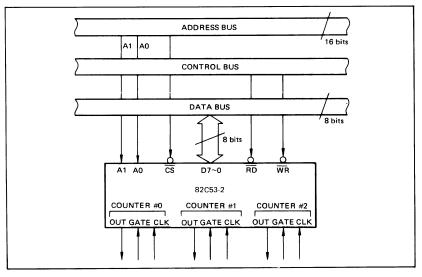


# ■ I/O·MSM82C53-2RS/GS/JS ■-

Pin Symbol	Name	Input/output	Function
D7 ~ D0	Bidirectional data bus	Input/output	Three-state 8-bit bidirectional data bus used when writing control words and count values, and reading count values upon reception of WR and RD signals from CPU.
ĊŚ	Chip select input	Input	Data transfer with the CPU is enabled when this pin is at low level. When at high level, the data bus $(D_0 \text{ thru } D_7)$ is switched to high impedance state where neither writing nor reading can be executed. Internal registers, however, remain unchanged.
RD	Read input	Input	Data can be transferred from MSM82C53 to CPU when this pin is at low level.
WR	Write input	Input	Data can be transferred from CPU to MSM82C53 when this pin is at low level.
A0, A1	Address input	Input	One of the three internal counters or the control word regis- ter is selected by A0/A1 combination. These two pins are normally connected to the two lower order bits of the address bus.
CLK0~2	Clock input	Input	Supply of three clock signals to the three counters incorpo- rated in MSM82C53.
GATE0~2	Gate input	Input	Control of starting, interruption, and restarting of counting in the three respective counters in accordance to the set con- trol word contents.
OUT0~2	Counter output	Output	Output of counter output waveform in accordance with the set mode and count value.

# DESCRIPTION OF PIN FUNCTIONS

# SYSTEM INTERFACING



# DESCRIPTION OF BASIC OPERATIONS

Data transfers between the internal registers and the external data bus is outlined in the following table.

cs	RD	WR	A1	AO	Function
0	1	0	0	0	Data bus to counter # 0 Writing
0	1	0	0	1	Data bus to counter # 1 Writing
0	1	0	1	0	Data bus to counter # 2 Writing
0	1	0	1	1	Data bus to control word register Writing
0	0	1	0	0	Data bus from counter # 0 Reading
0	0	1	0	1	Data bus from counter #1 Reading
0	0	1	1	0	Data bus from counter # 2 Reading
0	0	1	1	1	
1	x	x	x	×	Data bus in high impedance status
0	1	1	x	×	J

x denotes "not specified".

#### DESCRIPTION OF OPERATION

82C53 functions are selected by a control word from the CPU. In the required program sequence, the control word setting is followed by the count value setting and execution of the desired timer operation.

#### **Control Word and Count Value Program**

Each counter operation mode is set by control word programming. The control word format is outlined below.

	D7	Dę	D5	D4	D3	D2	D1	D0
	SC1	SC0	RL1	RLO	M2	M1	MO	BCD
1	Select Counter		Read	/Load	L	Mode		BCD
		(05	5 = 0,	A0, A	1 = 1,1	, RD	= 1, W	R = 0)

• Select Counter (SC0, SC1): Selection of set counter

SC1	SC0	Set Contents
0	0	Counter # 0 selection
0	1	Counter # 1 selection
1	0	Counter # 2 selection
1	1	Illegal combination

 Read/Load (RL1, RL0): Count value Reading/ Loading format setting

RL1	RL0	Set Contents
0	0	Counter Latch operation
0	1	Reading/Loading of Least Significant byte (LSB)
1	0	Reading/Loading of Most Significant byte (MSB)
1	1	Reading/Loading of LSB followed by MSB

Mode (M2, M1, M0): Operation waveform mode setting

M2	M1	мо	Set Contents
0	0	0	Mode 0 (Interrupt on Terminal Count)
0	0	1	Mode 1 (Programmable One-Shot)
x	1	0	Mode 2 (Rate Generator)
x	1	1	Mode 3 (Square Wave Generator)
1	0	0	Mode 4 (Software Triggered Strobe)
1	0	1	Mode 5 (Hardware Triggered Strobe)

x denotes "not specified".

BCD: Operation count mode setting

BCD	Set Contents
0	Binary Count (16-bits Binary)
1	BCD Count (4-decades Binary Coded Decimal)

After setting Read/Load, Mode, and BCD in each counter as outlined above, next set the desired count value. (In some Modes, counting is started immediately after the count value has been written). This count value setting must conform with the Read/Load format set in advance. Note that the internal counters are reset to OOOOH during control word setting. The counter value (OOOOH) can.t be read.

If the two bytes (LSB and MSB) are written at this stage (RL0 and RL1 = 1,1), take note of the following precaution.

Although the count values may be set in the three counters in any sequence after the control word has been set in each counter, count values must be set consecutively in the LSB – MSB order in any one counter.

#### I/O·MSM82C53-2RS/GS/JS =-

#### Example of control word and count value setting

	Counter # 0:	Read/Load	LSB	only,	Mode	З,
		Binary coun	t, coun	t value	зн	
ļ	Counter # 1:	Read/Load	MSB	only,	Mode	5,
		Binary cour	nt, cou	int valu	Je AAO	он
	Counter # 2:	Read/Load	LSB a	nd MSE	3, Mode	0,
		BCD count,	count v	value 12	234	
	B41/1 A 1	EU 7				

MVIA, 1EH	Counter #0 control word entting
	Counter #0 control word setting
MVIA, 6AH]	Counter #1 control word setting
MVIA, B1H	Counter #2 control word setting
MVI A, 03H ]	Counter #0 count value setting
OUT n0 」	Counter #0 count value setting
MVIA, AAH]	Counter #1 count value setting
OUT n1 」	Counter #1 count value setting
MVI A, 34H ]	
OUT n2	Counter #2 count value setting
MVIA, 12H	(LSB then MSB)
OUT n2 🚽	Counter #2 count value setting (LSB then MSB)

Note: n0: Counter #0 address

- n1: Counter #1 address
- n2: Counter #2 address
- n3: Control word register address

#### The minimum and maximum count values which can be counted in each mode are listed below.

Mode	Min.	Max,	Remarks
0	1	0	0 executes 10000H count (ditto in other modes)
1	1	0	
2	2	0	1 cannot be counted
3	2	1	1 executes 10001H count
4	1	0	
5	1	0	

#### Mode Definition

#### Mode 0 (terminal count)

The counter output is set to "L" level by the mode setting. If the count value is then written in the counter with the gate input at "H" level (that is, upon completion of writing the MSB when there are two bytes), the clock input counting is started. When the terminal count is reached, the output is switched to "H" level and is maintained in this status until the control word and count value are set again.

Counting is interrupted if the gate input is switched to "L" level, and restarted when switched back to "H" level.

When Count Values are written during counting, the operation is as follows:

- 1 byte Read/Load.... When the new count value is written, counting is stopped immediately, and then restarted at the new count value by the next clock.
- 2-byte Read/Load.... When byte 1 (LSB) of the new count value is written, counting is stopped immediately. Counting is restarted at the new count value when byte 2 (MSB) is written.

#### Mode 1 (programmable one-shot)

The counter output is switched to "H" level by the mode setting. Note that in this mode, counting is not started if only the count value is written. Since counting has to be started in this mode by using the leading edge of the gate input as a trigger, the counter output is switched to "L" level by the next clock after the gate input trigger. This "L" level status is maintained during the set count value, and is switched back to "H" level when the terminal count is reached.

Once counting has been started, there is no interruption until the terminal count is reached, even if the gate input is switched to "L" level in the meantime. And although counting continues even if a new count value is written during the counting, counting is started at the new count value if another trigger is applied by the gate input.

#### Mode 2 (rate generator)

The counter output is switched to "H" level by the mode setting. When the gate input is at "H" level, counting is started by the next clock after the count value has been written. And if the gate input is at "L" level, counting is started by using the rising edge of the gate input as a trigger after the count value has been set.

An "L" level output pulse appears at the counter output during a single clock duration once every n clock inputs where n is the set count value. If a new count value is written during while counting is in progress, counting is started at the new count value following output of the pulse currently being counted. And if the gate input is switched to "L" level during counting, the counter output is forced to switch to "H" level, the counting being restarted by the rising edge of the gate input.

#### Mode 3 (square waveform rate generator)

The counter output is switched to "H" level by the mode setting. Counting is started in the same way as described for mode 2 above.

The repeated square wave output appearing at the counter output contains half the number of counts as the set count value. If the set count value (n) is an odd number, the repeated square wave output consists of only (n + 1)/2 clock inputs at "H" level and (n - 1)/2 clock inputs at "L" level.

If a new count value is written during counting, the new count value is reflected immediately after the

change ("H" to "L" or "L" to "H") in the next counter output to be executed. The counting operation at the gate input is done the same as in mode 2.

• Mode 4 (software trigger strobe)

The counter output is switched to "H" level by the mode setting. Counting is started in the same way as described for mode 0. A single "L" pulse equivalent to one clock with is generated at the counter output when the terminal count is reached.

This mode differs from 2 in that the "L" level output appears one clock earlier in mode 2, and that pulses are not repeated in mode 4. Counting is

# I/O·MSM82C53-2RS/GS/JS =

stopped when the gate input is switched to "L" level, and restarted from the set count value when switched back to "H" level.

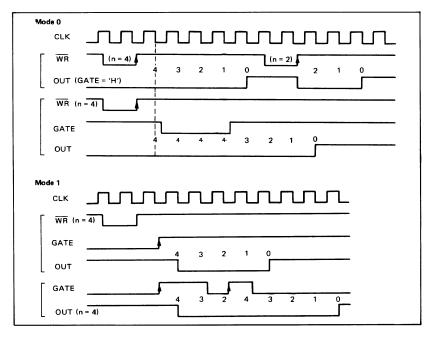
#### Mode 5 (hardware trigger strobe)

The counter output is switched to "H" level by the mode setting. Counting is started, and the gate input used, in the same way as in mode 1.

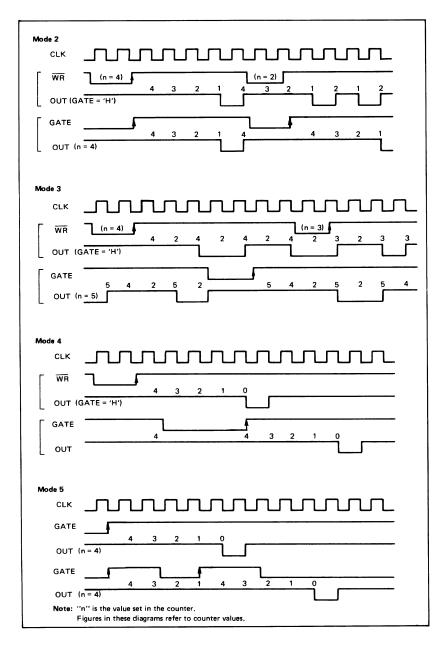
The counter output is identical to the mode 4 output.

The various roles of the gate input signals in the above modes are summarized in the following table.

Gate Mode	"L" Level Falling Edge	Rising Edge	"H" Level
0	Counting not possible		Counting possible
1		<ol> <li>(1) Start of counting</li> <li>(2) Retriggering</li> </ol>	
2	<ol> <li>Counting not possible</li> <li>Counter output forced to "H" level</li> </ol>	Start of counting	Counting possible
3	<ol> <li>Counting not possible</li> <li>Counter output forced to "H" level</li> </ol>	Start of counting	Counting possible
4	Counting not possible		Counting possible
5		(1) Start of counting (2) Retriggering	



■ I/O·MSM82C53-2RS/GS/JS ■-



#### **Reading of Counter Values**

All 82C53 counting is down-counting, the counting being in steps of 2 in mode 3. Counter values can be read during counting by (1) direct reading, and (2) counter latching ("read on the fly").

#### Direct reading

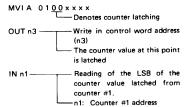
Counter values can be read by direct reading operations.

Since the counter value read according to the timing of the  $\overline{RD}$  and CLK signals is not guaranteed, it is necessary to stop the counting by a gate input signal, or to interrupt the clock input temporarily by an external circuit to ensure that the counter value is correctly read.

#### Counter latching

In this method, the counter value is latched by writing a counter latch command, thereby enabling a stable value to be read without effecting the counting in any way at all. An example of a counter latching program is given below.

Counter latching executed for counter #1 (Read/ Load 2-byte setting)



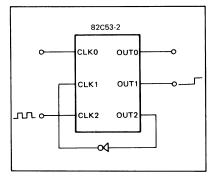
```
MOV B, A
Reading of MSB from counter
```

IN n1 ——— Heading of MSB from counter #1. MOV C, A

# - I/O·MSM82C53-2RS/GS/JS

Example of Practical Application

82C53 used as a 32-bit counter.



Use counter #1 and counter #2

Counter #1: mode 0, upper order 16-bit counter value

Counter #2: mode 2, lower order 16-bit counter value

This setting enables counting up to a maximum of  $2^{32}$ .

# OKI MSM82C55A Data Sheet<sup>1</sup>

This appendix contains the manufacturer data sheet for the MSM82C55A CMOS programmable peripheral interface manufactured by OKI Semiconductor, Inc. This interface is used on the Lab-PC-1200/AI.

<sup>&</sup>lt;sup>1</sup> Copyright © OKI Semiconductor, Inc. Reprinted with permission of copyright owner. All rights reserved. OKI Semiconductor. *Microprocessor Data Book 1993*.

# **OKI** semiconductor MSM82C55A-2RS/GS/VJS

CMOS PROGRAMMABLE PERIPHERAL INTERFACE

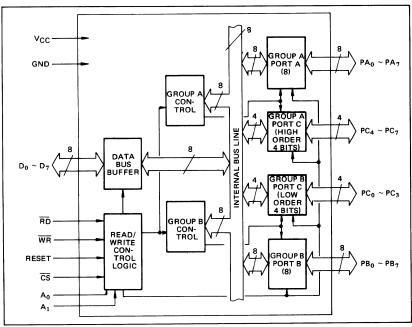
# GENERAL DESCRIPTION

The MSM82C55A is a programmable universal I/O interface device which operates as high speed and on low power consumption due to  $3 \,\mu$  silicon gate CMOS technology. It is the best fit as an I/O port in a system which employs the 8-bit parallel processing MSM80C85A CPU. This device has 24-bit I/O pins equivalent to three 8-bit I/O ports and all inputs/outputs are TTL interface compatible.

# FEATURES

- High speed and low power consumption due to  $3 \mu$  silicon gate CMOS technology
- 3 V to 6 V single power supply
- Full static operation
- Programmable 24-bit I/O ports
- Bidirectional bus operation (Port A)
- Bidirectional bus operation (Fort A)

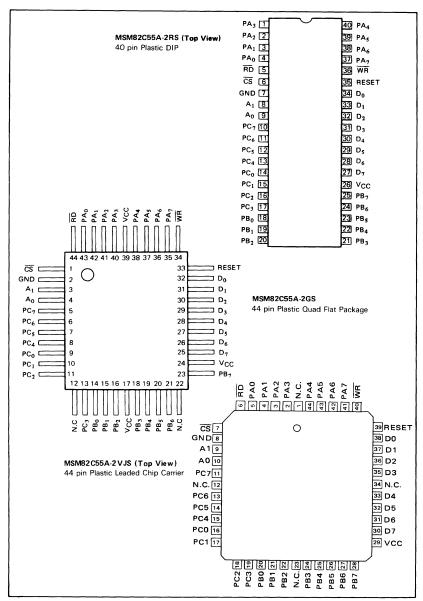
- Bit set/reset function (Port C)
- TTL compatible
- Compatible with 8255A-5
- •40 pin Plastic DIP (DIP40-P-600)
- •44 pin PLCC (QFJ44-P-S650)
- •44 pin-V Plastic QFP (QFP44-P-910-VK)
- •44 pin-VI Plastic QFP (QFP44-P-910-VIK)



278

# CIRCUIT CONFIGURATION

# I/O·MSM82C55A-2RS/GS/VJS



# **PIN CONFIGURATION**

# ■ I/O·MSM82C55A-2RS/GS/VJS ■-----

# ABSOLUTE MAXIMUM RATINGS

D	Gumbal	Conditions		Unit		
Parameter	Symbol	Conditions	MSM82C55A-2RS	MSM82C55A-2GS	MSM82C55A-2VJS	Unit
Ssupply Voltage	Vcc	Ta = 25°C		-0.5 to +7		V
Input Voltage	VIN	with respect	-0.5 to V <sub>cc</sub> + 0.5		V	
Output Voltage	VOUT	to GND	-0.5 to V <sub>cc</sub> + 0.5		V.	
Storage Temperature	T <sub>stg</sub>	-	- 55 to + 150		°C	
Power Dissipation	PD	Ta = 25°C	1.0	0.7	1.0	w

# **OPERATING RANGE**

Parameter	Symbol	Limits	Unit
Supply Voltage	Vcc	3 to 6	v
Operating Temperature	TOP	-40 to 85	°C

# RECOMMENDED OPERATING RANGE

Parameter	Symbol	Min.	Тур.	Max.	Unit
Supply Voltage	Vcc	4.5	5	5.5	v
Operating Temperature	Тор	-40	+25	+85	°C
"L" Input Voltage	VIL	-0.3		+0.8	v
"H" Input Voltage	VIH	2.2		V <sub>CC</sub> +0.3	v

# DC CHARACTERISTICS

-					MSM82C55A-2			
Parameter	Symbol	Conditions		Min.	Typ.	Max.	Unit	
"L" Output Voltage	VOL	I <sub>OL</sub> = 2.5 mA				0.4	V	
"H" Output Voltage	∨он	<sup>I</sup> OH = -40 µA		4.2			v	
		I <sub>OH</sub> = -2.5 mA		3.7			v	
Input Leak Current	1 <sub>L1</sub>	$0 \le V_{IN} \le V_{CC}$	V <sub>CC</sub> = 4.5V to 5.5V	-1		1	μA	
Output Leak Current	LO	$0 \le V_{OUT} \le V_{CC}$	$Ta = -40^{\circ}C$ to	-10		10	μA	
Supply Current (standby)	Iccs	$\label{eq:constraint} \begin{split} \overline{\text{CS}} &\geqq \text{V}_{\text{CC}} \ \text{-}0.2\text{V} \\ \text{V}_{\text{IH}} &\geqq \text{V}_{\text{CC}} \ \text{-}0.2\text{V} \\ \text{V}_{\text{IH}} &\geqq \text{0.2\text{V}} \end{split}$	+85°C (C <sub>L</sub> = 0pF)		0.1	10	μA	
Average Supply Current (active)	ICC	1/O wire cycle 82C55A-2 8MHzCPU timing				8	mA	

# - I/O·MSM82C55A-2RS/GS/VJS

# AC CHARACTERISTICS

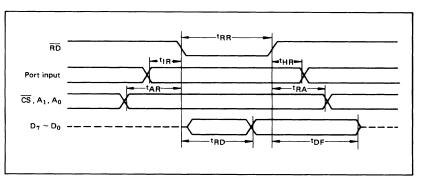
 $(V_{CC} = 4.5 \text{ to } 5.5 \text{V}, \text{ Ta} = -40 \text{ to } +80^{\circ} \text{C})$ 

	Symbol	MSM82C55A-2				
Parameter	Symbol	Min.	Max.	Unit	Remarks	
Setup Time of address to the falling edge of $\overline{\text{RD}}$	<sup>t</sup> AR	20		ns		
Hold Time of address to the rising edge of $\overline{RD}$	<sup>t</sup> RA	0		ns		
RD Pulse Width	<sup>t</sup> RR	100		ns		
Delay Time from the falling edge of $\overline{RD}$ to the output of defined data	<sup>t</sup> RD		120	ns		
Delay Time from the rising edge of $\overline{RD}$ to the floating of data bus	<sup>t</sup> DF	10	75	ns		
Time from the rising edge of RD or WR to the next falling edge of RD or WR	<sup>t</sup> RV	200		ns		
Setup Time of address before the falling edge of $\overline{WR}$	<sup>t</sup> AW	0		ns		
Hold Time of address after the rising edge or WR	tWA	20		ns		
WR Pulse Width	tww	150		ns		
Setup Time of bus data before the rising edge of $\overline{\text{WR}}$	tDW	50		ns		
Holt Time of bus data after the rising edge of WR	tWD	30		ns		
Delay Time from the rising edge of $\overline{WR}$ to the output of defined data	tWB		200	ns		
Setup Time of port data before the falling edge of $\overline{\text{RD}}$	tIR	20		ns		
Hold Time of port data after the rising edge of $\overline{\text{RD}}$	tHR	10		ns		
ACK Pulse Width	<sup>t</sup> AK	100		ns		
STB Pulse Width	<sup>t</sup> ST	100		ns	Load	
Setup Time of port data before the rising edge of STB	tPS	20		ns	150 pF	
Hold Time of port data after the rising edge of STB	tPH	50		ns		
Delay Time from the falling edge of ACK to the output of defined data	<sup>t</sup> AD		150	ns		
Delay Time from the rising edge of $\overline{ACK}$ to the floating of port (Port A in mode 2)	<sup>t</sup> KD	20	250	ns		
Delay Time from the rising edge of WR to the falling edge of $\overline{OBF}$	twoв		150	ns		
Delay Time from the falling edge of ACK to the rising edge of $\overline{\text{OBF}}$	<sup>t</sup> AOB		150	ns		
Delay Time from the falling edge of STB to the rising edge of IBF	<sup>t</sup> SIB		150	ns	_	
Delay Time from the rising edge of $\overline{\text{RD}}$ to the falling edge of IBF	trib		150	ns	_	
Delay Time from the falling edge of RD to the falling edge of INTR	tRIT		200	ns	-	
Delay Time from the rising edge of STB to the rising edge of INTR	<sup>t</sup> SIT		150	ns	_	
Delay Time from the rising edge of ACK to the rising edge of INTR	<sup>t</sup> AIT		150	ns		
Delay Time from the falling edge of $\overline{WR}$ to the falling edge of INTR	twiτ		250	ns		

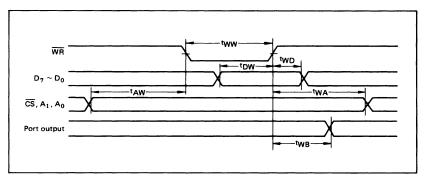
Note: Timing is measured at V  $_L$  = 0.8 V and V  $_H$  = 2.2 V for both input and outputs.

# ■ I/O·MSM82C55A-2RS/GS/VJS ■-

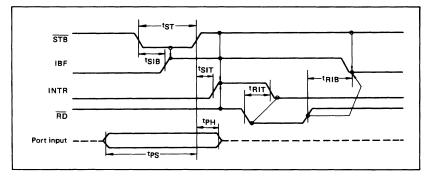




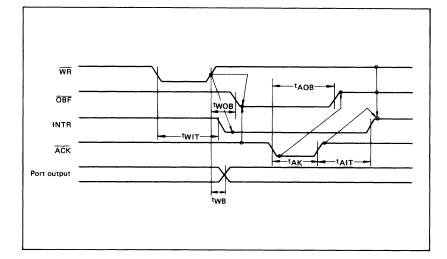
#### **Basic Output Operation (Mode 0)**



Strobe Input Operation (Mode 1)

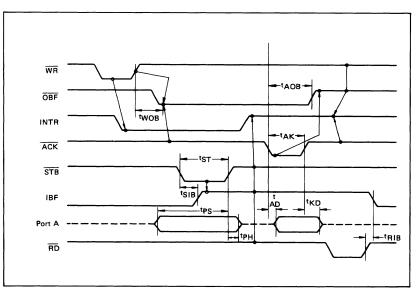


■ I/O·MSM82C55A-2RS/GS/VJS ■



Strobe Output Operation (Mode 1)

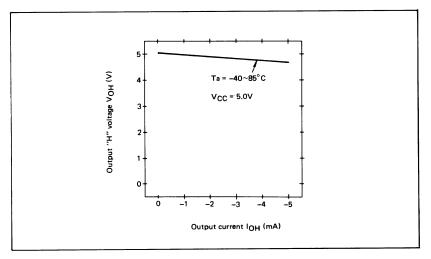




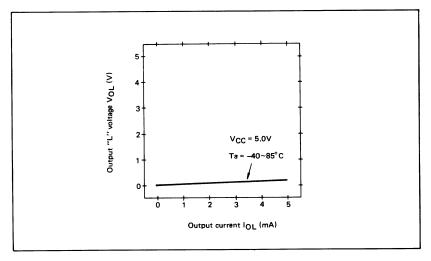
# ■ I/O·MSM82C55A-2RS/GS/VJS ■

# **OUTPUT CHARACTERISTICS (REFERENCE VALUE)**

1 Output "H" Voltage (VOH) vs. Output Current (IOH)



2 Output "L" Voltage (VOL) vs. Output Current (IOL)



Note: The direction of flowing into the device is taken as positive for the output current.

# - I/O·MSM82C55A-2RS/GS/VJS

Pin No.	Item	Input/Output	Function
D7 ~ D0	Bidirectional data bus	Input and output	These are three-state 8-bit bidirectional buses used to write and read data upon receipt of the WR and RD signals from CPUand also used when control words and bit set/reset data are trans- ferred from CPU to MSM82C55A.
RESET	Reset input	Input	This signal is used to reset the control register and all internal registers when it is in high level. At this time, ports are all made into the input mode (high impedance status). all port latches are cleared to 0, and all ports groups are set to mode 0
CS	Chip select input	Input	When the $\overline{CS}$ is in low level, data transmission is enabled with CPU. When it is in high level, the data bus is made into the high impedance status where no write nor read operation is performed. Internal registers hold their previous status, however.
RD	Read input	Input	When RD is in low level, data is transferred from MSM82C55A to CPU.
WR	Write input	Input	When WR is in low level, data or control words are transferred from CPU to MSM82C55A.
A0, A1	Port select input (address)	Input	By combination of A0 and A1, either one is selected from among port A, port B, port C, and control register. These pins are usually connected to low order 2 bits of the address bus.
PA7 ~ PA0	Port A	Input and output	These are universal 8-bit I/O ports. The direction of inputs/out- puts can be determined by writing a control word. Especially, port A can be used as a bidirectional port when it is set to mode 2
PB7 ~ PB0	Port B	Input and output	These are universal 8-bit I/O ports. The direction of inputs/out- puts can be determined by writing a control word.
PC7 ~ PC0	Port C	Input and output	These are universal 8-bit I/O ports. The direction of inputs/out- puts can be determined by writing a control word as 2 ports with 4 bits each. When port A or port B is used in mode 1 or mode 2 (port A only), they become control pins. Especially when port C is used as an output port, each bit can be set/reset independently.
Vcc			+5 V power supply.
GND			GND

# FUNCTIONAL DESCRIPTION OF PIN

#### BASIC FUNCTIONAL DESCRIPTION

#### Group A and Group B

When setting a mode to a port having 24 bits, set it by dividing it into two groups of 12 bits each.

- Group A: Port A (8 bits) and high order 4 bits of port C (PC7 ~ PC4) Group B: Port B (8 bits) and low order 4 bits of
  - port C (PC3 ~ PC0)

#### Mode 0, 1, 2

There are 3 types of modes to be set by grouping as follows:

- Mode 0: Basic input operation/output operation (Available for both groups A and B) Mode 1: Strobe input operation/output opera-
- tion (Available for both groups A and B) Mode 2: Bidirectional bus operation (Available for group A only)

When used in mode 1 or mode 2, however, port C has bits to be defined as ports for control signal for operation ports (port A for group A and port B for group B) of their respective groups.

### Port A, B, C

The internal structure of 3 ports is as follows:

- Port A: One 8-bit data output latch/buffer and one 8-bit data input latch
- Port B: One 8-bit data input/output latch/buffer and one 8-bit data input buffer
- Port C: One 8-bit data output latch/buffer and one 8-bit data input buffer (no latch for input)

#### Single bit set/reset function for port C

When port C is defined as an output port, it is possible to set (to turn to high level) or reset (to turn to low level) any one of 8 bits individually without affecting other bits.

# ■ I/O·MSM82C55A-2RS/GS/VJS ■---

# OPERATIONAL DESCRIPTION

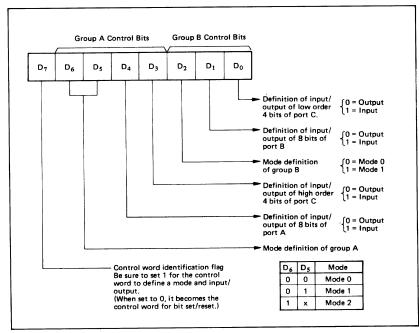
#### **Control Logic**

Operations by addresses and control signals, e.g., read and write, etc. are as shown in the table below:

Operation	A1	AO	cs	WR	RD	Operation
	0	0	0	1	0	Port A →Data Bus
Input	0	1	0	1	0	Port B → Data Bus
	1	0	0	1	0	Port C → Data Bus
Output	0	0	0	0	1	Data Bus → Port A
	0	1	0	0	1	Data Bus → Port B
	1	0	0	0	1	Data Bus →Port C
Control	1	1	0	0	1	Data Bus → Control Register
	1	1	0	1	0	Illegal Condition
Others	×	×	1	×	×	Data bus is in the high impedance status.

#### Setting of Control Word

The control register is composed of 7-bit latch circuit and 1-bit flag as shown below.



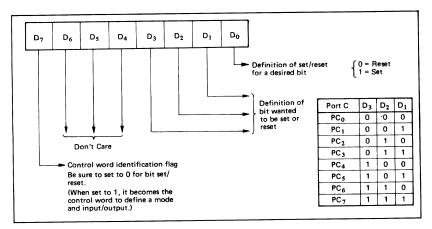
#### Precaution for mode selection

#### **Bit Set/Reset Function**

The output registers for ports A and C are cleared to  $\phi$  each time data is written in the command register and the mode is changed, but the port B state is undefined.

When port C is defined as output port, it is possible to set (set output to 1) or reset (set output to 0) any one of 8 bits without affecting other bits as shown next page.

#### ■ I/O·MSM82C55A-2RS/GS/VJS ■



#### Interrupt Control Function

When the MSM82C55A is used in mode 1 or mode 2, the interrupt signal for the CPU is provided. The interrupt request signal is output from port C. When the interral flip-flop INTE is set beforehand at this time, the desired interrupt request signal is output. When it is reset beforehand, however, the interrupt request signal is not output. The set/reset of the interral flip-flop is made by the bit set/reset operation for port C virtually.

#### Bit set $\rightarrow$ INTE is set $\rightarrow$ Interrupt allowed Bit reset $\rightarrow$ INTE is reset $\rightarrow$ Interrupt inhibited

#### Operational Description by Mode

1. Mode 0 (Besic input/output operation) Mode 0 makes the MSM82C55A operate as a basic input port or output port. No control signals such as interrupt request, etc. are required in this mode. All 24 bits can be used as two-8-bit ports and two 4-bit ports. Sixteen combinations are then possible for inputs/outputs. The inputs are not latched, but the outputs are.

			C	ontro	Wor	d			G	roup A	Group B		
Туре	D7	D <sub>6</sub>	D <sub>5</sub>	D4	D <sub>3</sub>	D <sub>2</sub>	D1	Do	Port A	High Order 4 Bits of Port C	Port B	Low Order 4 Bits of Port C	
1	1	0	0	0	0	0	0	0	Output	Output	Output	Output	
2	1	0	0	0	0	0	0	1	Output	Output	Output	Input	
3	1	0	0	0	0	0	1	0	Output	Output	Input	Output	
4	1	0	0	0	0	0	1	1	Output	Output	Input	Input	
5	1	0	0	0	1	0	0	0	Output	Input	Output	Output	
6	1	0	0	0	1	0	0	1	Output	Input	Output	Input	
7	1	0	0	0	1	0	1	0	Output	Input	Input	Output	
8	1	0	0	0	1	0	1	1	Output	Input	Input	Input	
9	1	0	0	1	0	0	0	0	Input	Output	Output	Output	
10	1	0	0	1	0	0	0	1	Input	Output	Output	Input	
11	1	0	0	1	0	0	1	0	Input	Output	Input	Output	
12	1	0	0	1	0	0	1	1	Input	Output	Input	Input	
13	1	0	0	1	1	0	0	0	Input	Input	Output	Output	
14	1	0	0	1	1	0	0	1	Input	Input	Output	Input	
15	1	0	0	1	1	0	1	0	Input	Input	Input	Output	
16	1	0	0	1	1	0	1	1	Input	Input	Input	Input	

Note: When used in mode 0 for both groups A and B

#### I/O·MSM82C55A-2RS/GS/VJS

#### 2. Mode 1 (Strobe input/output operation)

In mode 1, the strobe, interrupt and other control signals are used when input/output operations are made from a specified port. This mode is available for both groups A and B. In group A at this time, port A is used as the data line and port C as the control signal.

Following is a descrption of the input operation in mode 1.

#### STB (Strobe input)

 When this signal is low level, the data output from terminal to port is fetched into the internal latch of the port. This can be made independent from the CPU, and the data is not output to the data bus until the RD signal arrives from the CPU.

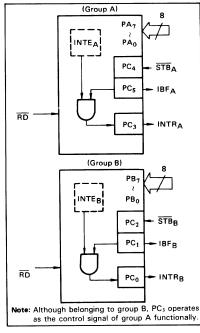
#### IBF (Input buffer full flag output)

 This is the response signal for the STB. This signal when turned to high level indicates that data is fetched into the input latch. This signal turns to high level at the falling edge of STB and to low level at the rising edge of RD.

#### INTR (Interrupt request output)

 This is the interrupt request signal for the CPU of the data fetched into the input latch. It is indicated by high level only when the internal INTE flip-flop is set. This signal turns to high level at the rising edge of the STB (IBF = 1 at this time)

#### Mode 1 Input



and low level at the falling edge of the  $\overline{\text{RD}}$  when the INTE is set.

 $INTE_A$  of group A is set when the bit for PC<sub>4</sub> is set, while  $INTE_B$  of group B is set when the bit for PC<sub>2</sub> is set.

Following is a description of the output operation of mode 1.

#### **OBF** (Output buffer full flag output)

 This signal when turned to low level indicates that data is written to the specified port upon receipt of the WR signal from the CPU. This signal turns to low level at the rising edge of the WR and high level at the falling edge of the ACK.

#### ACK (Acknowledge input)

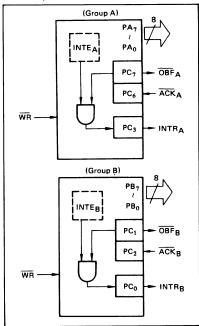
 This signal when turned to low level indicates that the terminal has received data.

#### INTR (Interrupt request output)

 This is the signal used to interrupt the CPU when a terminal receives data from the CPU via the MSM82C55A-5. It indicates the occurrence of the interrupt in high level only when the internal INTE flip-flop is set. This signal turns to high level at the rising edge of the ACK (OBF = 1 at this time) and low level at the falling edge of WR when the INTE<sub>B</sub> is set.

 $INTE_A$  of group A is set when the bit for PC<sub>6</sub> is set, while  $INTE_B$  of group B is set when the bit for PC<sub>2</sub> is set.

#### Mode 1 output



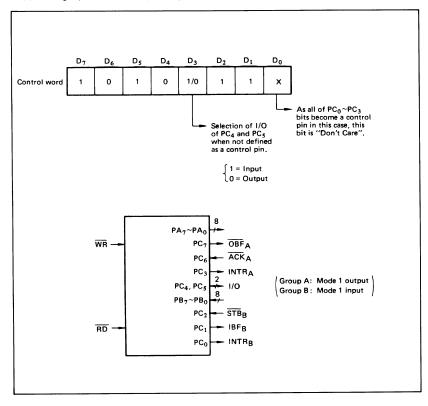
### ■ I/O·MSM82C55A-2RS/GS/VJS ■

Combination of Input/Output Port C	Group A: Input Group B: Input	Group A: Input Group B: Output	Group A: Output Group B: Input	Group A: Output Group B: Output
PC <sub>0</sub>	INTRB	INTRB	INTRB	INTRB
PC1	IBFB	OBFB	IBFB	OBFB
PC <sub>2</sub>	STBB	ACKB	STBB	ACKB
PC <sub>3</sub>	INTRA	INTRA	INTRA	INTRA
PC4	STBA	STBA	I/O	I/O
PC <sub>5</sub>	IBFA	IBFA	I/O	1/0
PC <sub>6</sub>	1/0	1/0	ACKA	ACKA
PC <sub>7</sub>	1/0	I/O	OBFA	OBFA

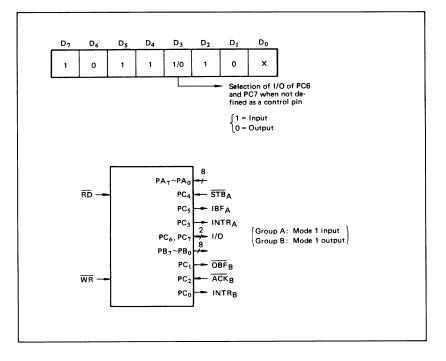
#### Port C Function Allocation in Mode 1

Note: I/O is a bit not used as the control signal, but it is available as a port of mode 0.

Examples of the relation between the control words and pins when used in mode 1 is shown below: (a) When group A is mode 1 output and group B is mode 1 input.



#### I/O·MSM82C55A-2RS/GS/VJS =



(b) When group A is mode 1 input and group B is mode 1 output.

#### 3. Mode 2 (Strobe bidirectional bus I/O operation)

In mode 2, it is possible to transfer data in 2 directions through a single 8-bit port. This operation is akin to a combination between input and output operations. Port C waits for the control signal in this case, too. Mode 2 is available only for group A, however.

#### Next, a description is made on mode 2. OBF (Output buffer full flag output)

 This signal when turned to low level indicates that data has been written to the internal output latch upon receipt of the WR signal from the CPU. At this time, port A is still in the high impedance status and the data is not yet output to the outside. This signal turns to low level at the rising edge of the ACK.

#### ACK (Acknowledge input)

 When a low level signal is input to this pin, the high impedance status of port A is cleared, the buffer is enabled, and the data written to the internal output latch is output to port A. When the input returns to high level, port A is made into the high impedance status.

#### STB (Strobe input)

When this signal turns to low level, the data output to the port from the pin is fetched into the internal input latch. The data is output to the data bus upon receipt of the RD signal from the CPU, but it remains in the high impedance status until then.

#### IBF (Input buffer full flag output)

 This signal when turned to high level indicates that data from the pin has been fetched into the input latch. This signal turns to high level at the falling edge of the STB and low level at the rising edge of the RD.

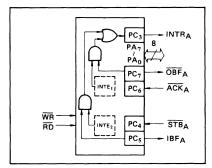
#### INTR (Interrupt request output)

 This signal is used to interrupt the CPU and its operation in the same as in mode 1. There are two INTE flip-flops internally available for input and output to select either interrupt of input or output operation. The INTE1 is used to control the interrupt request for output operation and it can be reset by the bit set for PC6. INTE2 is used to control the interrupt request for the input operation and it can be set by the bit set for PC4.

290

#### - I/O·MSM82C55A-2RS/GS/VJS

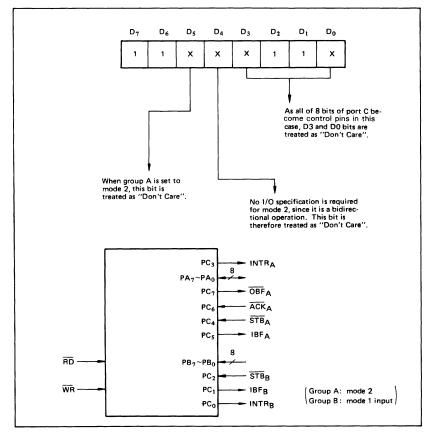
#### Mode 2 I/O Operation



Port C	Function
PC <sub>0</sub>	0 11 11 11
PC1	Confirmed to the group B mode
PC <sub>2</sub>	group b mode
PC <sub>3</sub>	INTRA
PC4	STBA
PC 5	IBFA
PC <sub>6</sub>	ACKA
PC7	OBFA

Port C Function Allocation in Mode 2

Following is an example of the relation between the control word and the pin when used in mode 2. When input in mode 2 for group A and in mode 1 for group B.



#### ■ I/O·MSM82C55A-2RS/GS/VJS ■

4. When Group A is Different in Mode from Group B Group A and group B can be used by setting them in different modes each other at the same time. When either group is set to model or mode 2, it is

possible to set the one not defined as a control pin in port C to both input and output as a port which operates in mode 0 at the 3rd and 0th bits of the control word.

	<b>C  A</b>	6				Por	rt C			
	Group A	Group B	PC7	PC <sub>6</sub>	PC <sub>5</sub>	PC4	PC3	PC <sub>2</sub>	PC1	PC <sub>0</sub>
1	Mode 1 input	Mode 0	1/0	1/0	IBFA	STBA	INTRA	I/O	1/0	1/0
2	Mode 0 output	Mode 0	OBFA	ACKA	1/0	1/0	INTRA	I/O	1/0	1/0
3	Mode 0	Mode 1 input	1/0	1/0	1/0	1/0	1/0	STBB	IBFB	INTRB
4	Mode 0	Mode 1 output	1/0	1/0	1/0	1/0	1/0	ACKB	OBFB	
5	Mode 1 input	Mode 1 input	1/0	I/O	IBFA	STBA	INTRA	STBB	IBFB	
6	Mode 1 input	Mode 1 output	1/0	1/0	IBFA	STBA	INTRA	ACKB	OBFB	INTRB
7	Mode 1 output	Mode 1 input	OBFA	ACKA	I/O	1/0	INTRA	STBB	IBFB	
8	Mode 1 output	Mode 1 output	OBFA	ACKA	I/O	1/0	INTRA	ACKB	OBFB	
9	Mode 2	Mode 0	OBFA	ĀCKĄ	IBFA	STBA	INTRA	1/0	1/0	1/0

(Mode combinations that define no control bit at port C)

Controlled at the 3rd bit (D3) of the control word

Controlled at the 0th bit (D0) of the control word

When the I/O bit is set to input in this case, it is possible to access data by the normal port C read operation.

When set to output, PC7 ~ PC4 bits can be accessed by the bit set/reset function only. Meanwhile, 3 bits from PC2 to PC0 can be accessed by normal write operation. The bit set/reset function can be used for all of PC3  $\sim$  PC0 bits. Note that the status of port C varies according to the combination of modes like this.

#### - I/O·MSM82C55A-2RS/GS/VJS -

#### 5. Port C Status Read

When port C is used for the control signal, that is, in either mode 1 or mode 2, each control signal and

bus status signal can be read out by reading the content of port C. The status read out is as follows:

					Sta	tus read or	n the data l	bus		
	Group A	Group B	D <sub>7</sub>	D <sub>6</sub>	D <sub>5</sub>	D <sub>4</sub>	D <sub>3</sub>	D <sub>2</sub>	D <sub>1</sub>	Do
1	Mode 1 input	Mode 0	1/0	1/0	IBFA	INTEA	INTRA	1/0	1/0	I/O
2	Mode 1 output/	Mode 0	OBFA	INTEA	·1/O	I/O	INTRA	1/0	ı/o	1/0
3	Mode 0	Mode 1 input	1/0	1/0	1/0	1/0	1/0	INTEB	IBFB	INTRB
4	Mode 0	Mode 1 output	1/0	1/0	1/0	1/0	1/0	INTEB	OBFB	INTRB
5	Mode 1 input	Mode 1 input	1/0	1/0	IBFA	INTEA	INTRA	INTEB	IBFB	INTRB
6	Mode 1 input	Mode 1 output	1/0	1/0	IBFA	INTEA	INTRA	INTEB	OBFB	
7	Mode 1 output	Mode 1 input	OBFA	INTEA	1/0	1/0	INTRA	INTEB	IBFB	INTRB
8	Mode 1 output	Mode 1 output	OBFA	INTEA	1/0	1/0		INTEB	OBFB	
9	Mode 2	Mode 0	OBFA	INTE <sub>1</sub>	IBFA	INTE <sub>2</sub>	INTRA	1/0	1/0	1/0
10	Mode 2	Mode 1 input	OBFA	INTE 1	IBFA	INTE2		INTEB	IBFB	
11	Mode 2	Mode 1 output	OBFA	INTE 1	IBFA	INTE2	INTRA	INTEB	OBFB	

#### 6. Reset of MSM82C55A

Be sure to keep the RESET signal at power ON in the high level at least for 50  $\mu$ s. Subsequently, it

becomes the input mode at a high level pulse above 500 ns.

#### Note: Comparison of MSM82C55A-5 and MSM82C55A-2

#### MSM82C55A-5

After a write command is executed to the command register, the internal latch is cleared in PORTA PORTC. For instance, 00H is output at the beginning of a write command when the output port is assigned. However, if PORTB is not cleared at this time, PORTB is unstable. In other words, PORTB only outputs ineffective data (unstable value according to the device) during the period from after a write command is executed till the first data is written to PORTB.

#### MSM82C55A-2

After a write command is executed to the command register, the internal latch is cleared in All Ports(PORTA,PORTB,PORTC). OOH is ontput at the beginning of a write command when the output port is assigned.

# E

# **Customer Communication**

For your convenience, this appendix contains forms to help you gather the information necessary to help us solve your technical problems and a form you can use to comment on the product documentation. When you contact us, we need the information on the Technical Support Form and the configuration form, if your manual contains one, about your system configuration to answer your questions as quickly as possible.

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United States: 512 794 5422 Up to 14,400 baud, 8 data bits, 1 stop bit, no parity United Kingdom: 01635 551422 Up to 9,600 baud, 8 data bits, 1 stop bit, no parity France: 01 48 65 15 59 Up to 9,600 baud, 8 data bits, 1 stop bit, no parity

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Canada (Ontario)	905 785 0085	905 785 0086
Canada (Quebec)	514 694 8521	514 694 4399
Denmark	45 76 26 00	45 76 26 02
Finland	09 725 725 11	09 725 725 55
France	01 48 14 24 24	01 48 14 24 14
Germany	089 741 31 30	089 714 60 35
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Italy	02 413091	02 41309215
Japan	03 5472 2970	03 5472 2977
Korea	02 596 7456	02 596 7455
Mexico	5 520 2635	5 520 3282
Netherlands	0348 433466	0348 430673
Norway	32 84 84 00	32 84 86 00
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Sweden	08 730 49 70	08 730 43 70
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## **Technical Support Form**

Photocopy this form and update it each time you make changes to your software or hardware, and use the completed copy of this form as a reference for your current configuration. Completing this form accurately before contacting National Instruments for technical support helps our applications engineers answer your questions more efficiently.

If you are using any National Instruments hardware or software products related to this problem, include the configuration forms from their user manuals. Include additional pages if necessary. Name Company Address Fax ( \_\_\_ ) \_\_\_\_\_Phone ( \_\_\_ ) \_\_\_\_\_ Computer brand Model Processor Operating system (include version number) Clock speed \_\_\_\_\_MHz RAM \_\_\_\_MB Display adapter \_\_\_\_\_ Mouse \_\_\_\_yes \_\_\_\_no Other adapters installed \_\_\_\_\_\_ Hard disk capacity \_\_\_\_\_MB Brand \_\_\_\_\_ Instruments used \_\_\_\_\_\_ National Instruments hardware product model Revision Configuration \_\_\_\_\_ National Instruments software product \_\_\_\_\_\_ Version \_\_\_\_\_ Configuration \_\_\_\_\_ The problem is: \_\_\_\_\_ List any error messages: The following steps reproduce the problem:

# Lab-PC-1200/AI Hardware and Software Configuration Form

Record the settings and revisions of your hardware and software on the line to the right of each item. Complete a new copy of this form each time you revise your software or hardware configuration, and use this form as a reference for your current configuration. Completing this form accurately before contacting National Instruments for technical support helps our applications engineers answer your questions more efficiently.

## **National Instruments Products**

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Interrupt level of hardware
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Edition Date: December 1997

**Part Number:** 341309A-01

Please comment on the completeness, clarity, and organization of the manual.

If you find errors in the manual, please record the page numbers and describe the errors.

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Prefix	Meanings	Value
p-	pico-	10-12
n-	nano-	10-9
μ-	micro-	10-6
m-	milli-	10-3
k-	kilo-	103
M-	mega-	106
G-	giga-	109

## Numbers/Symbols

%	percent
+	positive of, or plus
-	negative of, or minus
/	per
0	degree
Ω	ohm
2SDAC	two's complement DAC bit
_	
Α	
А	amperes
A/D	analog-to-digital
ADC	analog-to-digital converter-an electronic device, often an integrated
	circuit, that converts an analog voltage to a digital number
ADC resolution	the resolution of the ADC, which is measured in bits. An ADC with
	16 bits has a higher resolution, and thus a higher degree of accuracy, than a
	12-bit ADC.
ADCUNI/BI*	ADC unipolar/bipolar bit
address	character code that identifies a specific location (or series of locations) in
	memory
AI	analog input
ANSI	American National Standards Institute

### Glossary

В	
b	bit—one binary digit, either 0 or 1
B	byte—eight related bits of data, an eight-bit binary number. Also used
2	to denote the amount of memory required to store one byte of data.
BCD	binary coded decimal
binary	a number system with a base of 2
bipolar	a signal range that includes both positive and negative values (for example, $-5 \text{ V}$ to $+5 \text{ V}$ )
bus	the group of conductors that interconnect individual circuitry in a computer. Typically, a bus is the expansion vehicle to which I/O or other devices are connected. Examples of PC buses are the ISA and PCI
	bus.
С	
C	Celsius
CALDAC	calibration DAC
CALDACLD	calibration DAC load bit
CMOS	complementary metal-oxide semiconductor
CNTINT	counter interrupt status bit
CNTINTEN	counter interrupt enable bit
CW	control word
D	
D/A	digital-to-analog
DAC	digital-to-analog converter—an electronic device, often an integrated
DAC	circuit, that converts a digital number into a corresponding analog voltage or current
DAC0UNI/BI*	DAC0 unipolar/bipolar bit
DAC1UNI/BI*	DAC1 unipolar/bipolar bit
DAQ	data acquisition—(1) collecting and measuring electrical signals from
	sensors, transducers, and test probes or fixtures and inputting them to a computer for processing; (2) collecting and measuring the same kinds
	of electrical signals with A/D and/or DIO boards plugged into a computer, and possibly generating control signals with D/A and/or DIO
	boards in the same computer
DAVAIL	data available bit
dB	decibel—the unit for expressing a logarithmic measure of the ratio of
	two signal levels: dB=20log10 V1/V2, for signals in volts
DC	two signal levels: dB=20log10 V1/V2, for signals in volts direct current
DC DIFF	direct current differential mode—an analog input consisting of two terminals, both of
	direct current

DITHEREN DMA DOS DQINTEN	dither enable bit direct memory access—a method by which data can be transferred to/from computer memory from/to a device or memory on the bus while the processor does something else. DMA is the fastest method of transferring data to/from computer memory. disk operating system DAQ interrupt enable bit
E	
ECLKDRV ECLKRCV EEPROM	external clock drive bit external clock receive bit electrically erasable programmable read-only memory—ROM that can be erased with an electrical signal and reprogrammed
EEPROMCS EOIRCV ERRINTEN EXTCONV* EXTGATA0	EEPROM chip select bit external output interval clock receive bit error interrupt enable bit external conversion signal external gate A0 bit
F	
FIFOHF* FIFOINTEN ft	first-in first-out memory buffer—the first data stored is the first data sent to the acceptor. FIFOs are often used on DAQ devices to temporarily store incoming or outgoing data until that data can be retrieved or output. For example, an analog input FIFO stores the results of A/D conversions until the data can be retrieved into system memory, a process that requires the servicing of interrupts and often the programming of the DMA controller. This process can take several milliseconds in some cases. During this time, data accumulates in the FIFO for future retrieval. With a larger FIFO, longer latencies can be tolerated. In the case of analog output, a FIFO permits faster update rates, because the waveform data can be stored on the FIFO ahead of time. This again reduces the effect of latencies associated with getting the data from system memory to the DAQ device. FIFO half-full bit FIFO interrupt enable bit feet
G	
gain	the factor by which a signal is amplified, sometimes expressed in decibels
GAIN GATA0	gain bits gate A0 bit

## Η

h	hour
hex	hexadecimal
HFINTEN	FIFO half-full interrupt enable bit
HWTRIG	hardware trigger bit
Hz	hertz
I	
IEEE	Institute of Electrical and Electronics Engineers
in.	inches
INTSCAN	interval scanning bit
I/O	input/output—the transfer of data to/from a computer system involving
10	communications channels, operator interface devices, and/or data
	acquisition and control interfaces
IRQ	interrupt request
ISA	
ISA	industry standard architecture
К	
k	kilo—the standard metric prefix for 1,000, or $10^3$ , used with units of
ĸ	measure such as volts, hertz, and meters
К	kilo—the prefix for 1,024, or $2^{10}$ , used with B in quantifying data or
ĸ	
1-h	computer memory
kbytes/s	a unit for data transfer that means 1,000 or $10^3$ bytes/s
kS	1,000 samples
Kword	1,024 words of memory
L	
LDAC	Load DAC bit
LSB	least significant bit
Μ	
m	meters
М	(1) Mega, the standard metric prefix for 1 million or $10^6$ , when used with
	units of measure such as volts and hertz; (2) mega, the prefix for 1,048,576,
	or $2^{20}$ , when used with B to quantify data or computer memory
МА	multiplexer select bits
MB	megabytes of memory
Mbytes/s	a unit for data transfer that means 1 million or 10 <sup>6</sup> bytes/s
•	•
MS	million samples
MSB	most significant bit

## Ν

Ν	
NRSE	nonreferenced single-ended mode—all measurements are made with respect to a common (NRSE) measurement system reference, but the voltage at this reference can vary with respect to the measurement system ground
0	
operating system	base-level software that controls a computer, runs programs, interacts with users, and communicates with installed hardware or peripheral devices
OUT	counter output
OUTA1	output A1 bit
OVERFLOW	overflow error status bit
OVERRUN	overrun error status bit
Ρ	
Plug and Play devices	devices that do not require dip switches or jumpers to configure resources on the devices—also called switchless devices
Plug and Play ISA	a specification prepared by Microsoft, Intel, and other PC-related companies that result in PCs with plug-in boards that can be fully configured in software, without jumpers or switches on the boards
PRETRIG	pretrigger bit
PROMOUT	EEPROM output bit
R	
RAM	random-access memory
resolution	the smallest signal increment that can be detected by a measurement system. Resolution can be expressed in bits, in proportions, or in percent of full scale. For example, a system has 12-bit resolution, one part in 4,096 resolution, and 0.0244% of full scale.
RL	read/load select bit
rms	root mean square
RSE	referenced single-ended mode—all measurements are made with respect to a common reference measurement system or a ground. Also called a grounded measurement system.
RSE*/NRSE	referenced single-ended/nonreferenced single-ended bit
S	
s	seconds
S	samples
SC	counter select bit
SCANEN	scan enable bit

Glossary

SCANUP SCLK SDATA SE*/DIFF S/s SWTRIG	scan up bit serial clock bit serial data bit single-ended/differential bit samples per second—used to express the rate at which a DAQ board samples an analog signal software trigger bit
т	
TBSEL	time base select bit
TTL	transistor-transistor logic
TWOSCOMP	two's complement bit
U	
unipolar	a signal range that is always positive (for example, 0 to $+10$ V)
V	
V	volts
W	
WRTPRT*	write protect bit
Х	
Х	don't care bit

# Index

## Numbers

2SDAC0 bit, 2-7 to 2-8 2SDAC1 bit, 2-7 82C53 Counter/Timer Register Groups A and B configuration, 1-3 Counter A Mode Register description, 2-28 performing single A/D conversion, 3-7 programming update mode of analog output circuitry, 3-19 to 3-20 Counter A0 Data Register DAQ operation using internal timing, 3-8 description, 2-27 performing single A/D conversion, 3-7 programming, 3-10 Counter A1 Data Register DAQ operations using internal timing, 3-8 description, 2-27 programming, 3-11 Counter A2 Data Register description, 2-28 programming update mode of analog output circuitry, 3-20 Counter B Mode Register, 2-31 Counter B0 Data Register DAQ operations using internal timing, 3-9 description, 2-29 programming, 3-10 Counter B1 Data Register DAQ operations using internal timing, 3-8 to 3-9 description, 2-30 programming, 3-11 to 3-12 Counter B2 Data Register, 2-30

OKI MSM82C53 data sheet, C-1 overview, 2-26 programming general-purpose counter/timers, 3-19 to 3-20 register map, 2-2 Timer Interrupt Clear Register DAC interrupt programming, 3-20 description, 2-29 82C55A Digital I/O Register Group Digital Control Register, 2-34 OKI MSM82C55A data sheet, D-1 overview, 2-32 Port A Register, 2-33 Port B Register, 2-33 Port C Register, 2-34 register map, 2-3

## A

A/D conversions, single. See single A/D conversions, programming. A/D FIFO Clear Register clearing analog input circuitry, 3-5 description, 2-23 programming DAQ operation, 3-14 A/D FIFO Register description, 2-21 to 2-22 obtaining A/D conversion results, 3-7 overflow condition, 3-13 servicing DAQ operation, 3-12 to 3-13 straight binary mode, 2-21 to 2-22 two's complement binary mode, 2-22 ADCUNI/BI\* bit configuring analog input circuitry, 3-6 description, 2-16 analog input circuitry calibration, 4-4 to 4-7

bipolar input calibration procedure, 4-5 to 4-6 unipolar input calibration procedure, 4-6 to 4-7 configuration, 1-2 programming for single A/D conversions, 3-4 to 3-7 clearing analog input circuitry, 3-4 to 3-5 configuring analog input circuitry, 3-5 to 3-6 performing single A/D conversions, 3-7 Analog Input Register Group A/D FIFO Clear Register clearing analog input circuitry, 3-4 to 3-5 description, 2-23 programming DAQ operation, 3-14 A/D FIFO Register description, 2-21 to 2-22 obtaining A/D conversion results, 3-7 overflow condition, 3-13 servicing DAO operation, 3-12 to 3-13 two's complement binary mode, 2-22 DMATC Interrupt Clear Register, 2-23 overview, 2-20 register map, 2-2 Start Convert Register description, 2-23 performing single A/D conversions, 3-7 analog output circuitry calibration, 4-8 to 4-10 bipolar output calibration procedure, 4-8 to 4-9 unipolar output calibration procedure, 4-9 to 4-10 configuration, 1-2 to 1-3 programming, 3-17 to 3-20

analog output voltage versus digital code bipolar mode, two's complement coding (table), 3-18 unipolar mode, straight binary coding (table), 3-18 configuring analog output circuitry, 3-17 to 3-18 DAC interrupt programming, 3-20 update mode of analog output circuitry, 3-18 to 3-20 Analog Output Register Group DAC0 Low-Byte, DAC0 High-Byte, DAC1 Low-Byte, and DAC1 **High-Byte Registers** bipolar output calibration procedure, 4-9 DAC interrupt programming, 3-19 description, 2-25 programming update mode of analog output circuitry, 3-19 to 3-20 unipolar output calibration procedure, 4-10 overview, 2-24 register map, 2-2 analog output voltage versus digital code bipolar mode, two's complement coding (table), 3-18 unipolar mode, straight binary coding (table), 3-18

## B

bipolar input calibration, 4-5 to 4-6 gain calibration, 4-6 higher gains, 4-6 postgain offset calibration, 4-6 pregain offset coarse calibration, 4-5 pregain offset fine calibration, 4-5 bipolar input polarity, selecting, 3-5 to 3-6 bipolar mode, two's complement coding (table), 3-18 bipolar output calibration, 4-8 to 4-9 gain calibration, 4-9 offset calibration, 4-9 bipolar output polarity, selecting, 3-18 bits 2SDAC0, 2-7 to 2-8 2SDAC1, 2-7 ADCUNI/BI\*, 2-6, 3-16 CALDACLD, 2-13, 3-4 CNTINT, 2-17, 3-20 CNTINTEN, 2-9, 3-19, 3-20 CW<7..0>, 2-34 D<7..0> A/D FIFO Register, 2-22 Counter A0 Data Register, 2-27 Counter A1 Data Register, 2-27 Counter A2 Data Register, 2-28 Counter B0 Data Register, 2-29 Counter B1 Data Register, 2-30 Counter B2 Data Register, 2-30 DAC0 Low-Byte, DAC0 High\_Byte, DAC1 Low-Byte, and DAC1 High-Byte Registers, 2-25 Interval Counter Data Register, 2-36 Port A Register, 2-33 Port B Register, 2-33 Port C Register, 2-34 D<11..8>, 2-21 to 2-22, 2-25 D<15..8>, 2-22 D<15..12>, 2-25 DAC0UNI/BI\*, 2-16, 3-17 DAC1UNI/BI\*, 2-15, 3-17 DAVAIL, 2-18, 3-5, 3-7 DIOINTEN, 2-10 DITHEREN, 2-13, 4-5 DMAEN, 2-10, 3-17 DMATC, 2-17 DQINTEN, 2-15, 3-16 ECLKDRV, 2-12 ECLKRCV, 2-11, 3-5, 3-14 EEPROMCS, 2-13, 3-3, 4-2

EOIRCV, 2-12, 3-12, 3-15 ERRINTEN, 2-9, 3-16 EXTGATA0, 2-17 FIFOHF\*, 2-19, 3-5, 3-7 FIFOINTEN, 2-9, 3-16 GAIN<2..0>, 2-5, 3-6 GATA0, 2-17 HFINTEN, 2-15, 3-16 HWTRIG, 2-8, 3-5, 3-14 INTSCAN, 2-12, 3-12, 3-15 LDAC0, 2-7, 3-19 LDAC1, 2-7, 3-19 MA<2..0>, 2-6, 3-6 OUTA1, 2-19 OVERFLOW, 2-18, 3-5, 3-7, 3-13 OVERRUN, 2-18, 3-5, 3-7, 3-13 PRETRIG, 2-8, 3-15 PROMOUT, 2-19, 3-4 RSE\*/NRSE, 2-16, 3-6, 4-5 SCANEN, 2-5, 3-6 SCANUP, 2-15, 3-6 SCLK, 2-13, 3-3 to 3-4, 4-2 SDATA, 2-13, 4-2 SE\*/DIFF, 2-11, 3-6, 4-5 SWTRIG, 2-8, 3-4 to 3-5, 3-12, 3-15 TBSEL, 2-8, 3-10 TCINTEN, 2-9 to 2-10 TWOSCMP, 2-5 to 2-6, 3-6 WRTPRT\*, 2-14, 3-3, 4-2 board configuration, 1-2 to 1-3 bulletin board support, E-1

## C

CALDACLD bit description, 2-13 writing calibration constant to CALDAC, 3-4 calibration analog input calibration, 4-4 to 4-7

bipolar input calibration procedure, 4-5 to 4-6 unipolar input calibration procedure, 4-6 to 4-7 analog output calibration, 4-8 to 4-10 bipolar output calibration procedure, 4-8 to 4-9 unipolar output calibration procedure, 4-9 to 4-10 calibration DACs (CALDACs), 4-3 analog input circuitry characteristics (table), 4-3 analog output circuitry characteristics (table), 4-3 writing calibration constant to CALDAC, 3-4 EEPROM map (table), 4-10 to 4-17 initializing Lab-PC-1200/AI circuitry, 3-3 to 3-4 storing user-defined constants, 4-1 to 4-3 CLKB1 signal, 3-11 CNTINT bit DAC interrupt programming, 3-20 description, 2-17 **CNTINTEN** bit DAC interrupt programming, 3-20 description, 2-9 programming update mode of analog output circuitry, 3-19 Command Register 1 configuring analog input circuitry, 3-6 description, 2-5 to 2-6 **Command Register 2** clearing analog input circuitry, 3-4 to 3-5 description, 2-7 to 2-8 programming DAQ operation, 3-14 programming update mode of analog output circuitry, 3-19 Command Register 3 DAC interrupt programming, 3-20 DAQ interrupt programming, 3-16

description, 2-9 to 2-10 Command Register 4 configuring analog input circuitry, 3-6 description, 2-11 to 2-12 programming DAQ operations, 3-14, 3-15 Command Register 5 calibration process, 3-3 to 3-4 description, 2-13 to 2-14 Command Register 6 configuring analog input circuitry, 3-6 DAQ interrupt programming, 3-16 description, 2-15 to 2-16 configuration, 1-2 to 1-3 analog input configuration, 1-2 analog output configuration, 1-2 to 1-3 counter configuration, 1-3 digital I/O configuration, 1-3 Configuration and Status Register Group Command Register 1 configuring analog input circuitry, 3-6 description, 2-5 to 2-6 **Command Register 2** clearing analog input circuitry, 3-4 to 3-5 description, 2-7 to 2-8 programming DAQ operation, 3-14 programming update mode of analog output circuitry, 3-19 **Command Register 3** DAC interrupt programming, 3-20 DAQ interrupt programming, 3-16 description, 2-9 to 2-10 **Command Register 4** configuring analog input circuitry, 3-6 description, 2-11 to 2-12 programming DAQ operations, 3-14 **Command Register 5** calibration process, 3-3 to 3-4 description, 2-13 to 2-14

Command Register 6 configuring analog input circuitry, 3-6 DAQ interrupt programming, 3-16 description, 2-15 to 2-16 overview, 2-4 register map, 2-2 Status Register 1 description, 2-17 to 2-18 servicing DAQ operation, 3-13 Status Register 2, 2-19 controlled acquisition mode, 3-8 Counter A Mode Register description, 2-28 performing single A/D conversion, 3-7 programming update mode of analog output circuitry, 3-19 to 3-20 Counter A0 Data Register DAQ operation using internal timing, 3-8 description, 2-27 performing single A/D conversion, 3-7 programming, 3-10 Counter A1 Data Register DAQ operations using internal timing, 3-8 description, 2-27 programming, 3-11 Counter A2 Data Register description, 2-28 programming update mode of analog output circuitry, 3-19 Counter B Mode Register, 2-31 Counter B0 Data Register DAQ operations using internal timing, 3-9 description, 2-29 programming, 3-10 Counter B1 Data Register DAQ operations using internal timing, 3-8 to 3-9 description, 2-30 programming, 3-11 to 3-12 Counter B2 Data Register, 2-30

counter/timers. *See* 82C53 Counter/Timer Register Groups A and B; 82C55A Digital I/O Register Group. customer communication, *xii*, E-1 to E-2 CW<7..0> bits, 2-34

## D

D < 7..0 > bitsA/D FIFO Register, 2-22 Counter A0 Data Register, 2-27 Counter A1 Data Register, 2-27 Counter A2 Data Register, 2-28 Counter B0 Data Register, 2-29 Counter B1 Data Register, 2-30 Counter B2 Data Register, 2-30 DAC0 Low-Byte, DAC0 High-Byte, DAC1 Low-Byte, and DAC1 High-Byte Registers, 2-25 Interval Counter Data Register, 2-36 Port A Register, 2-33 Port B Register, 2-33 Port C Register, 2-34 D<11..8> bits A/D FIFO Register, 2-21 to 2-22 DAC0 Low-Byte, DAC0 High-Byte, DAC1 Low-Byte, and DAC1 High-Byte Registers, 2-25 D<15..8> bits, 2-22 D<15..12> bits. 2-25 DAC interrupt programming, 3-20 DAC0 Low-Byte, DAC0 High-Byte, DAC1 Low-Byte, and DAC1 High-Byte Registers bipolar output calibration procedure, 4-9 DAC interrupt programming, 3-20 description, 2-25 programming update mode of analog output circuitry, 3-19 to 3-20 unipolar output calibration procedure, 4 - 10DAC0UNI/BI\* bit

configuring analog output circuitry, 3-17 description, 2-16 DAC1UNI/BI\* bit configuring analog output circuitry, 3-17 description, 2-15 DAQ interrupt programming, 3-16 DAQ operation, programming controlled data acquisition mode, 3-8 definition of DAQ operation, 3-8 freerun data acquisition mode, 3-8 interval scanning data acquisition mode, 3-8 to 3-9 using external timing, 3-13 to 3-15 EXTCONV\* instead of Counter A0, 3-14 EXTTRIG in posttrigger mode, 3-14 EXTTRIG in pretrigger mode, 3-15 OUTB1 instead of Counter B1, 3-15 using internal timing, 3-8 to 3-13 Counter A0 and Counter B0, 3-10 Counter A1. 3-11 Counter B1 and Interval Counter Register, 3-11 to 3-12 servicing DAQ operation, 3-12 to 3-13 triggering DAQ operation, 3-12 DAVAIL bit clearing analog input circuitry, 3-5 description, 2-18 single A/D conversions, 3-7 Digital Control Register, 2-34 digital I/O circuitry configuration, 1-3 programming, 3-20 DIOINTEN bit, 2-10 **DITHEREN** bit analog input calibration, 4-5 description, 2-13 DMA programming, 3-17 DMAEN bit DAQ DMA programming, 3-16

description, 2-10 DMATC bit, 2-17 DMATC Interrupt Clear Register, 2-23 documentation conventions used in manual, *x-xi* National Instruments documentation, *xi-xii* organization of manual, *ix-x* related documentation, *xii* DQINTEN bit DAQ interrupt programming, 3-16 description, 2-15

## E

ECLKDRV bit, 2-12 ECLKRCV bit clearing analog input circuitry, 3-5 description, 2-11 programming DAQ operation, 3-14 EEPROM accidental overwriting (note), 4-1 calibration map (table), 4-10 to 4-17 reading single byte from EEPROM, 3-7 storing user-defined constants, 4-2 writing single byte to EEPROM, 3-7 EEPROMCS bit description, 2-13 reading single byte from EEPROM, 3-4 storing user-defined constants, 4-2 e-mail support, E-2 EOIRCV bit DAQ operations using OUTB1, 3-15 description, 2-12 programming counter B1, 3-12 ERRINTEN bit DAQ interrupt programming, 3-16 description, 2-9 EXTCONV\* signal clearing analog input circuitry, 3-4 to 3-5 programming DAQ operation, 3-14

EXTGATA0 bit, 2-17 EXTTRIG signal clearing analog input circuitry, 3-5 programming DAQ operation posttrigger mode, 3-14 pretrigger mode, 3-15 EXTUPDATE\* signal DAC interrupt programming, 3-20 programming update mode of analog output circuitry, 3-19 to 3-20

## F

fax and phone numbers for technical support, E-2 Fax-on-Demand support, E-2 FIFOHF\* bit clearing analog input circuitry, 3-5 description, 2-19 single A/D conversions, 3-7 FIFOINTEN bit DAQ interrupt programming, 3-16 description, 2-9 freerun data acquisition mode, 3-8 FTP support, E-1 Fujitsu MB88341/MB88342 data sheet, A-1

## G

GAIN<2..0> bits configuring analog input circuitry, 3-6 description, 2-5
GATA0 bit, 2-17
GATB0 signal, 3-9
GATB1 signal, 3-9
general-purpose counter/timers, programming, 3-21

## H

HFINTEN bit

DAQ interrupt programming, 3-16 description, 2-15 HWTRIG bit clearing analog input circuitry, 3-5 description, 2-8 programming DAQ operation, 3-14

## I

initializing Lab-PC-1200/AI circuitry, 3-3 to 3-4 input mode, selecting, 3-5 input polarity, selecting, 3-5 to 3-6 interrupts DAC interrupt programming, 3-20 DAQ interrupt programming, 3-16 Interval Counter Register Group DAQ operation using internal timing, 3-8 to 3-9 Interval Counter Data Register, 2-36 Interval Counter Strobe Register, 2-36 overview, 2-35 programming, 3-11 to 3-12 register map, 2-3 interval scanning acquisition mode, 3-8 to 3-9 **INTSCAN** bit DAQ operations using OUTB1, 3-15 description, 2-12 programming Counter B1, 3-12

## L

Lab-PC-1200/AI assigning resources, 3-2 companion disk, 3-2 general characteristics, 1-1 initializing circuitry, 3-3 to 3-4 LDAC0 bit description, 2-7 programming update mode of analog output circuitry, 3-19 LDAC1 bit description, 2-7 programming update mode of analog output circuitry, 3-19

## Μ

MA<2..0> bits configuring analog input circuitry, 3-6 description, 2-6 manual. *See* documentation.

## 0

OKI MSM82C53 data sheet, C-1 OKI MSM82C55A data sheet, D-1 **OUTA0** signal performing single A/D conversions, 3-7 programming Counter A0 and Counter B0. 3-10 OUTA1 bit, 2-19 **OUTA2** signal DAC interrupt programming, 3-20 programming update mode of analog output circuitry, 3-19 OUTB1 for programming DAQ operation, 3-15 **OVERFLOW** bit clearing analog input circuitry, 3-5 description, 2-18 overflow condition in DAQ operations, 3-13 single A/D conversions, 3-7 **OVERRUN** bit clearing analog input circuitry, 3-5 description, 2-18 overflow condition in DAQ operations, 3-13 single A/D conversions, 3-7

## Ρ

polarity, selecting analog input circuitry, 3-5 to 3-6 analog output circuitry, 3-17 to 3-18 Port A Register, 2-33 Port B Register, 2-33 Port C Register, 2-34 PRETRIG bit description, 2-8 programming DAQ operation in pretrigger mode, 3-15 programming. See also registers. analog input circuitry for single A/D conversions, 3-4 to 3-7 clearing analog input circuitry, 3-4 to 3-5 configuring analog input circuitry, 3-5 to 3-6 performing single A/D conversions, 3-7 analog output circuitry, 3-17 to 3-20 configuring analog output circuitry, 3-17 to 3-18 DAC interrupt programming, 3-20 update mode of analog output circuitry, 3-18 to 3-20 assigning Lab-PC-1200/AI resources, 3-2 DAQ DMA programming, 3-17 DAQ interrupt programming, 3-16 DAQ operation using external timing, 3-13 to 3-15 EXTCONV\* instead of Counter A0, 3 - 14EXTTRIG in posttrigger mode, 3-14 EXTTRIG in pretrigger mode, 3-15 OUTB1 instead of Counter B1, 3-15 DAO operation using internal timing, 3-8 to 3-13 Counter A0 and Counter B0, 3-10 Counter A1, 3-11

Counter B1 and Interval Counter Register, 3-11 to 3-12 servicing DAQ operation, 3-12 to 3-13 triggering DAQ operation, 3-12 digital I/O circuitry, 3-20 examples, 3-2 to 3-3 general-purpose counter/timers, 3-21 initializing Lab-PC-1200/AI circuitry, 3-3 to 3-4 Lab-PC-1200/AI companion disk, 3-2 Lab-PC-1200/AI companion diskette, 3-2 register programming considerations, 3-1 PROMOUT bit description, 2-19 reading from EEPROM, 3-7

## R

registers. *See also* specific register groups. programming considerations, 3-1. *See also* programming. register map, 2-2 to 2-3 resetting Lab-PC-1200/AI circuitry, 3-3 RSE\*/NRSE bit analog input calibration, 4-5 configuring analog input circuitry, 3-6 description, 2-16

## S

sample interval, defined, 3-8 scan interval, defined, 3-8 SCANEN bit configuring analog input circuitry, 3-6 description, 2-5 SCANUP bit configuring analog input circuitry, 3-6 description, 2-15 SCLK bit description, 2-13

reading single byte from EEPROM, 3-4 storing user-defined constants, 4-2 writing bits to CALDAC, 3-4 writing single byte to EEPROM, 3-3 SDATA bit description, 2-13 storing user-defined constants, 4-2 SE\*/DIFF bit analog input calibration, 4-5 configuring analog input circuitry, 3-6 description, 2-11 single A/D conversions, programming, 3-7 clearing analog input circuitry, 3-4 to 3-5 configuring analog input circuitry, 3-5 to 3-6 performing single A/D conversions, 3-7 Start Convert Register description, 2-23 performing single A/D conversions, 3-7 Status Register 1 description, 2-17 to 2-18 servicing DAQ operation, 3-13 Status Register 2, 2-19 SWTRIG bit clearing analog input circuitry, 3-4 to 3-5 description, 2-8 programming DAQ operation in pretrigger mode, 3-15 triggering DAQ operation, 3-12

## Τ

TBSEL bit DAQ operations using internal timing, 3-10 description, 2-8 TCINTEN bit, 2-9 to 2-10 technical support, E-1 to E-2 telephone and fax numbers for technical support, E-2 Timer Interrupt Clear Register DAC interrupt programming, 3-20 description, 2-29 triggering DAQ operation, 3-12 TWOSCMP bit configuring analog input circuitry, 3-6 description, 2-5 to 2-6

## U

unipolar input calibration, 4-6 to 4-7 gain calibration, 4-7 postgain offset calibration, 4-7 unipolar input polarity, selecting, 3-5 to 3-6 unipolar mode, straight binary coding (table), 3-18 unipolar output calibration, 4-9 to 4-10 gain calibration, 4-10 offset calibration, 4-10 unipolar output polarity, selecting, 3-17 update mode of analog output circuitry, programming, 3-18 to 3-20 user-defined calibration constants, 4-1 to 4-3

## W

WRTPRT\* bit description, 2-14 reading single byte from EEPROM, 3-7 storing user-defined constants, 4-2

## X

Xicor X25020 data sheet, B-1